

# GaNFast™ Power IC



## 1. Features

### GaNFast™ Power IC

- Monolithically-integrated gate drive
- Wide logic input range with hysteresis
- 5 V / 15 V input-compatible
- Wide  $V_{CC}$  range (10 to 30 V)
- Programmable turn-on  $dV/dt$
- 200 V/ns  $dV/dt$  immunity
- 800 V Transient Voltage Rating
- 650 V Continuous Voltage Rating
- Low 170 m $\Omega$  resistance
- Zero reverse recovery charge
- ESD protection – 2 kV (HBM), 1 kV (CDM)
- 2 MHz operation

### Small, low-profile SMT QFN

- 6 x 8 mm footprint, 0.85 mm profile
- Minimized package inductance

### Sustainability

- RoHS, Pb-free, REACH-compliant
- Up to 40% energy savings vs Si solutions
- System level 4kg CO<sub>2</sub> Carbon Footprint reduction

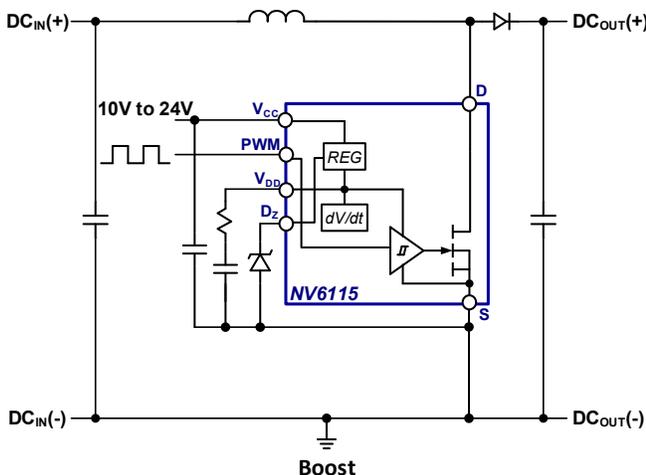
### Product Reliability

- 20-year limited product warranty (see Section 14 for details)

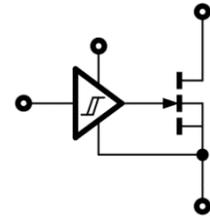
## 2. Topologies / Applications

- AC-DC, DC-DC, DC-AC
- QR Flyback, PFC, AHB, Buck, Boost, Half bridge, Full bridge, LLC resonant, Class D
- Wireless power, Solar Micro-inverters, LED lighting, TV SMPS, Server, Telecom

## 4. Typical Application Circuits



QFN 5 x 6 mm



Simplified schematic

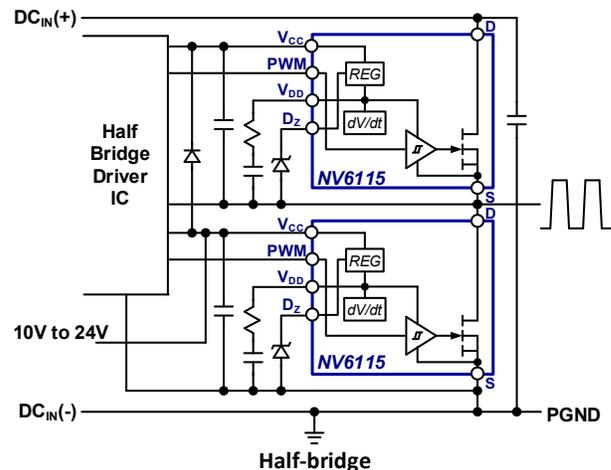
## 3. Description

This GaNFast power IC is optimized for high frequency, soft-switching topologies.

Monolithic integration of FET, drive and logic creates an easy-to-use ‘digital-in, power-out’ high-performance powertrain building block, enabling designers to create the fastest, smallest, most efficient power converters in the world.

The highest  $dV/dt$  immunity, high-speed integrated drive and industry-standard low-profile, low-inductance, 5 x 6 mm SMT QFN package allow designers to exploit Navitas GaN technology with simple, quick, dependable solutions for breakthrough power density and efficiency.

GaNFast power ICs extend the capabilities of traditional topologies such as flyback, half-bridge, resonant, etc. to MHz+ and enable the commercial introduction of breakthrough designs.



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## 6. Specifications

### 6.1. Absolute Maximum Ratings<sup>(1)</sup>

(with respect to Source (pad) unless noted)

SYMBOL	PARAMETER	MAX	UNITS
$V_{DS(TRAN)}$	Transient Drain-to-Source Voltage <sup>(2)</sup>	800	V
$V_{DS(CONT)}$	Continuous Drain-to-Source Voltage	-7 to +650	V
$V_{CC}$	Supply Voltage	30	V
$V_{PWM}$	PWM Input Pin Voltage	-3 to +30	V
$V_{DZ}$	$V_{DD}$ Setting Pin Voltage	6.6	V
$V_{DD}$	Drive Supply Voltage	7.2	V
$I_D$	Continuous Drain Current (@ $T_C = 100^\circ\text{C}$ )	8	A
$I_D$ PULSE	Pulsed Drain Current (10 $\mu\text{s}$ @ $T_J = 25^\circ\text{C}$ )	16	A
$I_D$ PULSE	Pulsed Drain Current (10 $\mu\text{s}$ @ $T_J = 125^\circ\text{C}$ )	12	A
dV/dt	Slew Rate on Drain-to-Source	200	V/ns
$T_J$	Operating Junction Temperature	-55 to 150	$^\circ\text{C}$
$T_{STOR}$	Storage Temperature	-55 to 150	$^\circ\text{C}$

(1) Absolute maximum ratings are stress ratings; devices subjected to stresses beyond these ratings may cause permanent damage.

(2)  $V_{DS(TRAN)}$  allows for surge ratings during non-repetitive events that are <100 $\mu\text{s}$  (for example start-up, line interruption) and repetitive events that are <100ns (for example repetitive leakage inductance spikes). Refer to Section 8.10 for detailed recommended design guidelines.

### 6.2. Recommended Operating Conditions<sup>(3)</sup>

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS
$V_{DZ}$	Drive Supply Set Zener Voltage <sup>(4)</sup>	5.8	6.2	6.6	V
$V_{DD}$	Drive Supply Voltage	5.5		7.0	V
$I_{DD\_EXT}$	Regulator External Load Current			3.0	mA
$R_{DD}$	Gate Drive Turn-On Current Set Resistance <sup>(5)</sup>	10	25		$\Omega$
$V_{PWM}$	PWM Input Pin Voltage	0	5	Min. of $(V_{CC}$ or 20)	V
$V_{CC}$	Supply Voltage	10		24	V
$T_C$	Operating Case Temperature	-40		125	$^\circ\text{C}$

(3) Exposure to conditions beyond maximum recommended operating conditions for extended periods of time may affect device reliability.

(4) Use of Zener diode other than 6.2 V is not recommended. See Table I for recommended part numbers of 6.2 V Zener diodes.

(5)  $R_{DD}$  resistor must be used. Minimum 10 Ohm to ensure application and device robustness.

### 6.3. ESD Ratings

SYMBOL	PARAMETER	MAX	UNITS
HBM	Human Body Model (per JS-001-2014)	2,000	V
CDM	Charged Device Model (per JS-002-2014)	1,000	V

### 6.4. Thermal Resistance

SYMBOL	PARAMETER	TYP	UNITS
$R_{\theta JC}^{(6)}$	Junction-to-Case	2.2	°C/W
$R_{\theta JA}^{(6)}$	Junction-to-Ambient	50	°C/W

(6)  $R_{\theta}$  measured on DUT mounted on 1 square inch 2 oz Cu (FR4 PCB)

## 6.5. Electrical Characteristics

Typical conditions:  $V_{DS} = 400\text{ V}$ ,  $V_{CC} = 15\text{ V}$ ,  $V_{DZ} = 6.2\text{ V}$ ,  $F_{SW} = 1\text{ MHz}$ ,  $T_{AMB} = 25\text{ °C}$ ,  $I_D = 4\text{ A}$ ,  $R_{DD} = 10\text{ }\Omega$  (or specified)

SYMBOL	PARAMETER	MIN	TYP	MAX	UNITS	CONDITIONS
<b><math>V_{CC}</math> Supply Characteristics</b>						
$I_{QCC}$	$V_{CC}$ Quiescent Current		0.85	1.5	mA	$V_{PWM} = 0\text{ V}$
$I_{QCC-SW}$	$V_{CC}$ Operating Current		2.3		mA	$F_{SW} = 1\text{ MHz}$ , $V_{DS} = \text{Open}$
<b>Low-Side Logic Input Characteristics</b>						
$V_{PVMH}$	Input Logic High Threshold (rising edge)			4	V	
$V_{PVML}$	Input Logic Low Threshold (falling edge)	1			V	
$V_{L-HYS}$	Input Logic Hysteresis		0.5		V	
$T_{ON}$	Turn-on Propagation Delay		11		ns	Fig.1, Fig.2
$T_{OFF}$	Turn-off Propagation Delay		9		ns	Fig.1, Fig.2
$T_R$	Drain rise time		6		ns	Fig.1, Fig.2
$T_F$	Drain fall time		3		ns	Fig.1, Fig.2
<b>Switching Characteristics</b>						
$F_{SW}$	Switching Frequency			2	MHz	
$t_{PW}$	Pulse width	0.02		1000	$\mu\text{s}$	
<b>GaN FET Characteristics</b>						
$I_{DSS}$	Drain-Source Leakage Current		0.2	25	$\mu\text{A}$	$V_{DS} = 650\text{ V}$ , $V_{PWM} = 0\text{ V}$
$I_{DSS}$	Drain-Source Leakage Current		7	50	$\mu\text{A}$	$V_{DS} = 650\text{ V}$ , $V_{PWM} = 0\text{ V}$ , $T_C = 125\text{ °C}$
$R_{DS(ON)}$	Drain-Source Resistance		170	240	m $\Omega$	$V_{PWM} = 6\text{ V}$ , $I_D = 4\text{ A}$
$R_{DS(ON)}$	Drain-Source Resistance		352		m $\Omega$	$V_{PWM} = 6\text{ V}$ , $I_D = 4\text{ A}$ , $T_C = 125\text{ °C}$
$V_{SD}$	Source-Drain Reverse Voltage		3.2	5	V	$V_{PWM} = 0\text{ V}$ , $I_{SD} = 4\text{ A}$
$Q_{OSS}$	Output Charge		16		nC	$V_{DS} = 400\text{ V}$ , $V_{PWM} = 0\text{ V}$
$Q_{RR}$	Reverse Recovery Charge		0		nC	
$C_{OSS}$	Output Capacitance		18		pF	$V_{DS} = 400\text{ V}$ , $V_{PWM} = 0\text{ V}$
$C_{O(er)}^{(7)}$	Effective Output Capacitance, Energy Related		24		pF	$V_{DS} = 400\text{ V}$ , $V_{PWM} = 0\text{ V}$
$C_{O(tr)}^{(8)}$	Effective Output Capacitance, Time Related		40		pF	$V_{DS} = 400\text{ V}$ , $V_{PWM} = 0\text{ V}$

(7)  $C_{O(er)}$  is a fixed capacitance that gives the same stored energy as  $C_{OSS}$  while  $V_{DS}$  is rising from 0 to 400 V

(8)  $C_{O(tr)}$  is a fixed capacitance that gives the same charging time as  $C_{OSS}$  while  $V_{DS}$  is rising from 0 to 400 V

### 6.6. Switching Waveforms

( $T_c = 25\text{ }^\circ\text{C}$  unless otherwise specified)

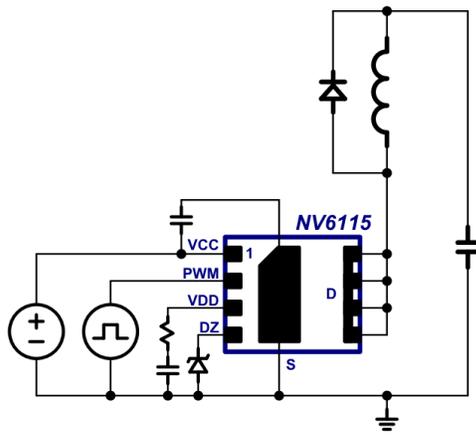


Fig. 1. Inductive switching circuit



Fig. 2. Propagation delay and rise/fall time definitions

### 6.7. Characteristic Graphs

(GaN FET,  $T_C = 25\text{ }^\circ\text{C}$  unless otherwise specified)

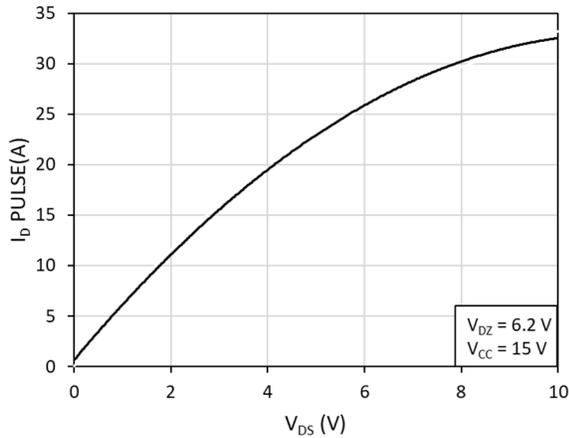


Fig. 3. Pulsed Drain current ( $I_D$  PULSE) vs. drain-to-source voltage ( $V_{DS}$ ) at  $T = 25\text{ }^\circ\text{C}$

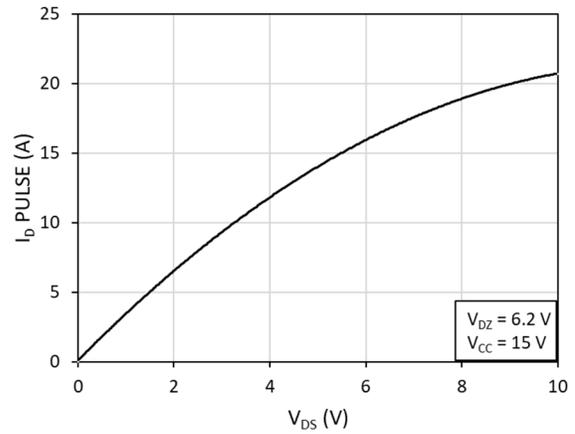


Fig. 4. Pulsed Drain current ( $I_D$  PULSE) vs. drain-to-source voltage ( $V_{DS}$ ) at  $T = 125\text{ }^\circ\text{C}$

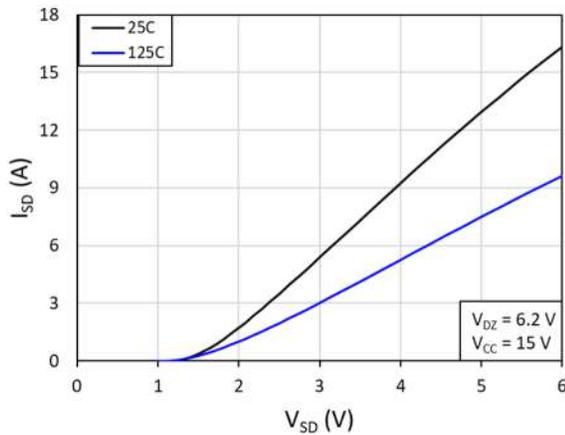


Fig. 5. Source-to-drain reverse conduction voltage

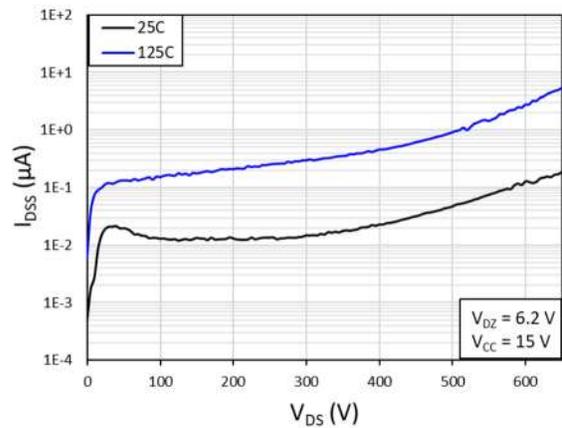


Fig. 6. Drain-to-source leakage current ( $I_{DSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

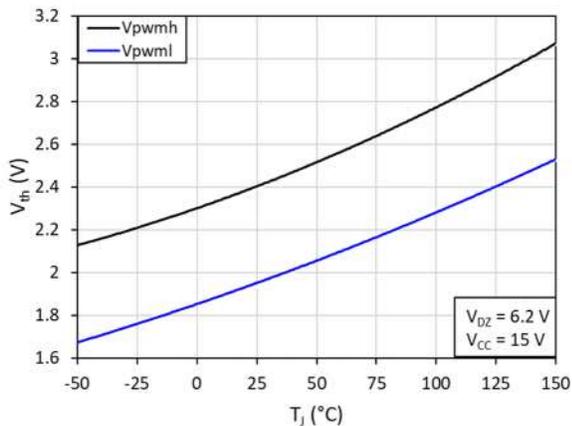


Fig. 7.  $V_{PWMH}$  and  $V_{PWMML}$  vs. junction temperature( $T_J$ )

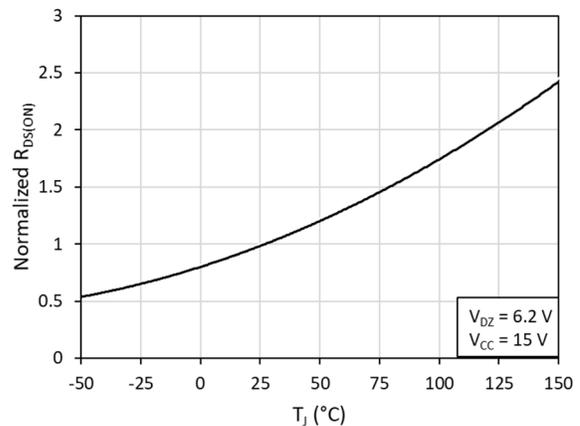


Fig. 8. Normalized on-resistance ( $R_{DS(ON)}$ ) vs. junction temperature ( $T_J$ )

Characteristic Graphs (Cont.)

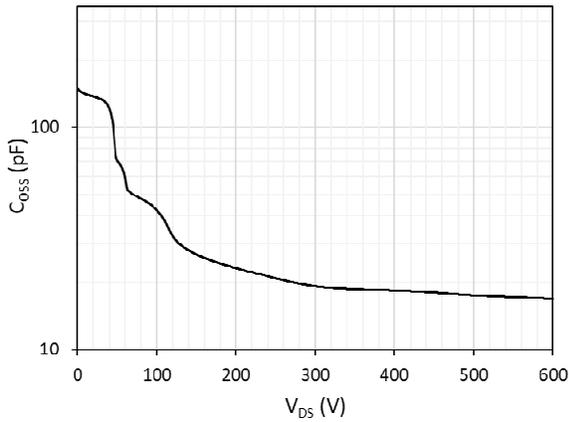


Fig. 9. Output capacitance ( $C_{OSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

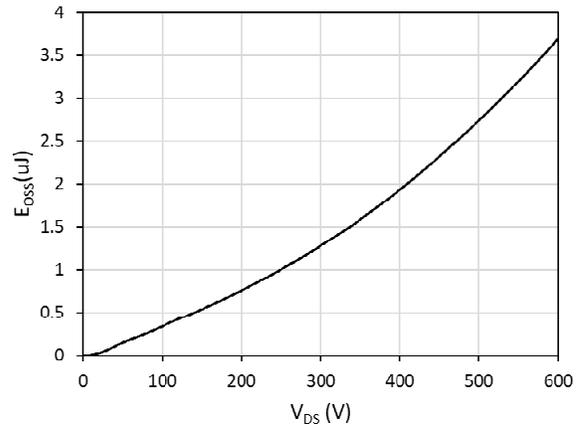


Fig. 10. Energy stored in output capacitance ( $E_{OSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

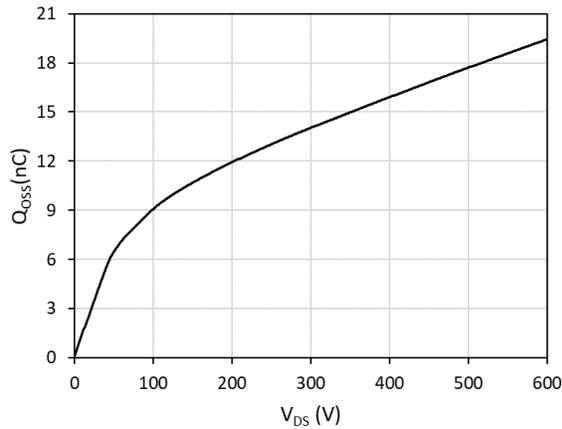


Fig. 11. Charge stored in output capacitance ( $Q_{OSS}$ ) vs. drain-to-source voltage ( $V_{DS}$ )

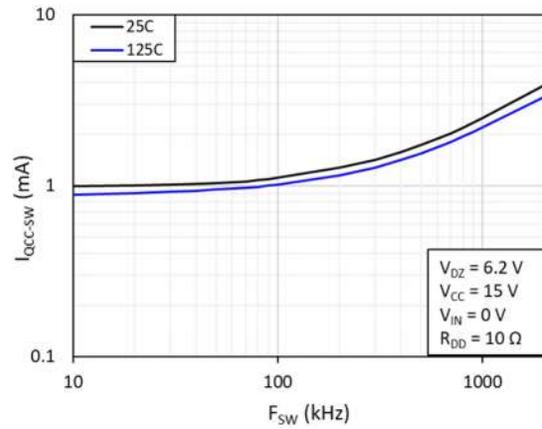


Fig. 12.  $V_{CC}$  operating current ( $I_{QCC-SW}$ ) vs. operating frequency ( $F_{SW}$ )

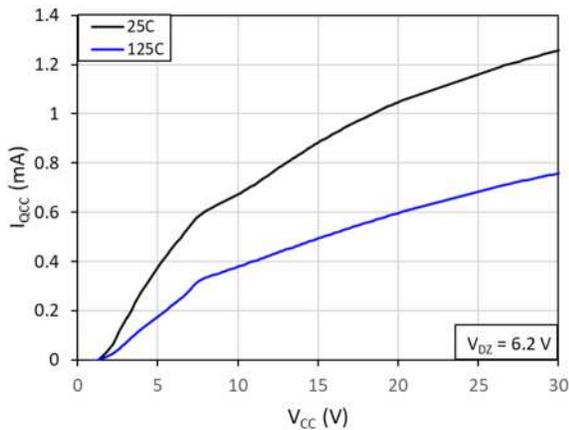


Fig. 13.  $V_{CC}$  quiescent current ( $I_{QCC}$ ) vs. supply voltage ( $V_{CC}$ )

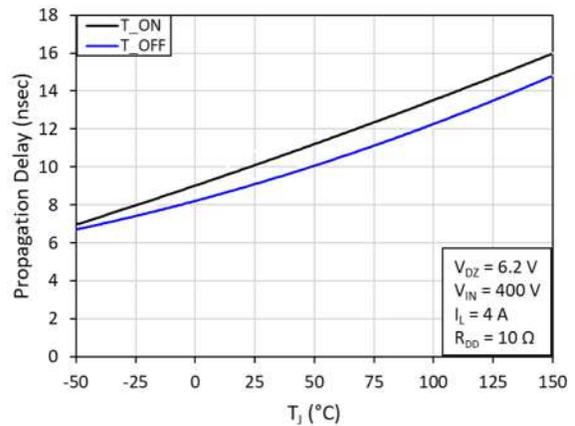


Fig. 14. Propagation delay ( $T_{ON}$  and  $T_{OFF}$ ) vs. junction temperature ( $T_J$ )

Characteristic Graphs (Cont.)

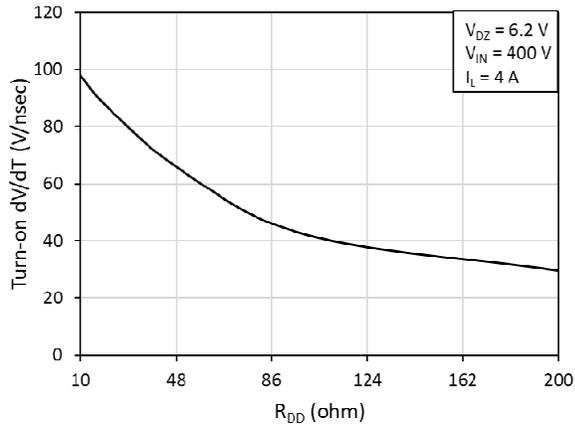


Fig. 15. Slew rate (dV/dt) vs. gate drive turn-on current set resistance ( $R_{DD}$ ) at  $T = 25\text{ }^{\circ}\text{C}$

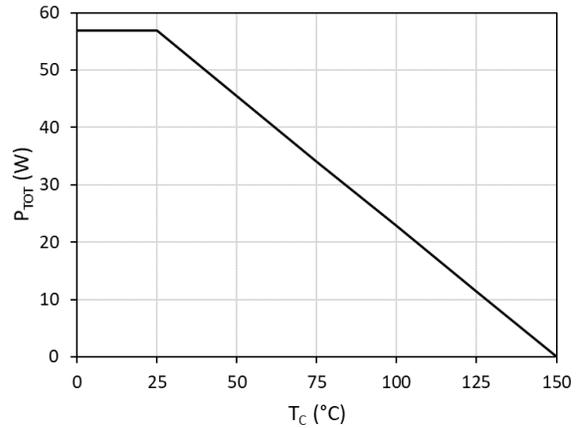


Fig. 16. Power dissipation ( $P_{TOT}$ ) vs. case temperature ( $T_C$ )

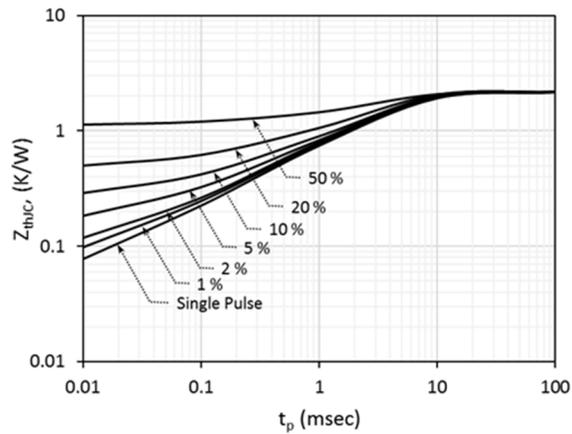
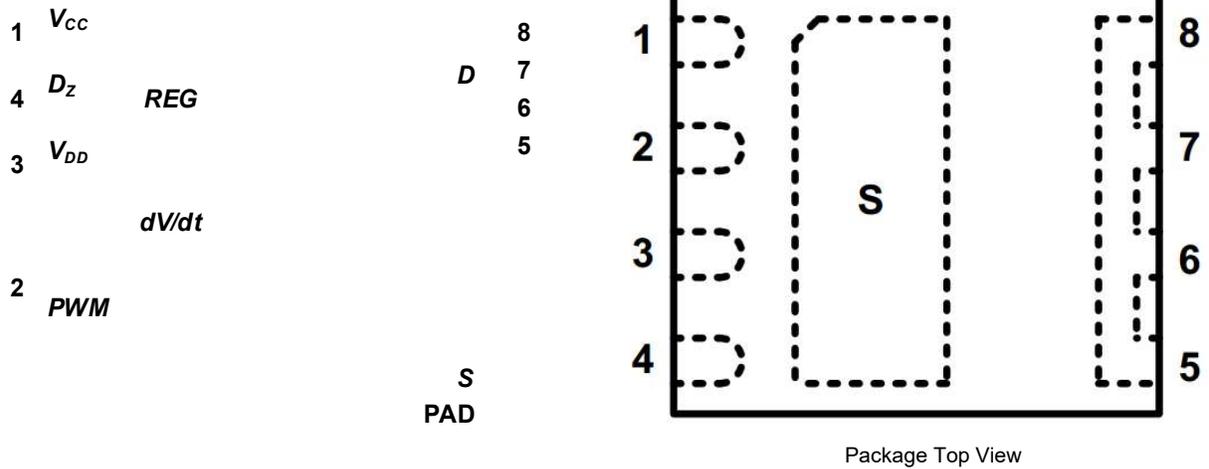


Fig. 17 Max. thermal transient impedance ( $Z_{thJC}$ ) vs. pulse width ( $t_p$ )

## 7. Internal Schematic, Pin Configurations and Functions



Pin		I/O <sup>(1)</sup>	Description
Number	Symbol		
1	$V_{CC}$	P	Supply voltage (10V to 24V)
2	PWM	I	PWM input
3	$V_{DD}$	I	Gate drive supply voltage. Gate drive turn-on current set pin (using $R_{DD}$ ).
4	$D_Z$	I	Gate drive supply voltage set pin (6.2 V Zener to GND).
5,6,7,8	$D$	P	Drain of power FET
PAD	$S$	O, G	Source of power FET & GaN IC supply ground. Metal pad on bottom of package.

(1) I = Input, O = Output, P = Power, G = GaN IC Ground

## 8. Functional Description

The following functional description contains additional information regarding the IC operating modes and pin functionality.

### 8.1. Start Up

When the  $V_{CC}$  supply is first applied to the NV6115, care should be taken such that the  $V_{DD}$  and  $D_Z$  pins are up at their correct voltage levels before the PWM input signal starts. The  $V_{DD}$  pin ramp up time is determined by the internal regulator current at this pin and the external  $C_{VDD}$  capacitor.  $C_{VDD}$  time constant should be calculated such that there is sufficient time to charge up the  $C_{VDD}$  capacitor to  $\sim 6V$ . In some scenarios, where fast startup is required, an optional diode in parallel with the  $R_{DD}$  can be used to ensure the  $C_{VDD}$  capacitor is fully charged before the first PWM pulse is applied. Also, since the  $D_Z$  pin voltage sets the  $V_{DD}$  voltage level, the  $V_{DD}$  pin will ramp up together with the  $D_Z$  pin (Fig. 18).

For half-bridge configurations, it is important that the  $V_{CC}$  supply, the  $D_Z$  pin, and the  $V_{DD}$  supply of the high-side NV6115 are all charged up to their proper levels before the first high-side PWM pulses start. For LLC applications, a long on-time PWM pulse to the low-side ( $> 10 \mu s$ ) is typically provided by the LLC controller to allow the supply pins of the high-side NV6115 to charge up (through the external bootstrap diode) to their correct levels before the first high-side PWM pulses start (Fig. 19).

For active clamp flyback (ACF) applications, the half-bridge must be ready very quickly due to the soft-start mode of the ACF controller. When the first few PWM pulses are generated by the ACF controller, the high-side supply pins of the NV6115 will require a few low-side pulses to charge up (through the external bootstrap diode) before the high-side starts to switch (Fig. 20).

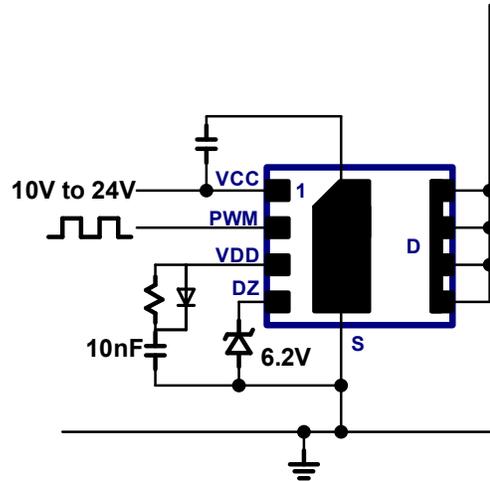


Fig. 18. Quick start-up circuit

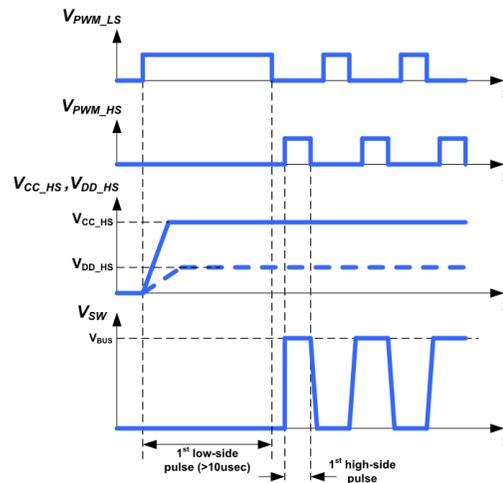


Fig. 19. LLC half-bridge start-up timing diagram

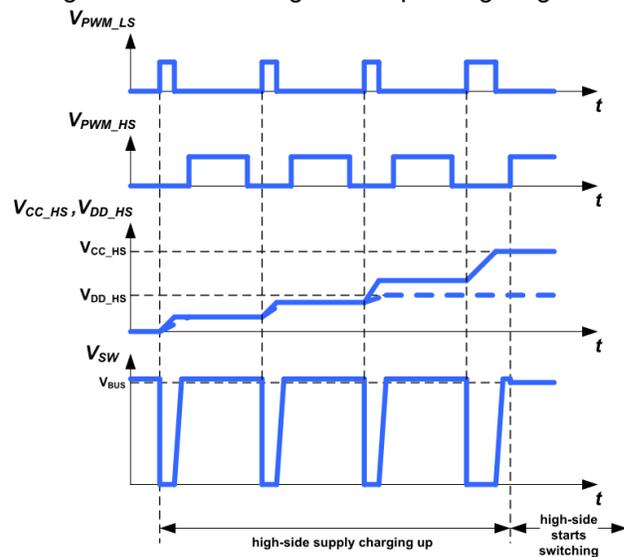


Fig. 20. ACF half-bridge start-up timing diagram

### 8.2. Normal Operating Mode

During Normal Operating Mode, all of the internal circuit blocks are active.  $V_{CC}$  is operating within the recommended range of 10 V to 24 V, the  $V_{DD}$  pin is at the voltage set by the Zener diode at the  $D_Z$  pin (6.2 V), and the internal gate drive and power FET are both enabled. The external PWM signal at the PWM pin determines the frequency and duty-cycle of the internal gate of the power FET. As the PWM voltage toggles above and below the rising and falling input thresholds (4 V and 1 V), the internal gate of the power FET toggles on and off between  $V_{DD}$  and 0 V (Fig. 21). The drain of the power FET then toggles between the source voltage (typically power ground) and a higher voltage level (650 V max), depending on the external power conversion circuit topology.

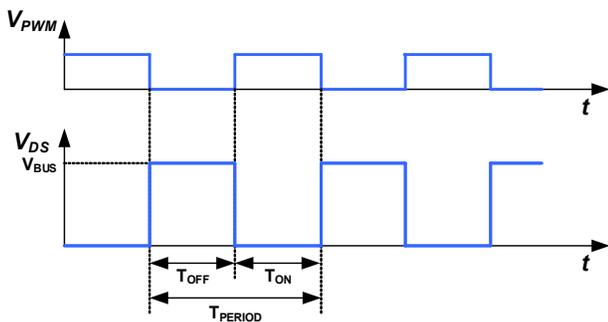


Fig. 21. Normal operating mode timing diagram

### 8.3. Standby Mode

For applications where a low standby power is required, an external series cut-off circuit (Fig. 22) can be used to disconnect  $V_{CC}$  of the NV6115 from the main  $V_{CC}$  supply of the power supply. This will reduce  $V_{CC}$  current consumption when the converter is in burst mode during light-load or open load conditions. The  $V_{CC}$  cut-off circuit consists of a series PMOS FET that is turned on and off with a pull-down NMOS FET. The gate of the NMOS is controlled by an external ENABLE signal that should be provided by the main controller of the power supply. The capacitor value at the  $V_{CC}$  pin should then be selected according to the desired start-up speed of the NV6115 each time the ENABLE signal toggles high. A 22 nF capacitor at  $V_{CC}$ , for example, will give a typical start-up time of approximately 2  $\mu$ s. An additional 200 k $\Omega$  resistor is placed across the PMOS cut-off FET to provide a small  $V_{CC}$  voltage level for proper start-up.

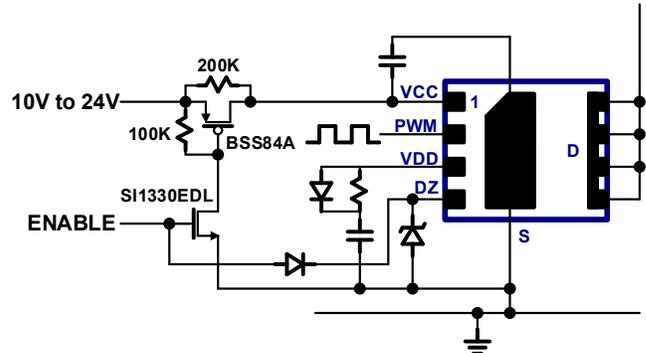


Fig. 22. Standby mode  $V_{CC}$  cut-off circuit

### 8.4. Programmable Turn-on dV/dt Control

During first start-up pulses or during hard-switching conditions, it is desirable to limit the slew rate ( $dV/dt$ ) of the drain of the power FET during turn-on. This is necessary to reduce EMI or reduce circuit switching noise. To program the turn-on  $dV/dt$  rate of the internal power FET, a resistor ( $R_{DD}$ ) is placed in between the  $V_{DD}$  capacitor and the  $V_{DD}$  pin. This resistor ( $R_{DD}$ ) sets the turn-on current of the internal gate driver and therefore sets the turn-on falling edge  $dV/dt$  rate of the drain of the power FET (Fig. 23). A typical turn-on slew-rate change with respect to  $R_{DD}$  is shown in Fig. 15.

**Minimum 10  $\Omega$   $R_{DD}$  is required.**

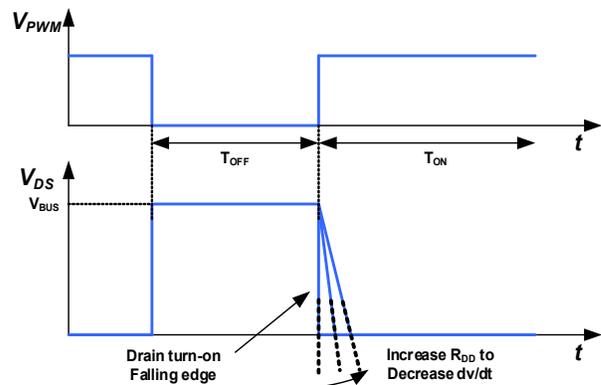


Fig. 23. Turn-on  $dV/dt$  slew rate control

### 8.5. Current Sensing

For many applications it is necessary to sense the cycle-by-cycle current flowing through the power FET. To sense the current flowing through the NV6115, a standard current-sensing resistor can be placed between the source and power ground (Fig. 24). In this configuration, all of the components around the NV6115 ( $C_{VCC}$ ,  $C_{VDD}$ ,  $D_Z$ , etc.) should be grounded with a single connection at the source. Also, an additional RC filter can be inserted between the PWM signal and the PWM pin (100  $\Omega$ , 100 pF typical). This filter is necessary to prevent false triggering due to high-frequency voltage spikes occurring at the source node due to external parasitic inductance from the source PCB trace or the current-sensing resistor itself.

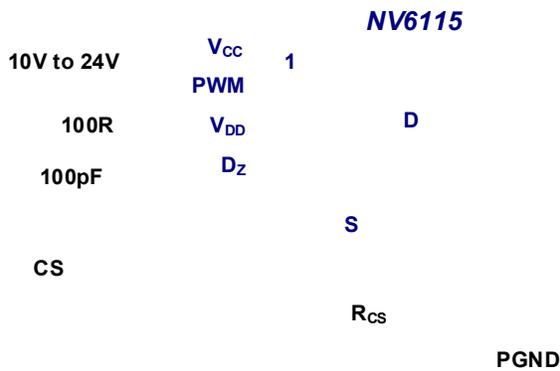


Fig. 24. Current sensing circuit

### 8.6. Paralleling Devices

For some applications it is desirable to parallel ICs in order to reduce conduction losses and temperatures. Two NV6115 ICs can be connected in parallel in a PFC boost application working in boundary-conduction mode (BCM) only. The parallel configuration for two NV6115 ICs is shown in Fig. 25. The paired pins that are connected together include the drain pins (D), the source pins (S), the  $V_{CC}$  pins, the PWM pins, and the  $D_Z$ . A single  $D_Z$  diode can be shared by both ICs. The  $V_{DD}$  pins are not connected together and require separate  $V_{DD}$  supply capacitors ( $C_{VDD1}$ ,  $C_{VDD2}$ ) and separate turn-on current set resistors ( $R_{DD1}$ ,  $R_{DD2}$ ). Each IC should have its own local  $V_{CC}$  supply filter capacitor ( $C_{VCC1}$ ,  $C_{VCC2}$ ). The PWM pins can have a single filter resistor ( $R_{PWM}$ ) but separate filter capacitors ( $C_{PWM1}$ ,  $C_{PWM2}$ ) should be placed at the PWM pin of each IC. When designing the PCB layout for the two paralleled ICs, the drain and source connections should be made as symmetrical as possible to avoid any parasitic inductance or capacitance mismatch. A proper PCB layout example for paralleling is shown in Section 10.

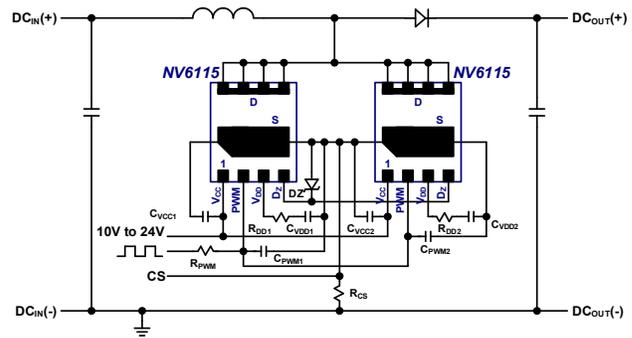


Fig. 25. Boost schematic using two parallel ICs

### 8.7. 3.3V PWM Input Circuit

For some applications where a 3.3 V PWM signal is required (DSP, MCU, etc.) an additional buffer can be placed before the PWM input pin (Fig. 26) with the buffer supply voltage connected to the V<sub>DD</sub> capacitor.

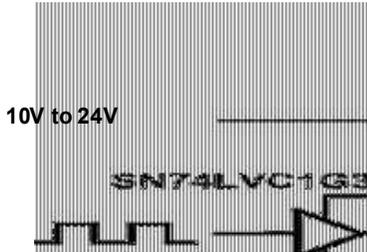


Fig. 26. 3.3 V PWM input buffer circuit

### 8.8. PCB Layout Guidelines

The design of the PCB layout is critical for good noise immunity, sufficient thermal management, and proper operation of the IC. Typical PCB layout examples for without current sensing resistor, with current sensing resistor, and paralleling, are all shown in Section 10.

The following rules should be followed carefully during the design of the PCB layout:

- 1) Place all IC filter and programming components directly next to the IC. These components include ( $C_{VCC}$ ,  $C_{VDD}$ ,  $R_{PWM}$ ,  $C_{PWM}$ ,  $R_{DD}$  and  $D_Z$ ).
- 2) Keep ground trace of IC filter and programming components separate from power GND trace. Do not run power GND currents through ground trace of filter components!
- 3) For best thermal management, place thermal vias in the source pad area to conduct the heat out through the bottom of the package and through the PCB board to other layers (see Section 10 for correct layout examples).
- 4) Use large PCB thermal planes (connected with thermal vias to the source pad) and additional PCB layers to reduce IC temperatures as much as possible (see Section 10 for correct layout examples).
- 5) For half-bridge layouts, do not extend copper planes from one IC across the components or pads of the other IC!
- 6) For high density designs, use a 4-layer PCB and 2 oz. copper to route signal connections. This allows layout to maintain large thermal copper planes and reduce power device temperature.

## 8.9. Recommended Component Values

The following table (Table I) shows the recommended component values for the external filter capacitors, Zener diode, and  $R_{DD}$  connected to the pins of this GaNFast power IC. These components should be placed as close as possible to the IC. Please see PCB Layout guidelines for more information. The Zener diode at the  $D_z$  pin should be a low-current type with a flat Zener, and the min/max limits must be followed.  $R_{DD}$  must be a minimum of 10  $\Omega$  to ensure application and device robustness.

SYM	DESCRIPTION	MIN	TYP	MAX	UNITS
$C_{VCC}$	Maximum $V_{CC}$ supply capacitor		0.1		$\mu\text{F}$
$C_{VDD}$	$V_{DD}$ supply capacitor		0.01		$\mu\text{F}$
$R_{DD}$	Gate drive turn-on current set resistor	10	25		$\Omega$
$R_{PWM}$	PWM filter resistor		100		$\Omega$
$C_{PWM}$	PWM filter capacitor		100		pF

Table I. Recommended component values.

### 8.9.1. Zener Selection

The Zener voltage is a critical parameter that sets the internal reference for gate drive voltage and other circuitry. The Zener diode needs to be selected such that the voltage on the  $D_z$  pin is within recommended operating conditions (5.8 V to 6.6 V) across operating temperature ( $-40^\circ\text{C}$  to  $125^\circ\text{C}$ ) and bias current (10  $\mu\text{A}$  to 1 mA). To ensure effective operation, the current vs. voltage characteristics of the Zener diode should be measured down to 10  $\mu\text{A}$  to ensure flat characteristics across the current operating range (10  $\mu\text{A}$  to 1 mA). The recommended part numbers meet these requirements (See ). If the Zener selected by user does not ensure that the voltage on the Zener pin is always within the recommended operating range, the functionality and reliability of the GaNFast power IC can be impacted.

Only the following Zener diodes are to be used ( ).

SYM	DESCRIPTION	PART NO.	SUPPLIER	MIN	TYP	MAX	UNITS
$D_z$	$V_{DD}$ set Zener diode ( $D_z$ pin)	BZT52B6V2 RHG	Taiwan Semiconductor Corporation	5.8	6.2	6.6	V
		MM3Z6V2ST1G	ON-Semiconductor				
		PDZ6.2B.115	Nexperia (NXP)				
		PLVA662A.215	Nexperia (NXP)				
		LM3Z6V2T1	Leshan Radio Company				

Table II. Qualified Zener diode components

An external resistor (~47 kΩ) between  $V_{CC}$  and  $D_Z$  can improve Zener voltage stability by adding bias current to the Zener pin to ensure the voltage on the  $D_Z$  pin is always within the recommended operating range (Fig. 27). This will add ~200 μA of quiescent current.

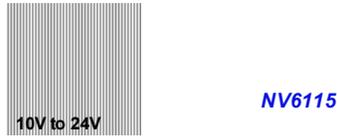


Fig. 27. Increasing Zener bias current for stable Zener voltage

### 8.10. GaN Power IC Drain-to-Source Voltage Design Considerations

GaN Power ICs have been designed and tested to provide significant design margin to handle transient and continuous voltage conditions that are commonly seen in single-ended topologies, such as quasi-resonant (QR) flyback applications. The different voltage levels and recommended margins in a typical QR flyback can be analyzed using . When the device is switched off, the energy stored in the transformer leakage inductance will cause  $V_{DS}$  to overshoot to the level of  $V_{SPIKE}$ . The clamp circuit should be designed to control the magnitude of  $V_{SPIKE}$ . It is recommended to apply an 80% derating from  $V_{DS(TRAN)}$  rating (800V) to 650 V max for repetitive  $V_{DS}$  spikes under the worst case steady-state operating conditions. After dissipation of the leakage energy, the device  $V_{DS}$  will settle to the level of the bus voltage plus the reflected output voltage which is defined in Fig. 28 as  $V_{PLATEAU}$ . It is recommended to design the system such that  $V_{PLATEAU}$  follows a typical derating of 80% (520V) from  $V_{DS(CONT)}$  (650V). Finally,  $V_{DS(TRAN)}$  (800V) rating is also provided for events that occur on a non-repetitive basis, such as line surge, lightning strikes, start-up, over-current, short-circuit, load transient, and output voltage transition. 800V  $V_{DS(TRAN)}$  ensures excellent device robustness and no-derating is needed for these non-repetitive events, assuming the surge duration is < 100 μs. For half-bridge based topologies, such as LLC,  $V_{DS}$  voltage is clamped to the bus voltage.  $V_{DS}$  should be designed such that it meets the  $V_{PLATEAU}$  derating guideline (520V).

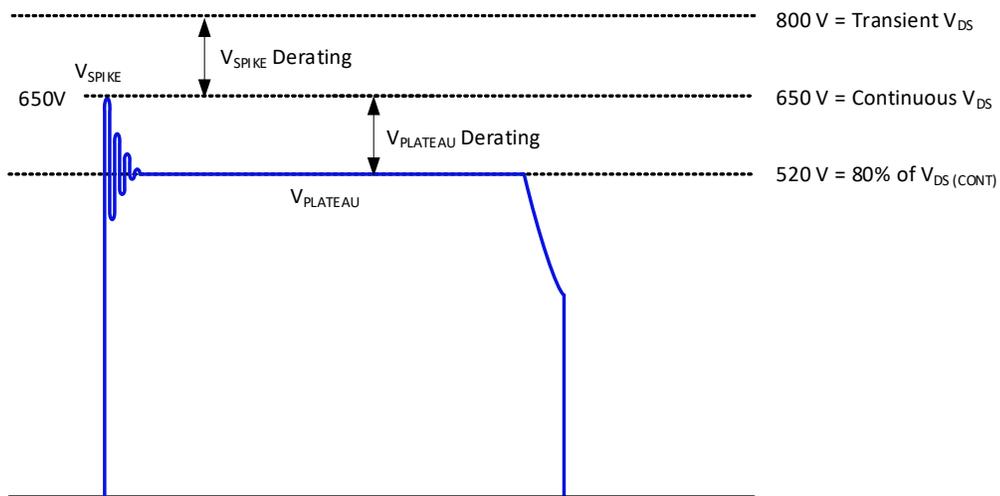
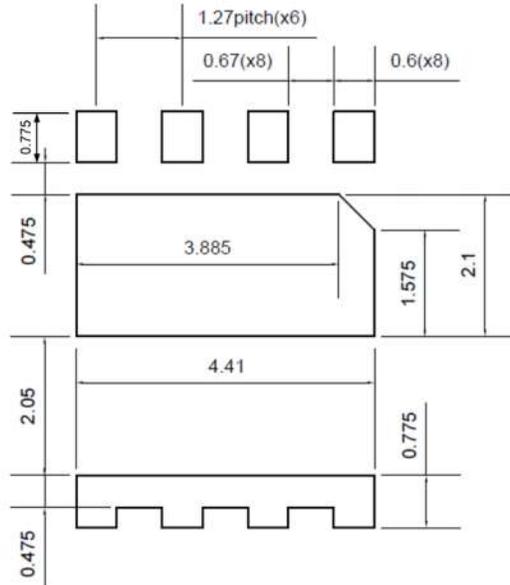


Fig. 28. QR flyback drain-to-source voltage stress diagram

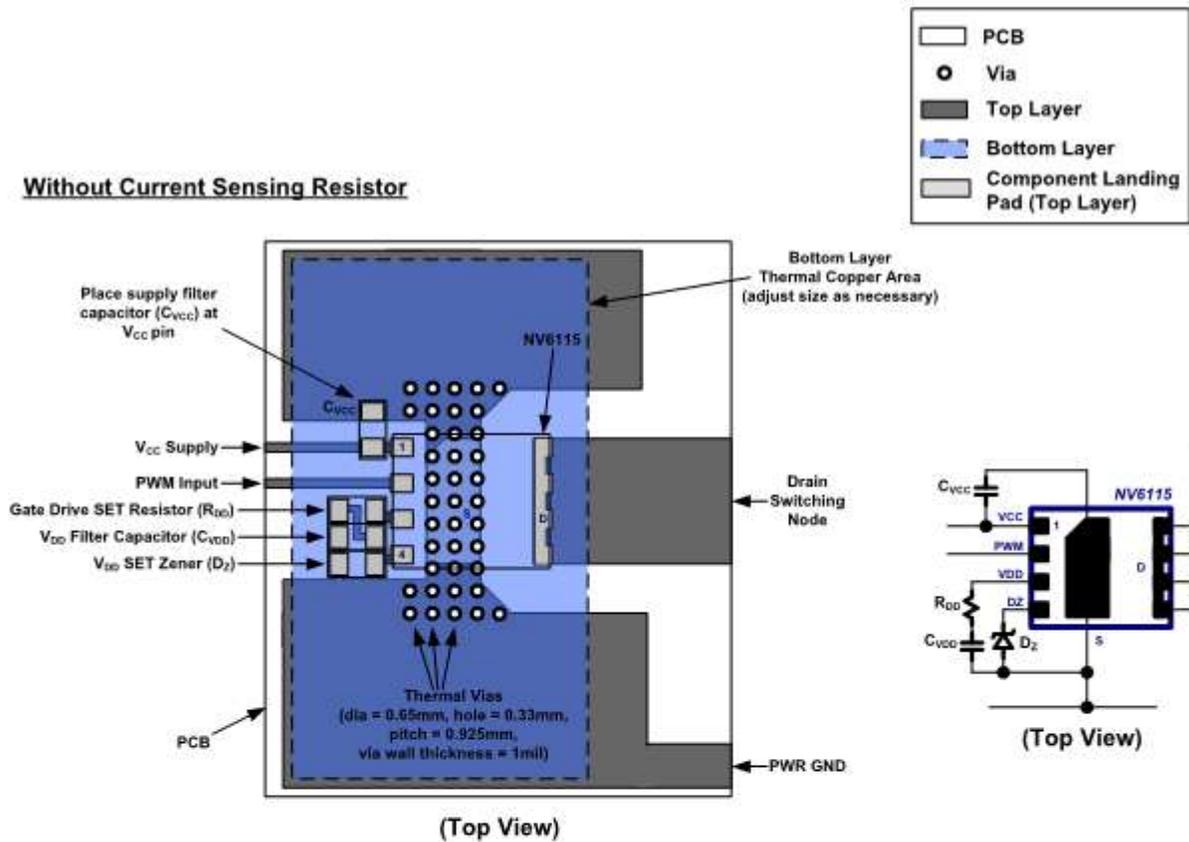
9. Recommended PCB Land Pattern



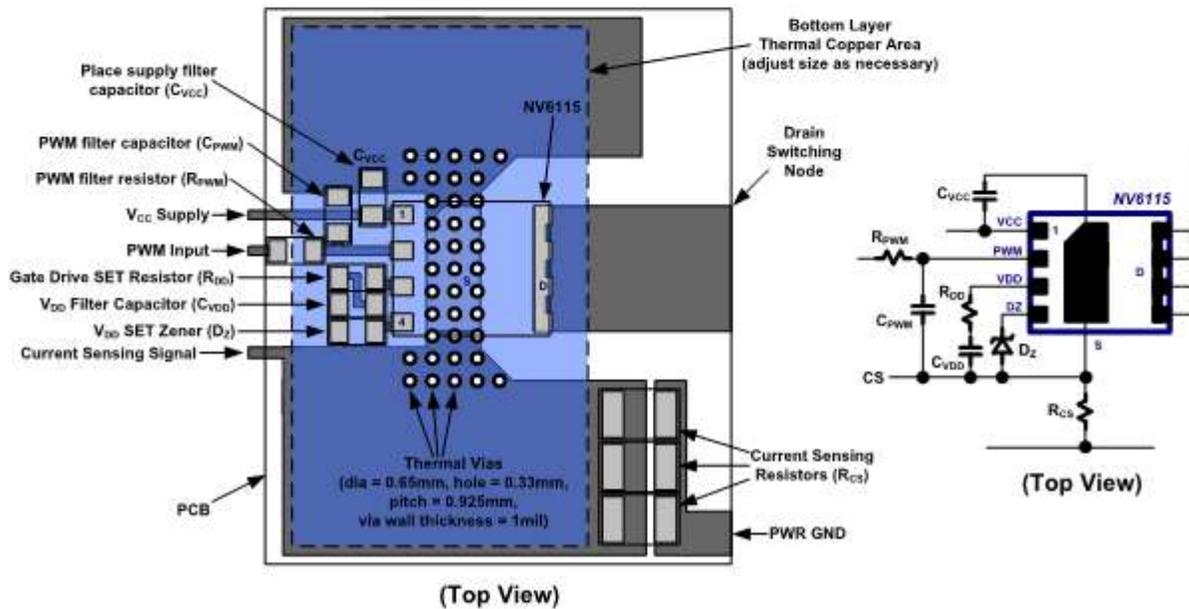
All dimensions are in mm

10. PCB Layout Guidelines

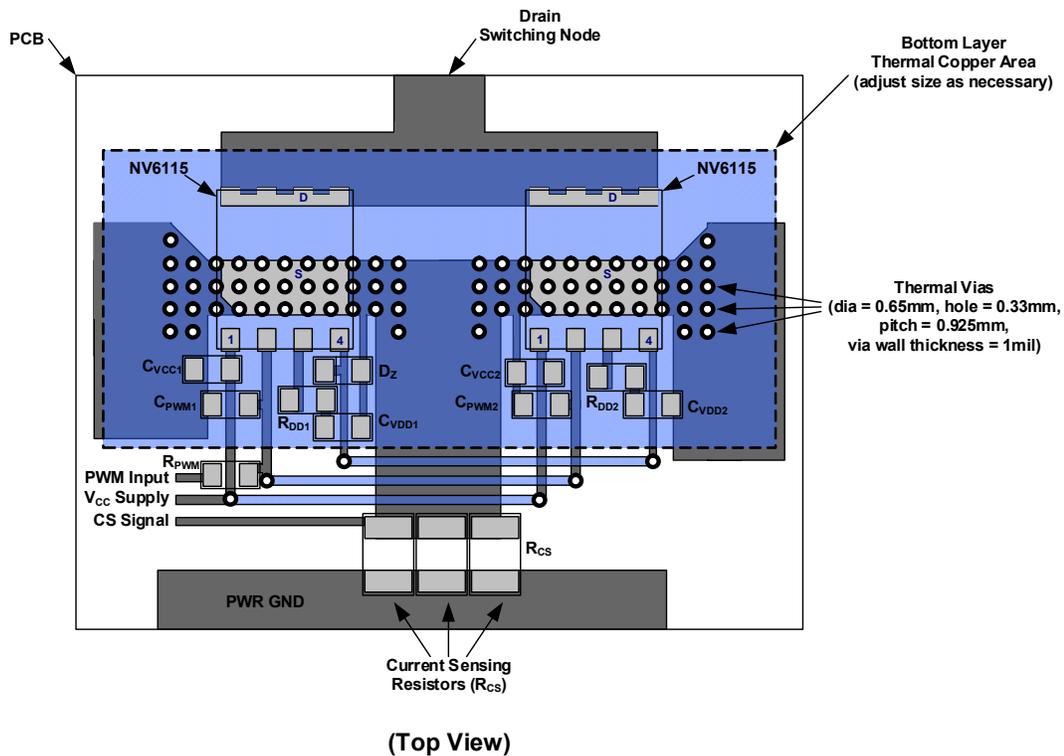
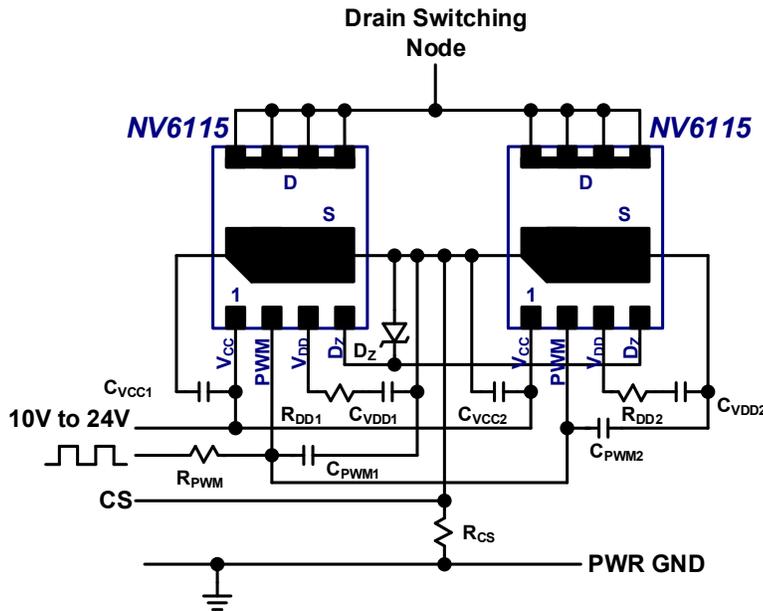
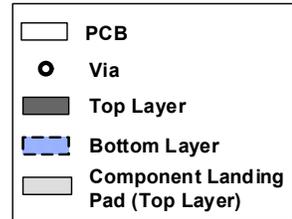
Without Current Sensing Resistor



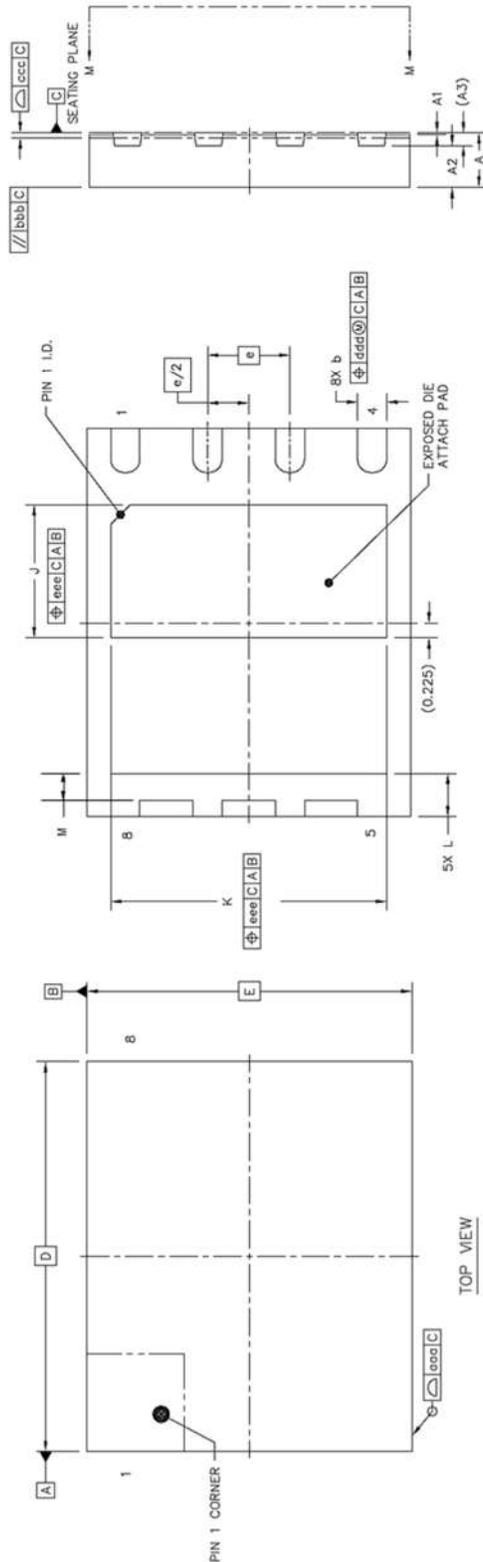
With Current Sensing Resistor



Paralleling 2 ICs (Boost PFC, BCM Mode only)



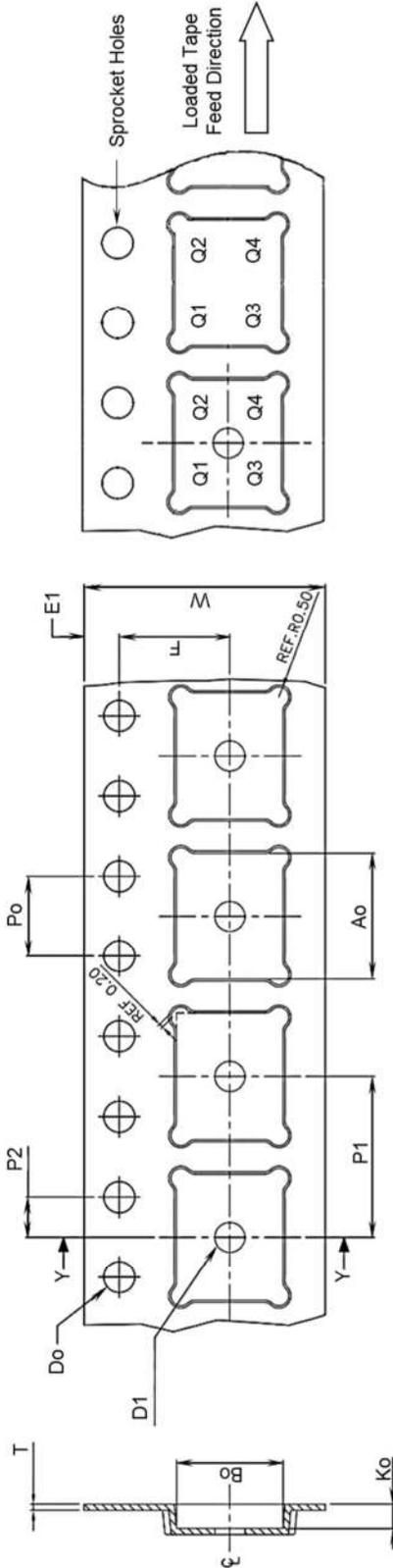
11. QFN Package Outline



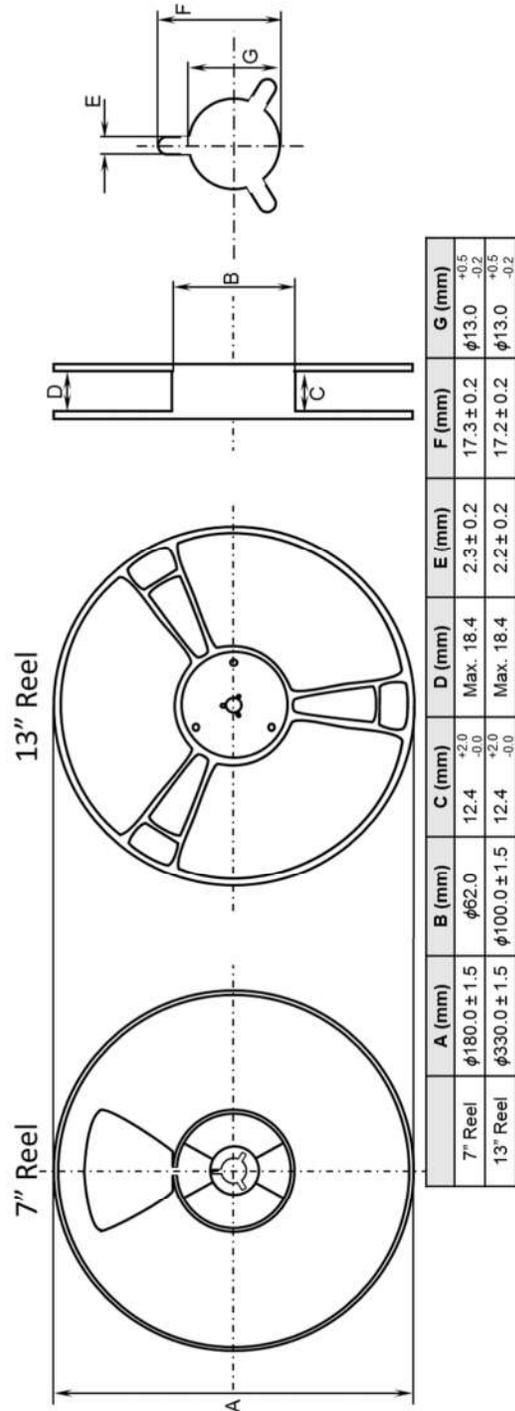
BOTTOM VIEW  
VIEW M-M

	SYMBOL	MIN	NOM	MAX	SYMBOL	MIN	NOM	MAX
TOTAL THICKNESS	A	0.8	0.85	0.9	X	J	1.95	2.05
STAND OFF	A1	0	0.035	0.05	Y	K	4.16	4.26
MOLD THICKNESS	A3	---	0.65	---	L	L	0.625	0.675
L/F THICKNESS	A3	---	0.203 REF	---	M	M	0.43	0.725
LEAD WIDTH	B	0.4	0.45	0.5	aaa	aaa	0.1	0.1
BODY SIZE	X	6 BSC			bbb	bbb	0.1	0.1
	Y	5 BSC			ccc	ccc	0.08	0.08
LEAD PITCH	e	1.27 BSC			ddd	ddd	0.1	0.1
					eee	eee	0.1	0.1

12. Tape and Reel Dimensions



Ao (mm)	Bo (mm)	Do (mm)	D1 (mm)	E1 (mm)	F (mm)	Ko (mm)	Po (mm)	P1 (mm)	P2 (mm)	T (mm)	W (mm)	Pin1 Quadrant
6.30 ± 0.1	5.30 ± 0.1	φ1.55 ± 0.05	min. φ1.50	1.75 ± 0.1	5.50 ± 0.1	1.20 ± 0.1	4.0 ± 0.1	8.00 ± 0.1	2.0 ± 0.1	0.30 ± 0.05	12.00 ± 0.1	Q1



	A (mm)	B (mm)	C (mm)	D (mm)	E (mm)	F (mm)	G (mm)
7" Reel	φ180.0 ± 1.5	φ62.0	12.4 <sup>+2.0</sup> <sub>-0.0</sub>	Max. 18.4	2.3 ± 0.2	17.3 ± 0.2	φ13.0 <sup>+0.5</sup> <sub>-0.2</sub>
13" Reel	φ330.0 ± 1.5	φ100.0 ± 1.5	12.4 <sup>+2.0</sup> <sub>-0.0</sub>	Max. 18.4	2.2 ± 0.2	17.2 ± 0.2	φ13.0 <sup>+0.5</sup> <sub>-0.2</sub>

### 13. Ordering Information

Part Number	Operating Temperature Grade	Storage Temperature Range	Package	MSL Rating	Packing (Tape & Reel)
NV6115 RA	-40 °C to +125 °C T <sub>CASE</sub>	-55 °C to +150 °C T <sub>CASE</sub>	5 x 6 mm QFN	1	1,000 : 7" Reel
NV6115	-40 °C to +125 °C TCASE	-55 °C to +150 °C TCASE	5 x 6 mm QFN	1	5,000 : 13" Reel

### 14. 20-Year Limited Product Warranty

The 20-year limited warranty applies to all packaged Navitas GaNFast Power ICs in mass production, subject to the terms and conditions of, Navitas' express limited product warranty, available at <https://navitassemi.com/terms-conditions>. The warranted specifications include only the MIN and MAX values only listed in Absolute Maximum Ratings, ESD Ratings and Electrical Characteristics sections of this datasheet. Typical (TYP) values or other specifications are not warranted.



### 15. Revision History

Date	Status	Notes
Aug 13, 2018	Initial Release	First publication
Nov 22, 2019	Revised	Updated MSL rating, added Section 8.10, updated Section 8.3 and figure 22.
Jun 22, 2020	Revised	Updated guidelines in Section 8.1: Fast startup. Updated R <sub>DD</sub> specification. This datasheet revision applies to product prior to PCN2020-03-30
Sept 8, 2020	Revised	Datasheet updated in alignment with PCN2020-03-30. Product is marked with Lot Code ending in Z for traceability
Apr 11, 2022	Revised	Added 20-Year Limited Product Warranty

### Additional Information

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