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AS3676

Flexible Lighting Management (CP, DCDC, 13 Current Sinks, ADC, LED Test, LDO, DLS and Ambient Light sensing)

1 General Description

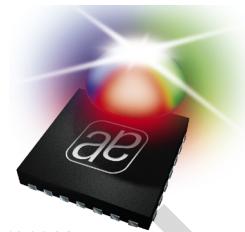
The AS3676 is a highly-integrated CMOS Power and Lighting Management Unit for mobile telephones, and other 1-cell Li+ or 3-cell NiMH powered devices.

The AS3676 incorporates one Step Up DC/DC Converter for white backlight LEDs, one high-power Charge Pump, one Analog-to-Digital Converter, 13 current sinks, LED in-circuit function test, a two wire serial interface, and control logic all onto a single device. It supports **ambient light sensor** processing and a Dynamic Luminance Scaling (**DLS**) input. Output voltages and output currents are fully programmable.

The AS3676 is part of the austriamicrosystems AS3675, AS3687/87XM and AS3689 lighting management unit family. It is software compatible to AS3687/87XM and AS3689 and pin and software compatible to AS3675.

2 Key Features

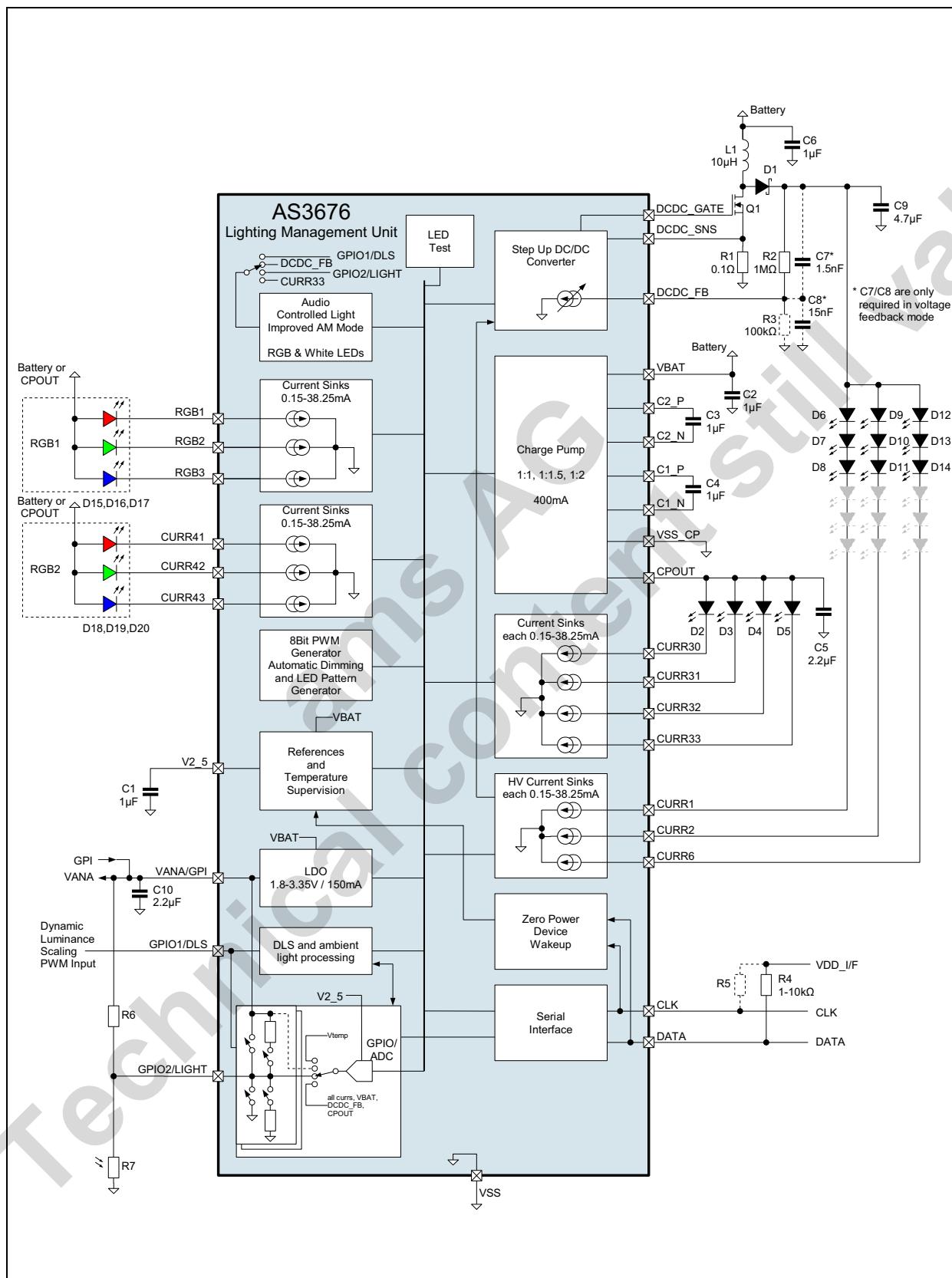
- High-Efficiency Step Up DC/DC Converter
 - Up to 26V/50mA for White LEDs
 - Programmable Output Voltage with External Resistors and Serial Interface
 - Over voltage Protection
- High-Efficiency High-Power Charge Pump
 - 1:1, 1:1.5, and 1:2 Mode
 - Automatic Up Switching (can be disabled and 1:2 mode can be blocked)
 - Output Current up to 400mA/500mA pulsed
 - Efficiency up to 95%
 - Very Low effective Resistance (2.5Ω typ. in 1:1.5)
 - Only 4 External Capacitors Required:
 - 2 x 1µF Flying Capacitors, 2 x 2.2µF Input/Output Capacitors
 - Supports LCD White Backlight LEDs, or RGB LEDs
- 13 Current Sinks
 - All 13 current sinks fully Programmable (8-bit) from: 0.15mA to 38.5mA (up to 75.6mA for CURR30...CURR33)
 - Three current sinks are High Voltage capable (CURR1, CURR2, CURR6)
 - Programmable Hardware Control (Strobe, and Preview or PWM)
 - Selectively Enable/Disable Current Sinks
 - Dynamic Luminance Scaling (DLS) support to improve backlight operating time (can adjust any current source)
- Light Sensor input with internal hardware processing to control backlight according to ambient light using 3 groups - any current source can select one of the groups or no light sensor control
- Internal PWM Generation
 - 8 Bit resolution
 - Autonomous Logarithmic up/down dimming
- Led Pattern Generator
 - Autonomous driving for Fun RGB LEDs
 - Support indicator LEDs
- 10-bit Successive Approximation ADC
 - 27µs Conversion Time
 - Selectable Inputs: VANA/GPI, GPIO1/DLS, GPIO2/LIGHT, all current sources, VBAT, CPOUT, DCDC_FB
 - Internal Temp. Measurement
 - Light Sensor input with Java support (JSR-256): read ADC processed value
- Support for automatic LED testing (open and shorted LEDs can be identified)
- Strobe Timeout protection
 - Up to 1600ms
 - Three different timing modes
- Two General Purpose Inputs/Output
 - GPIO1/DLS, GPIO2/LIGHT
 - Digital Input, Digital Output using VANA/GPI supply and Tristate
 - GPIOs Programmable Pull-Up/Down
- Programmable LDO
 - 1.8 to 3.35V, 150mA
 - Programmable via Serial Interface
- Standby LDO always on
 - Regulated 2.5V max. output 10mA
 - 3µA Quiescent Current
- Audio can be used to drive RGB LED(s)
 - Color and Brightness depends on audio amplitude
- Wide Battery Supply Range: 3.0 to 5.5V
- Two Wire Serial Interface Control
- Over current and Thermal Protection
- WL-CSP30 3x2.5mm, 0.5mm pitch Package



3 Applications

Power- and lighting-management for mobile telephones and other 1-cell Li+ or 3-cell NiMH powered devices.

Figure 1. AS3676 Block Diagram



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4 Pinout

Table 1. Pin Description for AS3676

Pin Number	Pin Name	Type	Description
A1	GPIO1/DLS	AIO	Digital Luminance Scaling PWM input and General Purpose Input Output 1
A2	VANA/GPI	AIO	LDO Output/General Purpose Input
A3	C2_N	AIO	Charge Pump flying capacitor; connect a ceramic capacitor of 500nF to this pin.
A4	C1_P	AIO	Charge Pump flying capacitor; connect a ceramic capacitor of 500nF to this pin.
A5	CPOUT	AO	Output voltage of the Charge Pump; connect a ceramic capacitor of 1µF ($\pm 20\%$).
A6	DATA	DIO	Serial interface data input/output.
B1	GPIO2/LIGHT	AIO	Ambient Light Sensor input and General Purpose Input Output 2
B2	VSS_CP	GND	Ground Pad for Charge Pump
B3	C1_N	AIO	Charge Pump flying capacitor; connect a ceramic capacitor of 500nF to this pin.
B4	C2_P	AIO	Charge Pump flying capacitor; connect a ceramic capacitor of 500nF to this pin.
B5	DCDC_GATE	AO	DCDC gate driver.
B6	CLK	DI	Clock input for serial interface.
C1	CURR41	AI	Analog current sink input
C2	RGB3	AI	Analog current sink input
C3	VSS	GND	Ground pad
C4	VBAT	S	Supply pad. Connect to battery.
C5	CURR30	AI	Analog current sink input, intended for activity icon LED
C6	DCDC_SNS	AI	Sense input of shunt resistor for Step Up DC/DC Converter.
D1	CURR43	AI	Analog current sink input
D2	RGB1	AI	Analog current sink input
D3	CURR33	AI	Analog current sink input, intended for activity icon LED
D4	CURR31	AI	Analog current sink input, intended for activity icon LED
D5	CURR2	AI_HV	Analog current sink input (intended for Keyboard backlight)
D6	DCDC_FB	AI	DCDC feedback. Connect to resistor string.
E1	CURR42	AI	Analog current sink input
E2	RGB2	AI	Analog current sink input
E3	CURR32	AI	Analog current sink input, intended for activity icon LED
E4	CURR6	AI_HV	Analog current sink input (intended for Keyboard backlight)
E5	CURR1	AI_HV	Analog current sink input (intended for Keyboard backlight)
E6	V2_5	AO3	Output voltage of the Low-Power LDO; always connect a ceramic capacitor of 1µF ($\pm 20\%$) or 2.2µF ($+100\%/-50\%$).

4.1 Pin Definitions

Table 2. Pin Type Definitions

Type	Description
DI	Digital Input
DO	Digital Output
DIO	Digital Input/Output
AIO	Analog Pad
AI	Analog Input
AI_HV	High-Voltage (26V) Pin
AO3	Analog Output (3.3V)
S	Supply Pad
GND	Ground Pad

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5 Absolute Maximum Ratings

Stresses beyond those listed in [Table 3](#) may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in [Table 4, "Operating Conditions," on page 6](#) is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 3. Absolute Maximum Ratings

Symbol	Parameter	Min	Max	Units	Comments
VIN_HV	26V Pins	-0.3	26	V	Applicable for high-voltage current sink pins CURR1, CURR2, CURR6
VIN_MV	5V Pins	-0.3	7.0	V	Applicable for 5V pins VBAT, CURR30-33, CURR41-43, RGB1-3, C1_N, C2_N, C1_P, C2_P, CPOUT, DCDC_FB, DCDC_GATE, CLK, DATA;
VIN_LV	3.3V Pins	-0.3	5.0	V	Applicable for 3.3V pins V2_5; DCDC_SNS, GPIO1/DLS, GPIO2/LIGHT, VANA/GPI
	Input Pin Current	-25	+25	mA	At 25°C, Norm: JEDEC 17
Tstrg	Storage Temperature Range	-55	125	°C	
IIN	Humidity	5	85	%	Non-condensing
VESD	Electrostatic Discharge	-2000	2000	V	Norm: MIL 883 E Method 3015
Pt	Total Power Dissipation		0.75	W	TA = 70 °C, Tjunc_max = 125°C
TBODY	Peak Body Temperature		260	°C	T = 20 to 40s, in accordance with IPC/JEDEC J-STD 020.
MSL	Moisture Sensitivity Level	MSL 1			Represents a max. floor life time of unlimited

6 Electrical Characteristics

Table 4. Operating Conditions

Symbol	Parameter	Condition	Min	Typ	Max	Unit
General Operating Conditions						
VHV	High Voltage	Applicable for high-voltage current sink pins CURR1, CURR2 and CURR6.	0.0		26.0	V
VBAT	Battery Voltage	Pin VBAT	3.0	3.6	5.5	V
VPERI	Periphery Supply Voltage	For serial interface pins.	1.5		5.5	V
V2_5	Voltage on Pin V2_5	Internally generated	2.4	2.5	2.6	V
TAMB	Operating Temperature Range		-30	25	85	°C
IACTIVE	Battery current	Normal Operating current (see Operating Modes on page 80)		110		µA
ISTANDBY	Standby Mode Current	Current consumption in standby mode. Only 2.5V regulator on, interface active		8	13	µA
ISHUTDOWN	Shutdown Mode Current	interface inactive (CLK and DATA set to 0V)		0.1	3	µA

7 Typical Operating Characteristics

Figure 2. DCDC Step Up Converter: Efficiency of +15V, Step Up to 15V vs. Load Current at VBAT=3.8V

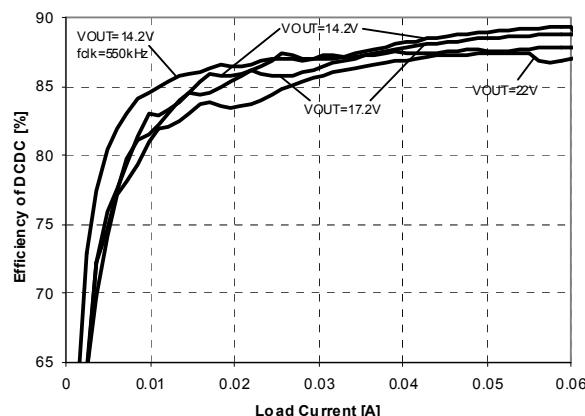


Figure 4. Charge Pump: Battery Current vs. VBAT

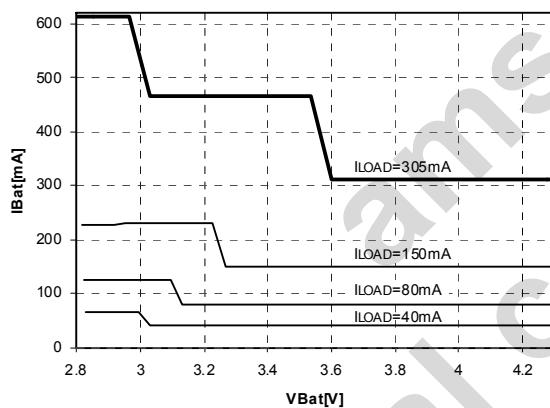


Figure 6.

Figure 3. Charge Pump: Efficiency vs. VBAT

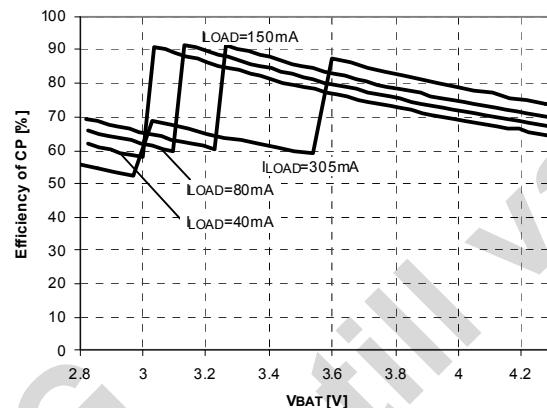


Figure 5. Current Sink CURR1 vs. V(CURRx)

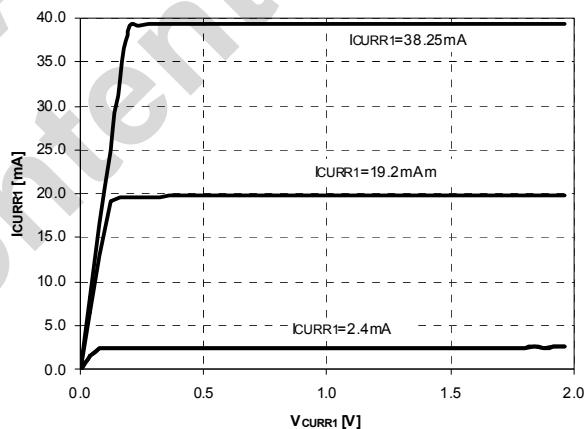


Figure 7. Current Sink CURR3x vs. VBAT

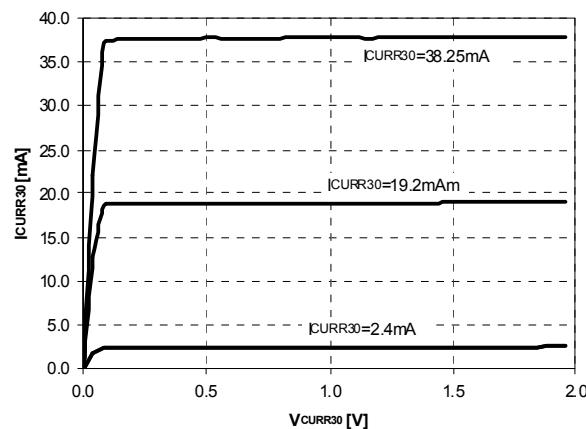


Figure 8. Charge Pump Input and Output Ripple
1:1.5 Mode, 100mA load

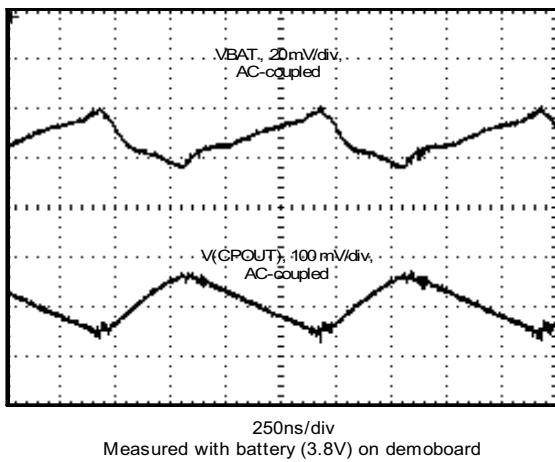
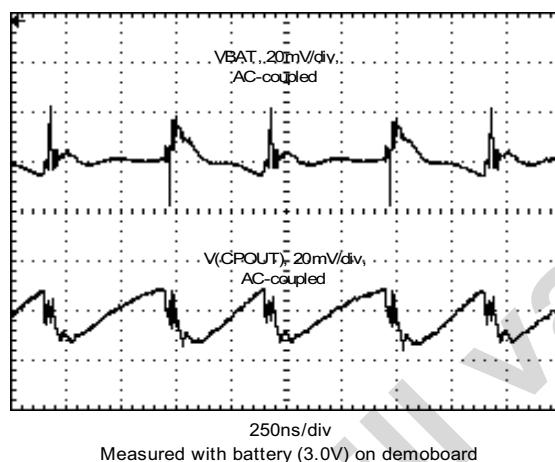


Figure 9. Charge Pump Input and Output Ripple
1:2 Mode, 100mA load



VBAT = 3.6V, TA = +25°C (unless otherwise specified).

8 Detailed Description

8.1 Analog LDO

The LDO is a general purpose LDO and the output pin connected to VANA/GPI. The design is optimized to deliver the best compromise between quiescent current and regulator performance for battery powered devices.

Stability is guaranteed with ceramic output capacitors of $1\mu\text{F} \pm 20\%$ (X5R) or $2.2\mu\text{F} +100/-50\%$ (Z5U). The low ESR of these capacitors ensures low output impedance at high frequencies. The low impedance of the power transistor enables the device to deliver up to 150mA even at nearly discharged batteries without any decrease in performance.

The LDO is off by default after start-up.

Figure 10. Analog LDO Block Diagram

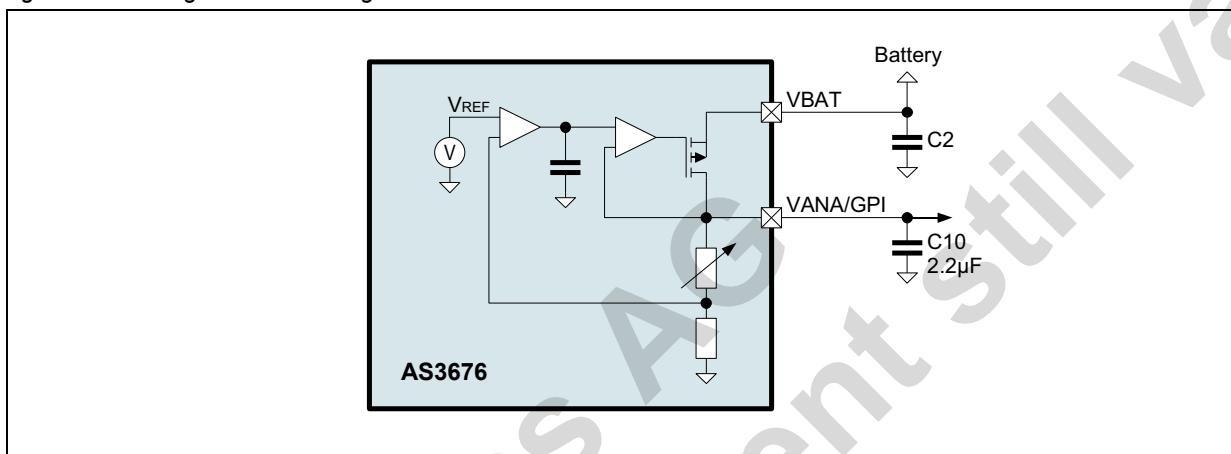


Table 5. Electrical Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Unit
V _{BAT}	Supply Voltage Range		3.0		5.5	V
R _{ON}	On Resistance	@150mA, full operating temperature range			1.0	Ω
V _{DROPOUT}	Dropout Voltage	@150mA			150	mV
		@50mA			50	mV
I _{ON}	Supply Current	Without load		50		µA
		With 150mA load		150		
I _{OFF}	Shutdown Current	Without load			100	nA
t _{start}	Start-up Time				200	µs
V _{out_tol}	Output Voltage Tolerance		-3		+3	%
V _{OUT}	Output Voltage	V _{BAT} = 3.0V and I _{OUT} =150mA	1.8		2.85	V
		Full Programmable Range; V _{BAT} > V _{OUT} + 150mV and I _{OUT} <=150mA	1.8		3.35	V
I _{LIMIT} ¹	LDO Current Limit	Pin VANA/GPI. LDO acts as current source if the output current exceeds I _{LIMIT} .	300	450 ²		mA

1. Not production tested – guaranteed by design and laboratory verification

2. During startup of the LDO the current limit is half the value of I_{LIMIT}

8.1.1 LDO Registers

Table 6. *Reg. control* Register

Addr: 00		Reg. control			
		This register enables/disables the LDOs, Charge Pumps, Charge Pump LEDs, current sinks, the Step Up DC/DC Converter, and low-power mode.			
Bit	Bit Name	Default	Access	Description	
0	ldo_on	0	R/W	0	Analog LDO is switched off
				1	Analog LDO is switched on

Table 7. *LDO Voltage* Register

Addr: 07h		LDO Voltage			
		This register sets the output voltage (VANA/GPI) for the LDO.			
Bit	Bit Name	Default	Access	Description	
4:0	ldo_voltage	00000b	R/W	Controls LDO voltage selection.	
				00000b	1.8V
				...	LSB=50mV
				11111b	3.35V

8.2 Step Up DC/DC Converter

The Step Up DC/DC Converter is a high-efficiency current mode PWM regulator, providing output voltage up to e.g. 25V/50mA¹. A constant switching-frequency results in a low noise on the supply and output voltages.

Figure 11. Step Up DCDC Converter Block Diagram Option: Current Feedback with Over voltage protection

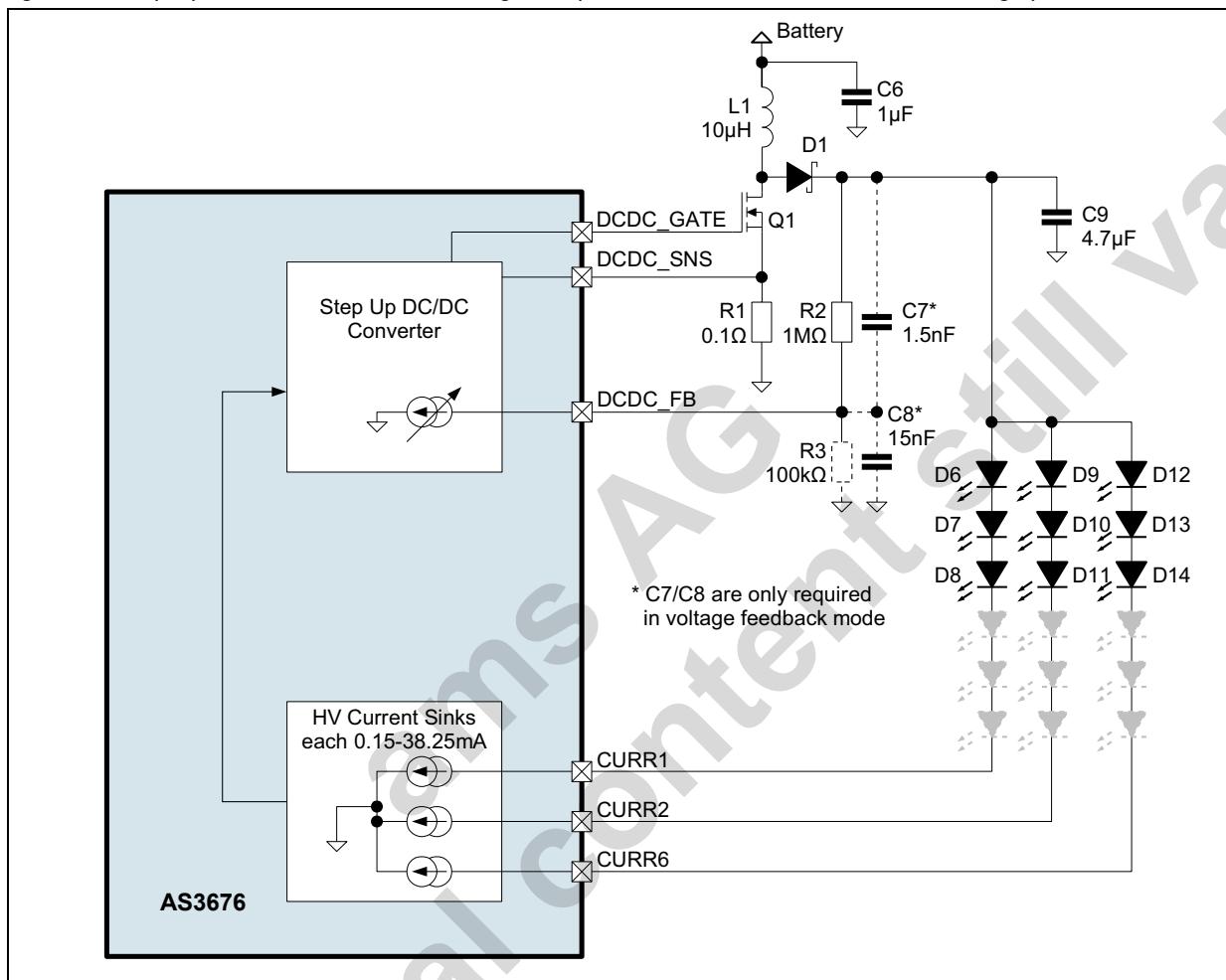


Table 8. Step Up DC/DC Converter Parameters

Symbol	Parameter	Condition	Min	Typ	Max	Unit
I _{VDD}	Quiescent Current	Pulse skipping mode.		140		µA
V _F B1	Feedback Voltage for External Resistor Divider	For constant voltage control. <i>step_up_res</i> = 1	1.20	1.25	1.30	V
V _F B2	Feedback Voltage for Current Sink Regulation	on CURR1, CURR2 or CURR6 in regulation. <i>step_up_res</i> = 0	0.4	0.5	0.6	V

1. The AS3676 internal driver structure allows output voltage higher than 25V. The [Over voltage Protection in Current Feedback Mode \(see page 13\)](#) or [Voltage Feedback \(see page 14\)](#) should be set to fit to the external components used (maximum voltage rating of Q1, C9 and D1).

Table 8. Step Up DC/DC Converter Parameters

Symbol	Parameter	Condition	Min	Typ	Max	Unit
I _{DCDC_FB}	Additional Tuning Current at Pin DCDC_FB and over voltage protection	Adjustable by software using Register DCDC control1 1µA step size (0-31µA) $V_{PROTECT} = 1.25V + I_{DCDC_FB} * R_2$	0		31	µA
	Accuracy of Feedback Current at full scale		-6		6	%
V _{sense_max}	Current Limit Voltage at R ₁	e.g., 1.32A for 0.1Ω sense resistor R ₁ .	92	132	170	mV
		For fixed startup time of 500us	50	66	86	
		If <code>step_up_lowcur=1</code>	60	86	114	
R _{sw}	Switch Resistance	ON-resistance of external switching transistor.			1	Ω
I _{LOAD}	Load Current	At 26V output voltage	0		50	mA
f _{IN}	Switching Frequency	Internally trimmed	0.9	1	1.1	MHz
C _{OUT}	Output Capacitor	Ceramic, ±20%. Use nominal 4.7µF capacitors to obtain at least 0.7µF under all conditions (voltage dependence of capacitors)	0.7	4.7		µF
L	Inductor	Use inductors with small C _{parasitic} (<100pF) to get high efficiency.	7	10	13	µH
t _{MIN_ON}	Minimum on Time		90	140	190	ns
MDC	Maximum Duty Cycle		90			%
Vripple	Voltage ripple >20kHz	Cout=4.7µF, Iout=0..45mA, VBAT=3.0...4.2V			160	mV
	Voltage ripple <20kHz				40	mV
Efficiency	Efficiency	Iout=20mA, Vout=17V, VBAT=3.8V		85		%

To ensure soft startup of the dc/dc converter, the over current limits are reduced for a fixed time after enabling the dc/dc converter. The total startup time for an output voltage of e.g. 26V is less than 2ms. If C7 and C8 are mounted and the bit `step_up_prot` is set, the total startup time can exceed 2ms.

Note: If the DCDC converter is only used in current feedback mode (CURR1, CURR2 or CURR6 - and not used for a constant voltage source), the capacitors C7 and C8 can be removed.

8.2.1 Feedback Selection

Register DCDC control1 and DCDC control2 selects the type of feedback for the Step Up DC/DC Converter.

The feedback for the DC/DC converter can be selected either by current sinks (CURR1, CURR2, CURR6) or by a voltage feedback at pin DCDC_FB. If the register bit `step_up_fb_auto` is set, the feedback path is automatically selected between CURR1, CURR2 and CURR6 (the lowest voltage of these current sinks is used).

Setting `step_up_fb` enables feedback on the pins CURR1, CURR2 or CURR6. The Step Up DC/DC Converter is regulated such that the required current at the feedback path can be supported. (Bit `step_up_res` should be set to 0 in this configuration)

Note: Always choose the path with the highest voltage drop as feedback to guarantee adequate supply for the other (unregulated) paths or enable the register bit `step_up_fb_auto`.

8.2.2 Over voltage Protection in Current Feedback Mode

The over voltage protection in current feedback mode (`step_up_fb` = 01, 10 or 11 or `step_up_fb_auto` = 1) works as follows: Only resistor R2 and C7/C8² is soldered and R3 is omitted. An internal current source (sink) is used to generate a voltage drop across the resistor R2. If then the voltage on DCDC_FB is above 1.25V and `step_up_prot`=1, the DCDC is momentarily disabled to avoid too high voltages on the output of the DCDC converter. When the voltage on DCDC_FB drops below 1.25V, the DCDC automatically resumes operation.

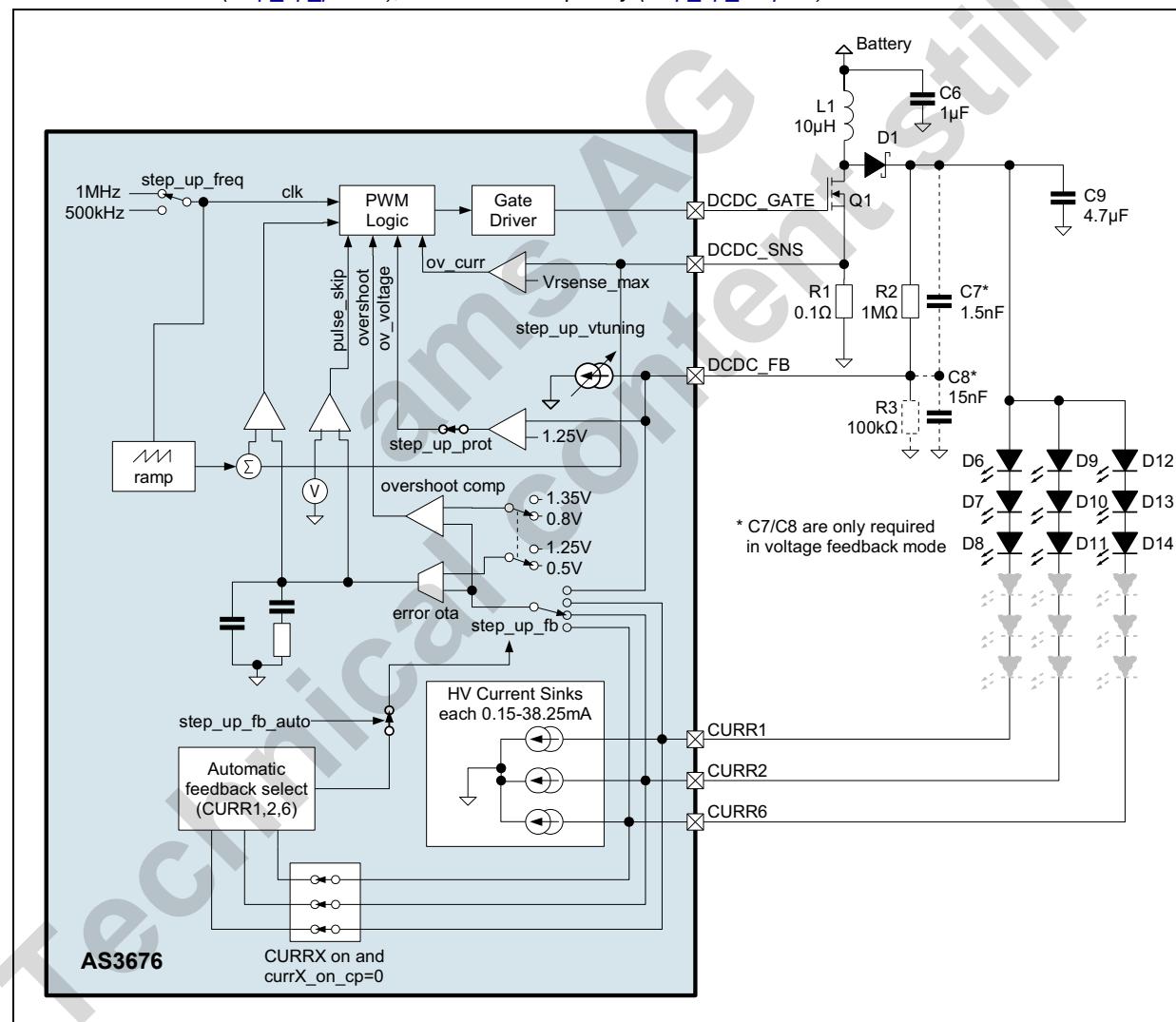
The protection voltage can be calculated according to the following formula:

$$V_{PROTECT} = 1.25V + IDCDC_FB * R2 \quad (EQ\ 1)$$

Note: The voltage on the pin DCDC_FB is limited by an internal protection diode to VBAT + one diode forward voltage (typ. 0.6V).

If the over voltage protection is not used in current feedback mode, connect DCDC_FB to ground.

Figure 12. Step Up DC/DC Converter Detail Diagram; Option: Regulated Output Current, Feedback is automatically selected between CURR1, CURR2, CURR6 (`step_up_fb_auto`=1); over voltage protection is enabled (`step_up_prot`=1); 1MHz clock frequency (`step_up_freq`=0)



2. If the DCDC converter is only used in current feedback mode (CURR1, CURR2 or CURR6 - and not used for a constant voltage source), the capacitors C7 and C8 can be removed.

8.2.3 Voltage Feedback

Setting bit `step_up_fb` (see page 15) = 00 enables voltage feedback at pin DCDC_FB. Capacitors C7 and C8 have to be soldered in this operating mode.

The output voltage is regulated to a constant value, given by (Bit `step_up_res` should be set to 1 in this configuration)

$$U_{Step\ up\ out} = (R_2 + R_3)/R_3 * 1.25 + I_{DCDC_FB} * R_2 \quad (EQ\ 2)$$

If R3 is not used, the output voltage is by (Bit `step_up_res` should be set to 0 in this configuration)

$$U_{Step\ up\ out} = 1.25 + I_{DCDC_FB} * R_2 \quad (EQ\ 3)$$

Where:

$U_{Step\ up\ out}$ = Step Up DC/DC Converter output voltage

R_2 = Feedback resistor R2

R_3 = Feedback resistor R3

I_{DCDC_FB} = Tuning current at ball DCDC_FB; 0 to 31 μ A

Table 9. Voltage Feedback Example Values

I_{DCDC_FB} µA	$U_{Step\ up\ out}$ $R_2 = 1M\Omega$, R_3 not used	$U_{Step\ up\ out}$ $R_2 = 500k\Omega$, $R_3 = 50k\Omega$
0	-	13.75
1	-	14.25
2	-	14.75
3	-	15.25
4	-	15.75
5	6.25	16.25
6	7.25	16.75
7	8.25	17.25
8	9.25	17.75
9	10.25	18.25
10	11.25	18.75
11	12.25	19.25
12	13.25	19.75
13	14.25	20.25
14	15.25	20.75
15	16.25	21.25
...
30	31.25	28.75
31	32.25	29.25

Note: The voltage on CURR1, CURR2 and CURR6 must not exceed 26V (see page 25)

8.2.4 PCB Layout Hints

To ensure good EMC performance of the DCDC converter, keep its external power components C6, R1, L1, Q1, D1 and C9 close together. Connect the ground of C6, R1 and C9 locally together and connect this with a short path to AS3676 VSS. This ensures that local high-frequency currents will not flow to the battery.

8.2.5 Unused DCDC converter

If the DCDC converter is not used, connect DCDC_SNS to GND. DCDC_FB³ and DCDC_GATE can be left open.

Table 12. DCDC control2 Register

Addr: 22h		DCDC control2		
This register controls the Step Up DC/DC Converter and low-voltage current sinks CURR3x.				
Bit	Bit Name	Default	Access	Description
0	step_up_res	0	R/W	Gain selection for Step Up DC/DC Converter
				0 Select 0 if Step Up DC/DC Converter is used with current feedback (CURR1, CURR2, CURR6) or if DCDC_FB is used with current feedback only – R2, C7, C8 connected, R3 not used
				1 Select 1 if DCDC_FB is used with external resistor divider using 2 resistors: R2 and R3
1	skip_fast	0	R/W	Step Up DC/DC Converter output voltage at low loads, when pulse skipping is active
				0 Accurate output voltage, more ripple
				1 Elevated output voltage, less ripple
2	step_up_prot	1	R/W	Step Up DC/DC Converter protection
				0 No over voltage protection
				1 Over voltage protection on pin DCDC_FB enabled voltage limitation = 1.25V on DCDC_FB
3	step_up_lowcur	0	R/W	Step Up DC/DC Converter coil current limit
				0 Normal current limit
				1 Current limit reduced by approx. 33%
7	step_up_fb_auto	0	R/W	step_up_fb selects the feedback of the DCDC converter
				1 If step_up_fb is not DCDC_FB (00), then feedback is automatically chosen within the current sinks CURR1, CURR2 and CURR6. Only those are used for this selection, which are enabled (currX_mode must not be 00) and not connected to the charge pump (currX_on_cp must be 0).

8.3 Charge Pump

The Charge Pump uses two external flying capacitors C3, C4 to generate output voltages higher than the battery voltage. There are three different operating modes of the charge pump itself:

- 1:1 Bypass Mode
 - Battery input and output are connected by a low-impedance switch
 - battery current = output current.
- 1:1.5 Mode
 - The output voltage is up to 1.5 times the battery voltage (without load), but is limited to VCPOUTmax all the time
 - battery current = 1.5 times output current.
- 1:2 Mode
 - The output voltage is up to 2 times the battery voltage (without load), but is limited to VCPOUTmax all the time
 - battery current = 2 times output current

As the battery voltage decreases, the Charge Pump must be switched from 1:1 mode to 1:1.5 mode and eventually in 1:2 mode in order to provide enough supply for the current sinks. Depending on the actual current the mode with best overall efficiency can be automatically or manually selected:

Examples:

- Battery voltage = 3.7V, LED dropout voltage = 3.5V. The 1:1 mode will be selected and there is 200mV drop on the current sink and on the Charge Pump switch. Efficiency 95%.
- Battery voltage = 3.5V, LED dropout voltage = 3.5V. The 1:1.5 mode will be selected and there is 1.5V drop on the current sink and 250mV on the Charge Pump. Efficiency 66%.

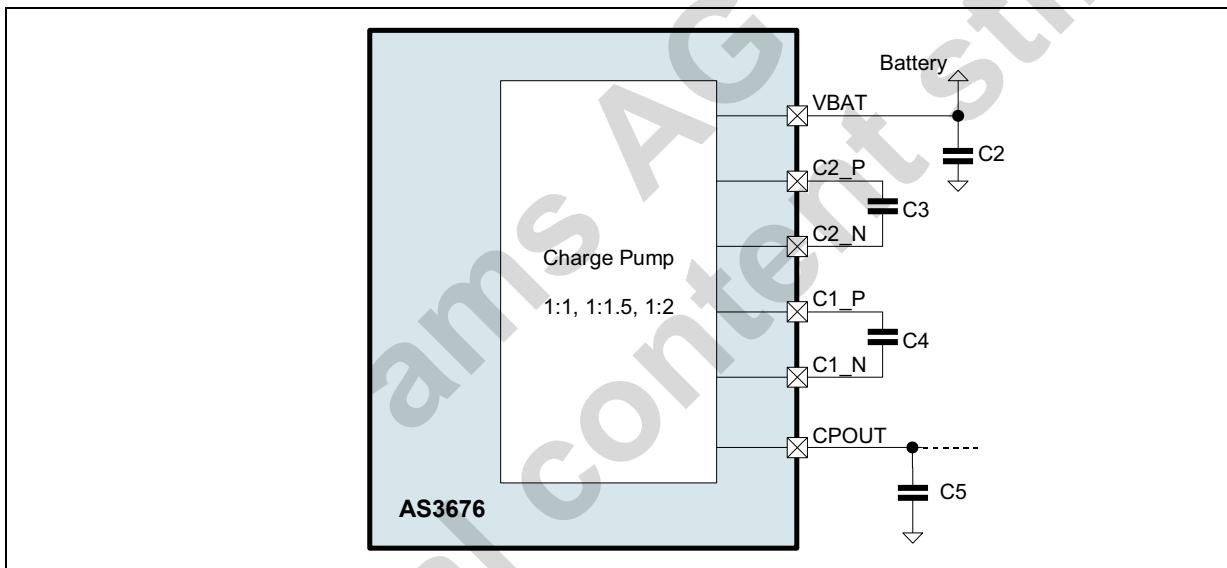
The efficiency is dependent on the LED forward voltage given by:

$$Eff = (V_{LED} \cdot I_{out}) / (U_{in} \cdot I_{in}) \quad (EQ\ 4)$$

The charge pump mode switching can be done manually or automatically with the following possible software settings:

- Automatic up all modes allowed (1:1, 1:1.5, 1:2)
 - Start with 1:1 mode
 - Switch up automatically 1:1 to 1:1.5 to 1:2
- Automatic up, but only 1:1 and 1:1.5 allowed
 - Start with 1:1 mode
 - Switch up automatically only from 1:1 to 1:1.5 mode; 1:2 mode is not used
- Manual
 - Set modes 1:1, 1:1.5, 1:2 by software

Figure 13. Charge Pump Pin Connections



The Charge Pump requires the external components listed in the following table:

Table 13. Charge Pump External Components

Symbol	Parameter	Condition	Min	Typ	Max	Unit
C ₂	External Decoupling Capacitor	Ceramic low-ESR capacitor between pins VBAT and VSS.		1.0		µF
C ₃ , C ₄	External Flying Capacitor (2x)	Ceramic low-ESR capacitor between pins C _{1_P} and C _{1_N} , between pins C _{2_P} and C _{2_N} and between VBAT and VSS		1.0		µF
C ₅	External Storage Capacitor	Ceramic low-ESR capacitor between pins CPOUT and VSS, pins CPOUT and VSS. Use nominal 2.2µF capacitors (size 0603)		2.2		µF

Note: The connections of the external capacitors C₂, C₃, C₄ and C₅ should be kept as short as possible.

The maximum voltage on the flying capacitors C3 and C4 is VBAT.

Table 14. Charge Pump Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Unit
ICPOUT	Output Current Continuous	Depending on PCB layout	0.0		400	mA
	Output Current Pulsed	max. 200ms $V_{CPOUT} = V_{BAT} * CPMODE - ILOAD * RCP$	0.0		500	mA
VCPOUTmax	Output Voltage	Internally limited, Including output ripple			5.6	V
η	Efficiency	Including current sink loss; ICPOUT < 100mA.	60		90	%
ICP1_1.5	Power Consumption without Load fclk = 1 MHz	1:1.5 Mode		3.4		mA
ICP1_2		1:2 Mode		3.8		
Rcp1_1	Effective Charge Pump Output Resistance (Open Loop, fclk = 1MHz)	1:1 Mode; VBAT \geq 3.5V		0.57		Ω
Rcp1_1.5		1:1.5 Mode; VBAT \geq 3.3V		2.65		
Rcp1_2		1:1.2 Mode; VBAT \geq 3.1V		3.25		
fclk Accuracy	Accuracy of Clock Frequency		-10		10	%
currhv_switch	CURR1, 2, 6 minimum voltage				0.45	V
curriv_switch	CURR30-33, RGB1-3, CURR41-3, minimum voltage	If the voltage drops below this threshold, the charge pump will use the next available mode (1:1 \rightarrow 1:1.5 or 1:1.5 \rightarrow 1:2)			0.2	V
	CURR30-33 0-75.6mA range for strobe if curr3x_strobe_high=1				0.4	V
tdeb	CP automatic up-switching debounce time	cp_start_debounce=0		240		μ sec
		After switching on CP (cp_on set to 1), if cp_start_debounce=1		2000		μ sec

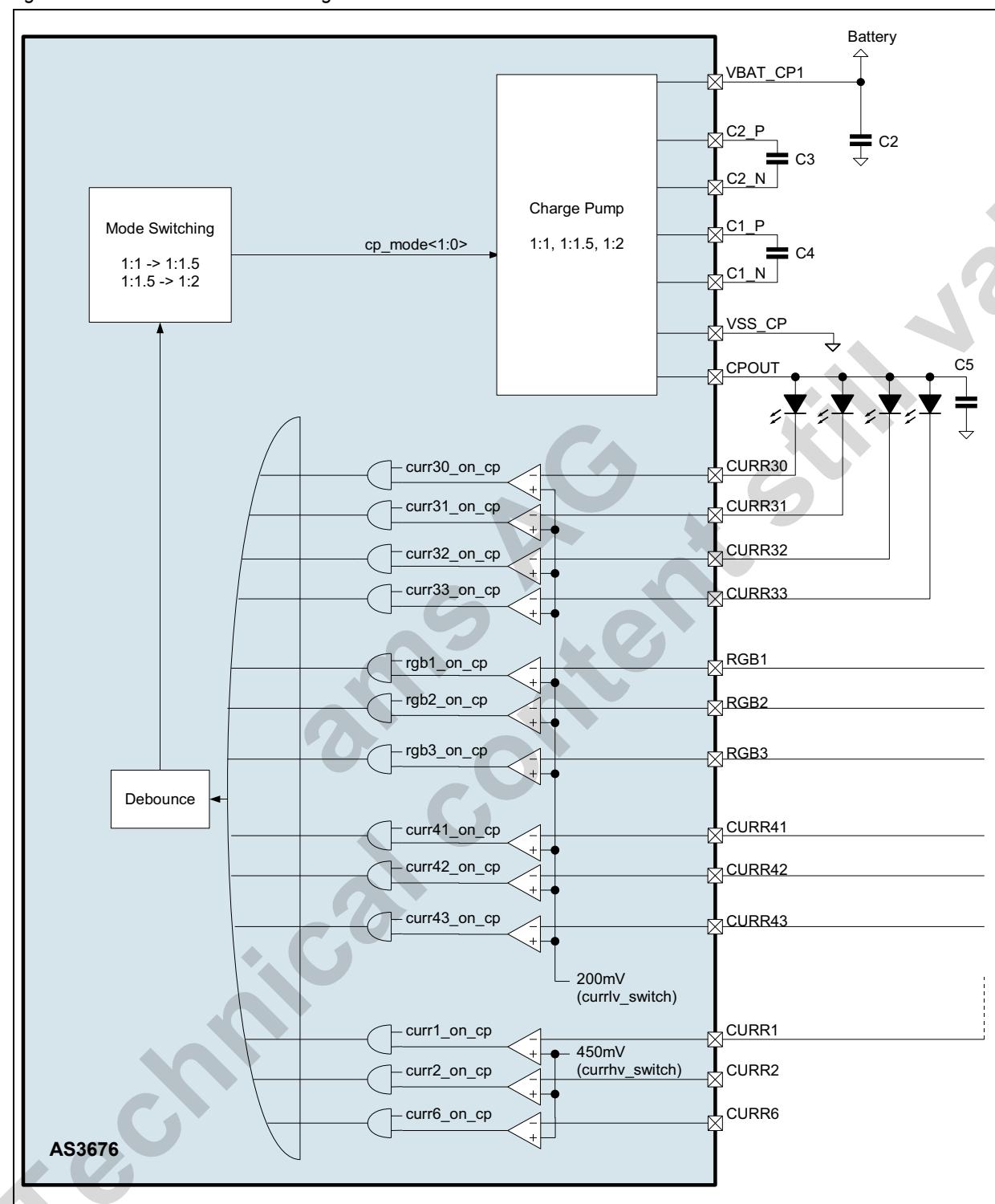
8.3.1 Charge Pump Mode Switching

If automatic mode switching is enabled ([cp_mode_switching](#) (see page 20) = 00 or [cp_mode_switching](#) = 01) the charge pump monitors the current sinks, which are connected via a led to the output CPOUT. To identify these current sources (sinks), the registers [CP mode Switch1](#) and [CP mode Switch2](#) (register bits [curr30_on_cp](#) (see page 21) ... [curr33_on_cp](#), [rgb1_on_cp](#) ... [rgb3_on_cp](#), [curr1_on_cp](#), [curr2_on_cp](#), [curr41_on_cp](#) ... [curr43_on_cp](#) and [curr6_on_cp](#)) should be setup before starting the charge pump ([cp_on](#) (see page 20) = 1). If any of the voltage on these current sources drops below the threshold (curriv_switch, currhv_switch), the next higher mode is selected after the debounce time.

To avoid switching into 1:2 mode (battery current = 2 times output current), set [cp_mode_switching](#) = 01.

If the currX_on_cp=0 and the according current sink is connected to the charge pump, the current sink will be functional, but there is no up switching of the charge pump, if the voltage compliance is too low for the current sink to supply the specified current.

Figure 14. Automatic Mode Switching



8.3.2 Soft Start

An implemented soft start mechanism reduces the inrush current. Battery current is smoothed when switching the charge pump on and also at each switching condition. This precaution reduces electromagnetic radiation significantly.

Table 20. *Curr low voltage status2 Register (Continued)*

Addr: 2Bh		Curr low voltage status2			
		Indicates the low voltage status of the current sinks. If the currX_low_v bit is set, the voltage on the current sink is too low, to drive the selected output current			
Bit	Bit Name	Default	Access	Description	
1	curr2_low_v	NA	R	0	voltage of current Sink CURR2 >currhv_switch
				1	voltage of current Sink CURR2 <currhv_switch
2	curr41_low_v	NA	R	0	voltage of current Sink CURR41 >currlv_switch
				1	voltage of current Sink CURR41 <currlv_switch
3	curr42_low_v	NA	R	0	voltage of current Sink CURR42 >currlv_switch
				1	voltage of current Sink CURR42 <currlv_switch
4	curr43_low_v	NA	R	0	voltage of current Sink CURR43 >currlv_switch
				1	voltage of current Sink CURR43 <currlv_switch

8.4 Current Sinks

The AS3676 contains general purpose current sinks intended to control RGB LEDs, white LEDs (e.g. backlights) and can also be used for buzzers or vibrators. All current sinks have an integrated over voltage protection.

CURR1, CURR2 and CURR6 are also used as feedback for the Step Up DC/DC Converter (regulated to 0.5V in this configuration) see [Feedback Selection on page 12](#).

- Current sinks CURR1, CURR2 and CURR6 are high-voltage compliant (26V) current sinks, used e.g., for series of white LEDs
- Current sinks CURR3x (CURR30, CURR31, CURR32 and CURR33) are parallel 5V current sinks, used for backlighting, indicator LEDs or RGB LEDs.
- Current sinks RGB1, RGB2, and RGB3 are general purpose current sinks e.g. for a fun LED.
- Current sinks CURR4x (CURR41, CURR42, and CURR43) are general purpose current sinks.

Table 21. Current Sink Function Overview

Current Sink	Max. Voltage (V)	Max. Current (mA)	Resolution		Software Current Control	Hardware On/Off Control	Can be assigned to Audio Controlled LED Channel
			(Bits)	(mA)			
CURR1	26.0	38.25	8	0.15	Separate	LED Pattern; Internal PWM; external PWM at GPIO1/DLS	ch1
CURR2							ch2
CURR6							ch3
CURR30	VBAT (5.5V)	38.25 (75.6mA for strobe if curr3x_strobe_high=1)	8	0.15	Combined in Strobe/ Preview or Separated	Flash LED Strobe (CURR1 or CURR30) & Preview (CURR2); Internal PWM; LED Pattern; external PWM at GPIO1/DLS	Completely individual assignment of the audio channels ch1, ch2 and ch3 to the outputs
CURR31							
CURR32							
CURR33							
RGB1		38.25	8	0.15	Separate	LED Pattern; Internal PWM; external PWM at GPIO1/DLS	ch1
RGB2							ch2
RGB3							ch3
CURR41		38.25	8	0.15	Separate	LED Pattern; Internal PWM; external PWM at GPIO1/DLS	ch1
CURR42							ch2
CURR43							ch3

8.4.1 Unused Current Sinks

Unused current sinks can be left open or used as a ADC input (see [Analog-to-Digital Converter on page 65](#)).

8.4.2 High Voltage Current Sinks CURR1, CURR2, CURR6

The high voltage current sinks have a resolution of 8 bits.

Table 22. HV Current Sinks Characteristics

Symbol	Parameter	Condition		Min	Typ	Max	Unit
I _{BIT7}	Current sink if Bit7 = 1	For V(CURRx) > 0.45V			19.2		mA
I _{BIT6}	Current sink if Bit6 = 1				9.6		
I _{BIT5}	Current sink if Bit5 = 1				4.8		
I _{BIT4}	Current sink if Bit4 = 1				2.4		
I _{BIT3}	Current sink if Bit3 = 1				1.2		
I _{BIT2}	Current sink if Bit2 = 1				0.6		
I _{BIT1}	Current sink if Bit1 = 1				0.3		
I _{BIT0}	Current sink if Bit0 = 1				0.15		
Δm^1	matching Accuracy	CURR1,CURR2,CURR6	I _{BIT7} and I _{BIT6}	-5		+5	%
			all other bits	-10		+10	%
Δ^2	absolute Accuracy			-15		+15	%
V _{CURR1,2,6x}	Voltage compliance			0.45		26	V

1. Variation between currents within this group
2. Variation between the programmed current and the actual current

High Voltage Current Sinks CURR1, CURR2, CURR6 Registers

Table 23. Curr1 current Register

Addr: 09h		Curr1 current			
		This register controls the High voltage current sink current.			
Bit	Bit Name	Default	Access	Description	
7:0	curr1_current	0	R/W	Defines current into current sink curr1	
				00h	0 mA
				01h	0.15 mA
			
				FFh	38.25 mA

Table 24. Curr2 current Register

Addr: 0Ah		Curr2 current			
		This register controls the High voltage current sink current.			
Bit	Bit Name	Default	Access	Description	
7:0	curr2_current	0	R/W	Defines current into current sink curr2	
				00h	0 mA
				01h	0.15 mA
			
				FFh	38.25 mA

Table 25. curr6 current Register

Addr: 2Fh		curr6 current			
		This register controls the High voltage current sink current.			
Bit	Bit Name	Default	Access	Description	
7:0	curr6_current	0	R/W	Defines current into current sink CURR6	
				00h	0 mA
				01h	0.15 mA
			
				FFh	38.25 mA

Table 26. curr12 control Register

Addr: 01h		curr12 control			
		This register select the mode of the current sinks controls High voltage current sink current.			
Bit	Bit Name	Default	Access	Description	
1:0	curr1_mode	0	R/W	Select the mode of the current sink curr1	
				00b	off
				01b	on
				10b	PWM controlled
				11b	LED pattern controlled
3:2	curr2_mode	0	R/W	Select the mode of the current sink curr2	
				00b	off
				01b	on
				10b	PWM controlled
				11b	LED pattern controlled

Table 27. curr rgb control Register

Addr: 02h		curr rgb control			
		This register select the mode of the current sinks CURR6.			
Bit	Bit Name	Default	Access	Description	
7:6	curr6_mode	0	R/W	Select the mode of the current sink CURR6	
				00b	off
				01b	on
				10b	PWM controlled
				11b	LED pattern controlled

8.4.3 Current Sinks CURR30, CURR31, CURR32, CURR33

These current sinks have a resolution of 8 bits and can sink up to 38.25mA. The current values can be controlled individually with `curr30_current` – `curr33_current` or common with `curr3x_strobe` or `curr3x_preview`.

Table 28. Current Sinks CURR30,31,32,33 Parameters

Symbol	Parameter	Condition		Min	Typ	Max	Unit
I_{BIT7}	Current sink if Bit7 = 1	For $V(CURR3x) > 0.2V$			19.2		mA
I_{BIT6}	Current sink if Bit6 = 1				9.6		
I_{BIT5}	Current sink if Bit5 = 1				4.8		
I_{BIT4}	Current sink if Bit4 = 1				2.4		
I_{BIT3}	Current sink if Bit3 = 1				1.2		
I_{BIT2}	Current sink if Bit2 = 1				0.6		
I_{BIT1}	Current sink if Bit1 = 1				0.3		
I_{BIT0}	Current sink if Bit0 = 1				0.15		
Δm^1	matching Accuracy	CURR30-33	I_{BIT7} and I_{BIT6}	-5		+5	%
			all other bits	-10		+10	%
Δ^2	absolute Accuracy			-15		+15	%
V_{CURR3X}	Voltage compliance			0.2		CPO UT	V
		<code>curr3x_strobe_high=1</code> and strobe function		0.4			

1. Variation between currents within this group
2. Variation between the programmed current and the actual current

Current Sinks CURR3x Registers

Table 29. `Curr3 control2` Register

Addr: 12h		Curr3 control2			
		This register selects the modes of the current sinks30..33 current.			
Bit	Bit Name	Default	Access	Description	
0	<code>preview_off_after_strobe</code>	0b	R/W	Select the switch off mode after strobe pulse	
				0	normal preview/strobe mode
				1	switch off preview after strobe duration has expired. To reinitiate the torch mode the <code>preview_ctrl</code> has to be set off and on again
2:1	<code>preview_ctrl</code>	00b	R/W	Preview is triggered by	
				00b	off
				01b	software trigger (setting this bit automatically triggers preview)
				10b	CURR2 active high; set <code>gpi_curr2_en=1</code>
				11b	CURR2 active low; set <code>gpi_curr2_en=1</code>
5	<code>curr3x_strobe_high</code>	0b	R/W	Double current on CURR30..CURR33 during strobe function	
				0	normal strobe current (0-37.8mA)
				1	double strobe current (0-75.6mA)

Table 29. *Curr3 control2* Register (Continued)

Addr: 12h		Curr3 control2			
		This register selects the modes of the current sinks30..33 current.			
Bit	Bit Name	Default	Access	Description	
7	strobe_pin	0	R/W	Select strobe input pin and current sink outputs (only if <i>strobe_ctrl</i> =10 or 11)	
				0	CURR1 is strobe input; CURR30...CURR33 flash output; set gpi_curr1_en=1
				1	CURR30 is strobe input; CURR1, CURR2, CURR6 flash output; set gpi_curr30_en=1

Table 30. *Curr3 strobe control* Register

Addr: 11h		Curr3 strobe control			
		This register selects the modes of the current sinks30..33 current.			
Bit	Bit Name	Default	Access	Description	
1:0	strobe_ctrl	00b	R/W	Strobe is triggered by	
				00b	off
				01b	software trigger (setting this bit automatically triggers strobe)
				10b	CURR1 (or CURR30 see <i>strobe_pin</i>) active high
				11b	CURR1 (or CURR30 see <i>strobe_pin</i>) active low
3:2	strobe_mode	00b	R/W	Selects strobe mode	
				00b	Mode1 (Tstrobe=Ts; strobe trigger signal $\geq 10\mu s$)
				01b	Mode 2 (Tstrobe=max Ts)
				10b	Mode 3 (Tstrobe = strobe signal)
				11b	not used

Table 30. *Curr3 strobe control* Register (Continued)

Addr: 11h		Curr3 strobe control			
		This register selects the modes of the current sinks30..33 current.			
Bit	Bit Name	Default	Access	Description	
7:4	strobe_timing	0000b	R/W	Selects strobe time (Ts)	
				0000b	100 msec
				0001b	200 msec
				0010b	300 msec
				0011b	400 msec
				0100b	500 msec
				0101b	600 msec
				0110b	700 msec
				0111b	800 msec
				1000b	900 msec
				1001b	1000 msec
				1010b	1100 msec
				1011b	1200 msec
				1100b	1300 msec
				1101b	1400 msec
				1110b	1500 msec
				1111b	1600 msec

Table 31. *Curr3x strobe* Register

Addr: 0Eh		Curr3x strobe			
		This register selects the strobe current of the current sinks30..33			
Bit	Bit Name	Default	Access	Description	
5:0	curr3x_strobe	00	R/W	Defines Strobe current of Current sinks curr30-33	
				00h	0 mA
				01h	0.6 mA (1.2mA if curr3x_strobe_high=1)
			
				3Fh	37.8 mA (75.6mA if curr3x_strobe_high=1)

Table 32. *Curr3x preview* Register

Addr: 0Fh		Curr3x preview			
		This register selects the preview current of the current sinks30..33			
Bit	Bit Name	Default	Access	Description	
5:0	curr3x_preview	00	R/W	Defines Preview current of Current sinks curr30-33	
				00h	0 mA
				01h	0.6 mA
			
				3Fh	37.8 mA

Table 37. Curr33 current Register

Addr: 43h		Curr33 current		
		This register selects the current of the current sink33		
Bit	Bit Name	Default	Access	Description
7:0	curr33_current	00	R/W	Selects curr33 current, if curr33 is not used for strobe/preview (curr33_mode=11b)
				00h 0 mA
				01h 0.15 mA
			
				FFh 38.25 mA

Table 38. curr3 control1 Register

Addr: 03h		curr3 control1		
		This register select the mode of the current sinks30 - 33		
Bit	Bit Name	Default	Access	Description
1:0	curr30_mode	0	R/W	Select the mode of the current sink curr30
				00b off
				01b strobe/preview
				10b curr30_current or curr3x_other PWM controlled
				11b curr30_current or curr3x_other - don't use curr3x_other if softdim_pattern=1, use curr30_current instead
3:2	curr31_mode	0	R/W	Select the mode of the current sink curr31
				00b off
				01b strobe/preview
				10b curr31_current or curr3x_other PWM controlled
				11b curr31_current - don't use curr3x_other if softdim_pattern=1, use curr31_current instead
5:4	curr32_mode	0	R/W	Select the mode of the current sink CURR32
				00b off
				01b strobe/preview
				10b curr32_current or curr3x_other PWM controlled
				11b curr32_current or curr3x_other - don't use curr3x_other if softdim_pattern=1, use curr32_current instead
7:6	curr33_mode	0	R/W	Select the mode of the current sink curr33
				00b off
				01b strobe/preview
				10b curr33_current or curr3x_other PWM controlled
				11b curr33_current or curr3x_other - don't use curr3x_other if softdim_pattern=1, use curr33_current instead

Table 39. *Pattern control* Register

Addr: 18h		Pattern control			
		This register controls the LED pattern			
Bit	Bit Name	Default	Access	Description	
4	curr30_pattern	0b	R/W	Additional CURR33 LED pattern control bit	
				0b	CURR30 controlled according curr30_mode register
				1b	CURR30 controlled by LED pattern generator
5	curr31_pattern	0b	R/W	Additional CURR33 LED pattern control bit	
				0b	CURR31 controlled according curr31_mode register
				1b	CURR31 controlled by LED pattern generator
6	curr32_pattern	0b	R/W	Additional CURR33 LED pattern control bit	
				0b	CURR32 controlled according curr32_mode register
				1b	CURR32 controlled by LED pattern generator
7	curr33_pattern	0b	R/W	Additional CURR33 LED pattern control bit	
				0b	CURR33 controlled according curr33_pattern register
				1b	CURR33 controlled by LED pattern generator

Table 42. *Rgb1 current Register*

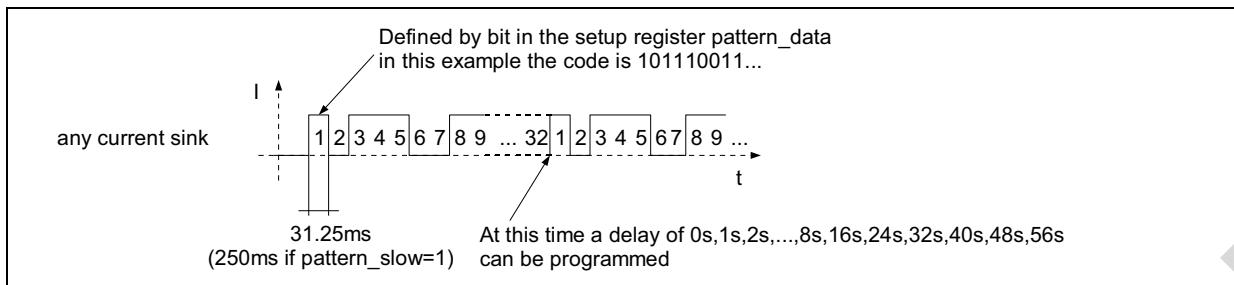
Addr: 0Bh		Rgb1 current			
		This register controls the RGB current sink current.			
Bit	Bit Name	Default	Access	Description	
7:0	rgb1_current	0	R/W	Defines current into Current sink RGB1	
				00h	0 mA
				01h	0.15 mA
			
				FFh	38.25 mA

Table 43. *Rgb2 current Register*

Addr: 0Ch		Rgb2 current			
		This register controls the RGB current sink current.			
Bit	Bit Name	Default	Access	Description	
7:0	rgb2_current	0	R/W	Defines current into Current sink RGB2	
				00h	0 mA
				01h	0.15 mA
			
				FFh	38.25 mA

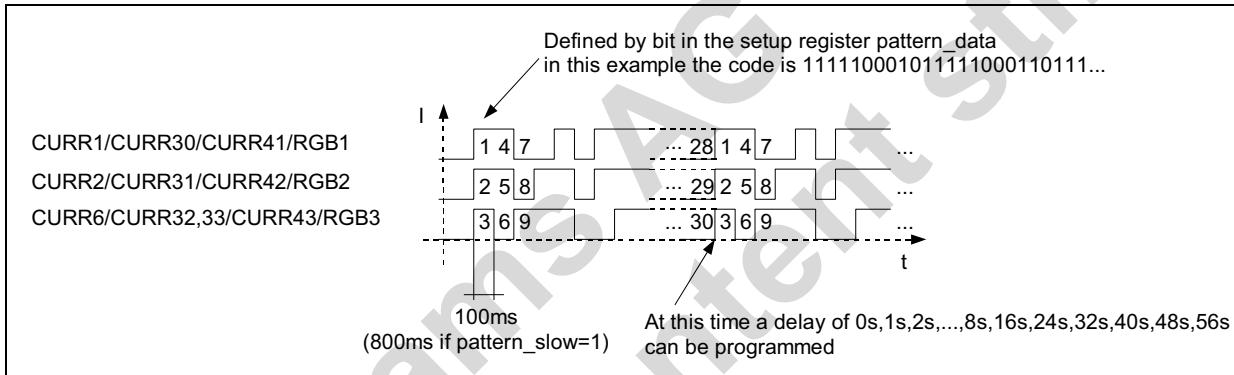
Table 44. *Rgb3 current Register*

Addr: 0Dh		Rgb3 current			
		This register controls the RGB current sink current.			
Bit	Bit Name	Default	Access	Description	
7:0	rgb3_current	0	R/W	Defines current into Current sink RGB3	
				00h	0 mA
				01h	0.15 mA
			
				FFh	38.25 mA

Figure 15. LED Pattern Generator AS3676 for *pattern_color* = 0

To select the different current sinks to be controlled by the LED pattern generator, see the 'xxxx'_mode registers (where 'xxxx' stands for the to be controlled current sink, e.g. curr1_mode for CURR1 current sink). See also the description of the different current sinks.

To allow the generator of a color patterns set the bit *pattern_color* to '1'. Then the pattern can be connected to CURRx as follows:

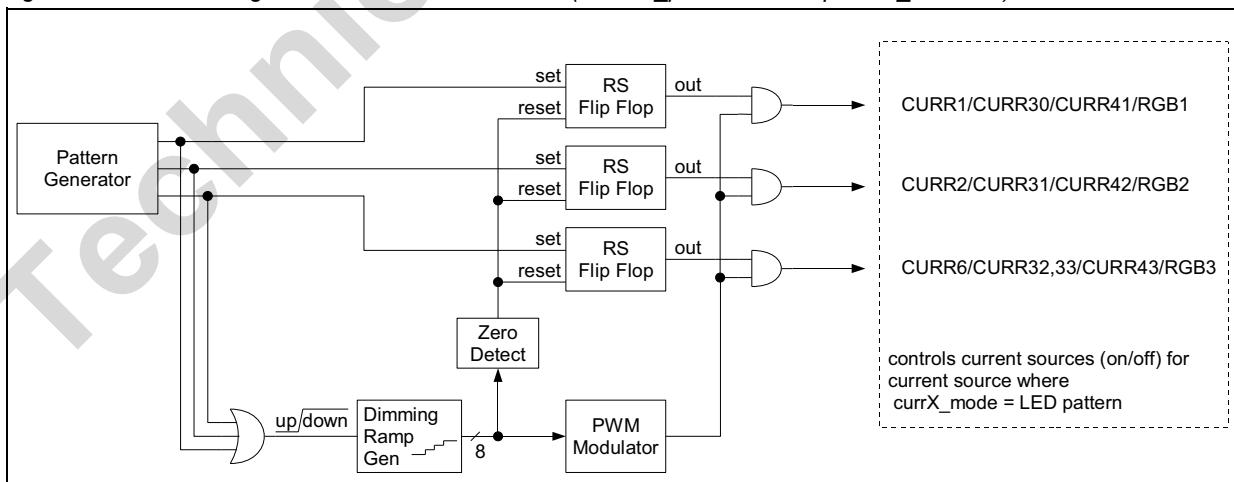
Figure 16. LED Pattern Generator AS3676 for *pattern_color* = 1

Only those current sinks will be controlled, where the 'xxxx'_mode register is configured for LED pattern.

If the register bit *pattern_slow* is set, all pattern times are increased by a factor of eight. (bit duration: 250ms if *pattern_color*=0 / 800ms if *pattern_color*=1, delays between pattern up to 56s).

Soft Dimming for Pattern

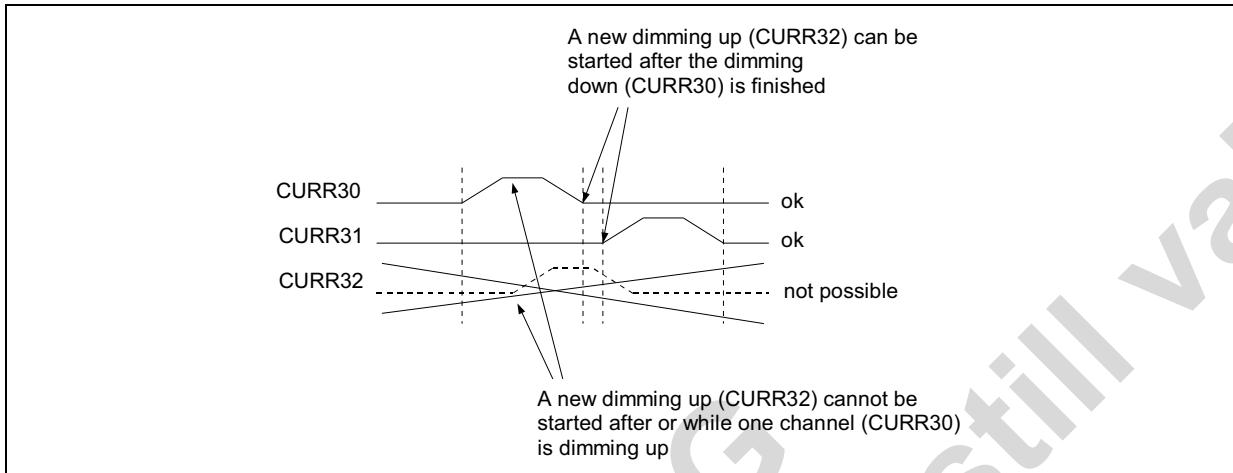
The internal pattern generator can be combined with the internal pwm dimming modulator to obtain as shown in the following figure:

Figure 17. Soft dimming Architecture for the AS3676 (*softdim_pattern*=1 and *pattern_color* = 1)

With the AS3676 smooth fade-in and fade-out effects can be automatically generated.

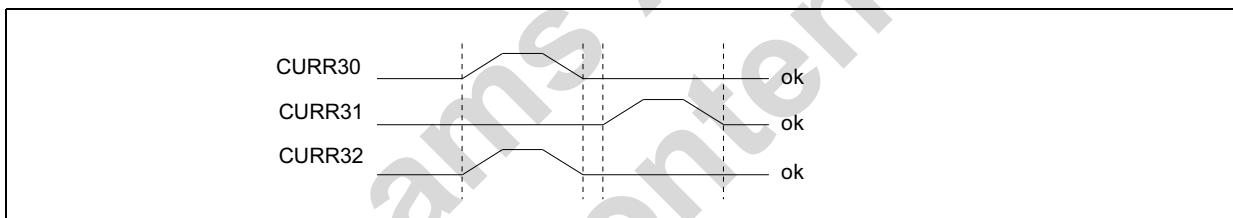
As there is only one dimming ramp generator and one pwm modulator following constraints have to be considered when setting up the pattern (applies only if `pattern_color=1`):

Figure 18. Soft dimming example Waveform for CURR30-32



However using the identical dimming waveform for two channels is possible as shown in the following figure:

Figure 19. Soft dimming example Waveform for CURR30-32



LED Pattern Registers

Table 50. Pattern data0...Pattern data3 Registers

Addr: 19h,1Ah,1Bh,1Ch		Pattern data0, Pattern data1, Pattern data2, Pattern data3			
		This registers contains the pattern data for the current sinks.			
Bit	Bit Name	Default	Access	Description	
7:0	pattern_data[7:0] ¹	0	R/W	Pattern data0	
7:0	pattern_data[15:8]	0	R/W	Pattern data1	
7:0	pattern_data[23:16]	0	R/W	Pattern data2	
7:0	pattern_data[31:24]	0	R/W	Pattern data3	

1. Update any of the pattern register only if none of the current sources is connected to the pattern generator ('xxxx'_mode must not be 11b). The pattern generator is automatically started at the same time when any of the current sources is connected to the pattern generator

Table 54. LED Pattern timing

pattern_slow	pattern_delay2	pattern_delay[1..0]	bit duration [ms]		delay [s] between patterns	pattern duration [s] (total cycle time: pattern + delay)
	delay between patterns		pattern_color=0	pattern_color=1		
0	0	10	31	100	2	3
0	0	11	31	100	3	4
0	1	00	31	100	4	5
0	1	01	31	100	5	6
0	1	10	31	100	6	7
0	1	11	31	100	7	8
1	0	00	250	800	0	8
1	0	01	250	800	8	16
1	0	10	250	800	16	24
1	0	11	250	800	24	32
1	1	00	250	800	32	40
1	1	01	250	800	40	48
1	1	10	250	800	48	56
1	1	11	250	800	56	64

1. Even by setting 000 for pattern delay, there is a small delay before the new patterns starts.

8.4.7 PWM Generator

The PWM generator can be used for any current sink. The setting applies for all current sinks, which are controlled by the pwm generator (e.g. CURR1 is pwm controlled if curr1_mode = 10). The pwm modulated signal can switch on/off the current sinks and therefore depending on its duty cycle change the brightness of an attached LED.

Internal PWM Generator

The internal PWM generator uses the 2MHz internal clock as input frequency and its dimming range is 6 bits digital (2MHz / 2^6 = 31.3kHz pwm frequency) and 2 bits analog. Depending on the actual code in the register `pwm_code` the following algorithm is used:

If `pwm_code` bit 7 = 1

Then the upper 6 bits (Bits 7:2) of `pwm_code` are used for the 6 bits PWM generation, which controls the selected currents sinks directly

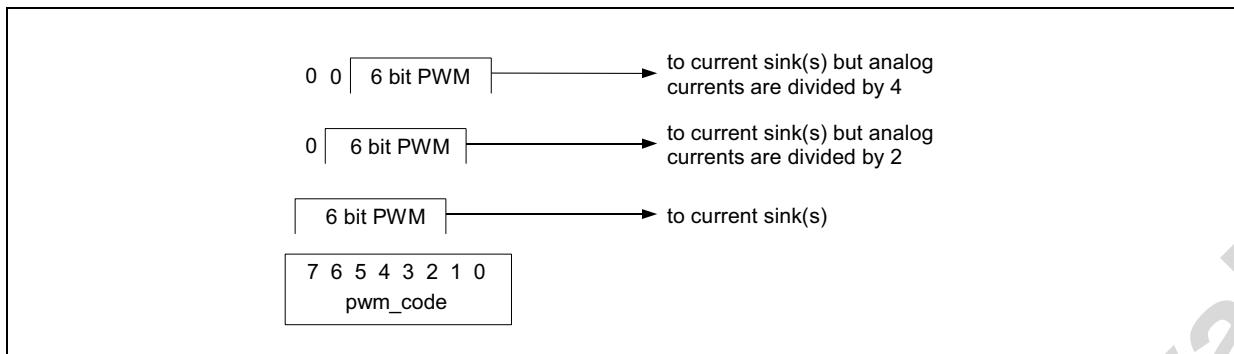
If `pwm_code` bit 7 =0 and bit 6 = 1

Then bits 6:1 of `pwm_code` are used for the 6 bits PWM generation. This signal controls the selected current sinks, but the analog current of these sinks is divided by 2

If `pwm_code` bit 7 and bit 6 = 0

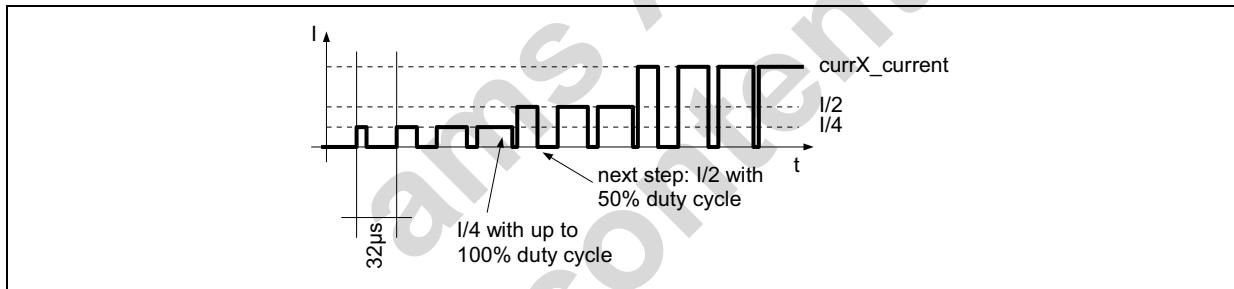
Then bits 5:0 of `pwm_code` are used for the 6 bits PWM generation. This signal controls the selected current sinks, but the analog current of these sinks is divided by 4

Figure 20. PWM Control

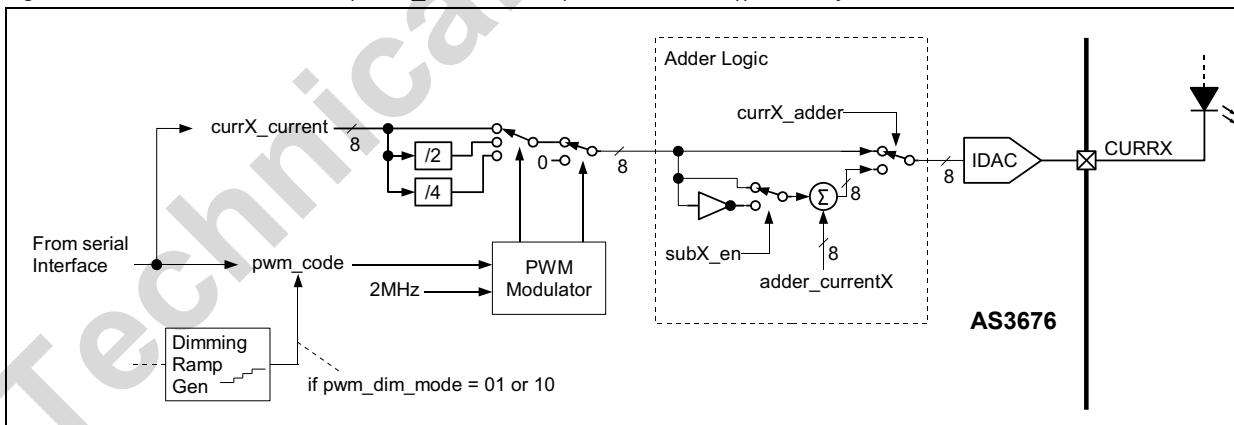
**Automatic Up/Down Dimming**

If the register `pwm_dim_mode` is set to 01 (up dimming) or 10 (down dimming) the value within the register `pwm_code` is increased (up dimming) or decreased (down dimming) every time and amount (either 1/4th or 1/8th) defined by the register `pwm_dim_speed`. The maximum value of 255 (completely on) and the minimum value of 0 (off) is never exceeded. It is used to smoothly and automatically dim the brightness of the LEDs connected to any of the current sinks. The PWM code is readable all the time (also during up and down dimming).

The waveform for up dimming looks as follows (cycles omitted for simplicity):

Figure 21. PWM Dimming Waveform for up dimming (`pwm_dim_mode` = 01); `currX_mode` = PWM controlled (not all steps shown)

The internal pwm modulator circuit controls the current sinks as shown in the following figure:

Figure 22. PWM Control Circuit (`currX_mode` = 10b (PWM controlled)); X = any current sink

The adder logic (available for all current sinks) is intended to allow dimming not only from 0% to 100% (or 100% to 0%) of `currX_current`, but also e.g. from 10% to 110% (or 110% to 10%) of `currX_current`. The starting current for up dimming is defined by $0 + \text{currX_adder}$ and the end current is defined by $\text{currX_current} + \text{currX_adder}$.

Table 55. PWM Dimming Table

	Decrease by 1/4th every step		Decrease by 1/8th every step		Seconds	Seconds	Seconds	Seconds
Step	%Dimming	PWM	%Dimming	PWM	50msec/Step	25msec/Step	5msec/Step	2.5msec/Step
25			5,5	14	1,20s	0,60s	0,120s	0,060s
26			5,1	13	1,25s	0,63s	0,125s	0,063s
27			4,7	12	1,30s	0,65s	0,130s	0,065s
28			4,3	11	1,35s	0,68s	0,135s	0,068s
29			3,9	10	1,40s	0,70s	0,140s	0,070s
30			3,5	9	1,45s	0,73s	0,145s	0,073s
31			3,1	8	1,50s	0,75s	0,150s	0,075s
32			2,7	7	1,55s	0,78s	0,155s	0,078s
33			2,4	6	1,60s	0,80s	0,160s	0,080s
34			2,0	5	1,65s	0,83s	0,165s	0,083s
35			1,6	4	1,70s	0,85s	0,170s	0,085s
36			1,2	3	1,75s	0,88s	0,175s	0,088s
37			0,8	2	1,80s	0,90s	0,180s	0,090s
38			0,4	1	1,85s	0,93s	0,185s	0,093s
39			0,0	0	1,90s	0,95s	0,190s	0,095s

PWM Generator Registers

Table 56. Pwm control Register

Addr: 16h		Pwm control				
		This register controls PWM generator				
Bit	Bit Name	Default	Access	Description		
2:1	pwm_dim_mode	00b	R/W	Selects the dimming mode		
				00b	no dimming; actual content of register pwm_code is used for pwm generator	
				01b	logarithmic up dimming (codes are increased). Start value is actual pwm_code	
				10b	logarithmic down dimming (codes are decreased). Start value is actual pwm_code ; switch off the dimmed current source after dimming is finished to avoid unnecessary quiescent current	
				11b	NA	

Table 56. Pwm control Register (Continued)

Addr: 16h		Pwm control		
		This register controls PWM generator		
Bit	Bit Name	Default	Access	Description
5:3	pwm_dim_speed	000b	R/W	Defines dimming speed by increase/decrease pwm_code
				000b by 1/4 th every 50 msec (total dim time 1.0s)
				001b by 1/8 th every 50 msec (total dim time 1.9s)
				010b by 1/4 th every 25 msec (total dim time 0.5s)
				011b by 1/8 th every 25 msec (total dim time 0.95s)
				100b by 1/4 th every 5 msec (total dim time 100ms)
				101b by 1/8 th every 5 msec (total dim time 190ms)
				110b by 1/4 th every 2.5 msec (total dim time 50ms)
				111b by 1/8 th every 2.5 msec (total dim time 95ms)

Table 57. pwm code Register

Addr: 17h		pwm code		
		This register controls the Pwm code.		
Bit	Bit Name	Default	Access	Description
7:0	pwm_code	00b	R/W	Selects the PWM code
				00h 0% duty cycle
			
				FFh 100% duty cycle

Table 58. Adder Current 1 Register

Addr: 30h		Adder Current 1		
		This register defines the current which can be added to CURR1, CURR30, CURR41, RGB1		
Bit	Bit Name	Default	Access	Description
7:0	adder_current1	00b	R/W	Selects the added current value – do not exceed together with currX_current the internal 8 Bit range (see text)
				00h 0 (represents 0mA)
			
				FFh 255 (represents 38.25mA)

Table 59. Adder Current 2 Register

Addr: 31h		Adder Current 2			
		This register defines the current which can be added to CURR2, CURR31, CURR42, RGB2			
Bit	Bit Name	Default	Access	Description	
7:0	adder_current2	00b	R/W	Selects the added current value – do not exceed together with currX_current the internal 8 Bit range (see text)	
				00h	0 (represents 0mA)
			
				FFh	255 (represents 38.25mA)

Table 60. Adder Current 3 Register

Addr: 32h		Adder Current 3			
		This register defines the current which can be added to CURR6, CURR32, CURR43, RGB3			
Bit	Bit Name	Default	Access	Description	
7:0	adder_current3	00b	R/W	Selects the added current value – do not exceed together with currX_current the internal 8 Bit range (see text)	
				00h	0 (represents 0mA)
			
				FFh	255 (represents 38.25mA)

Table 61. Adder Current 4 Register

Addr: 52h		Adder Current 4			
		This register defines the current which can be added to CURR33			
Bit	Bit Name	Default	Access	Description	
7:0	adder_current4	00b	R/W	Selects the added current value – do not exceed together with currX_current the internal 8 Bit range (see text)	
				00h	0 (represents 0mA)
			
				FFh	255 (represents 38.25mA)

Table 62. Adder Enable 1 Register

Addr: 33h		Adder Enable 1			
		Enables the adder circuit for the selected current sources			
Bit	Bit Name	Default	Access	Description	
0	rgb1_adder	0	R/W	Enables adder circuit for current source RGB1	
				0	Normal Operation of the current source
				1	adder_current1 gets added to the current source current
1	rgb2_adder	0	R/W	Enables adder circuit for current source RGB2	
				0	Normal Operation of the current source
				1	adder_current2 gets added to the current source current

This filtered signal can be readout from the register `amb_result<7:0>`.

Each of the available three channels ($N=1, 2$ or 3) has six 8-bit registers:

- `groupN_y0`: define current multiplier for values below `groupN_X1`
- `groupN_y3`: define current multiplier for high values (actual starting point defined by `groupN_x1`, `groupN_k1` and `groupN_x2`, `groupN_k2`)
- `groupN_x1`, `groupN_k1`: If ADC reading is > `groupN_x1` then `groupN_k1` divided by 32 defines the slope of the first ramp
- `groupN_x2`, `groupN_k2`: If ADC reading is > `groupN_x2` then `groupN_k2` divided by 32 defines the slope of the second ramp

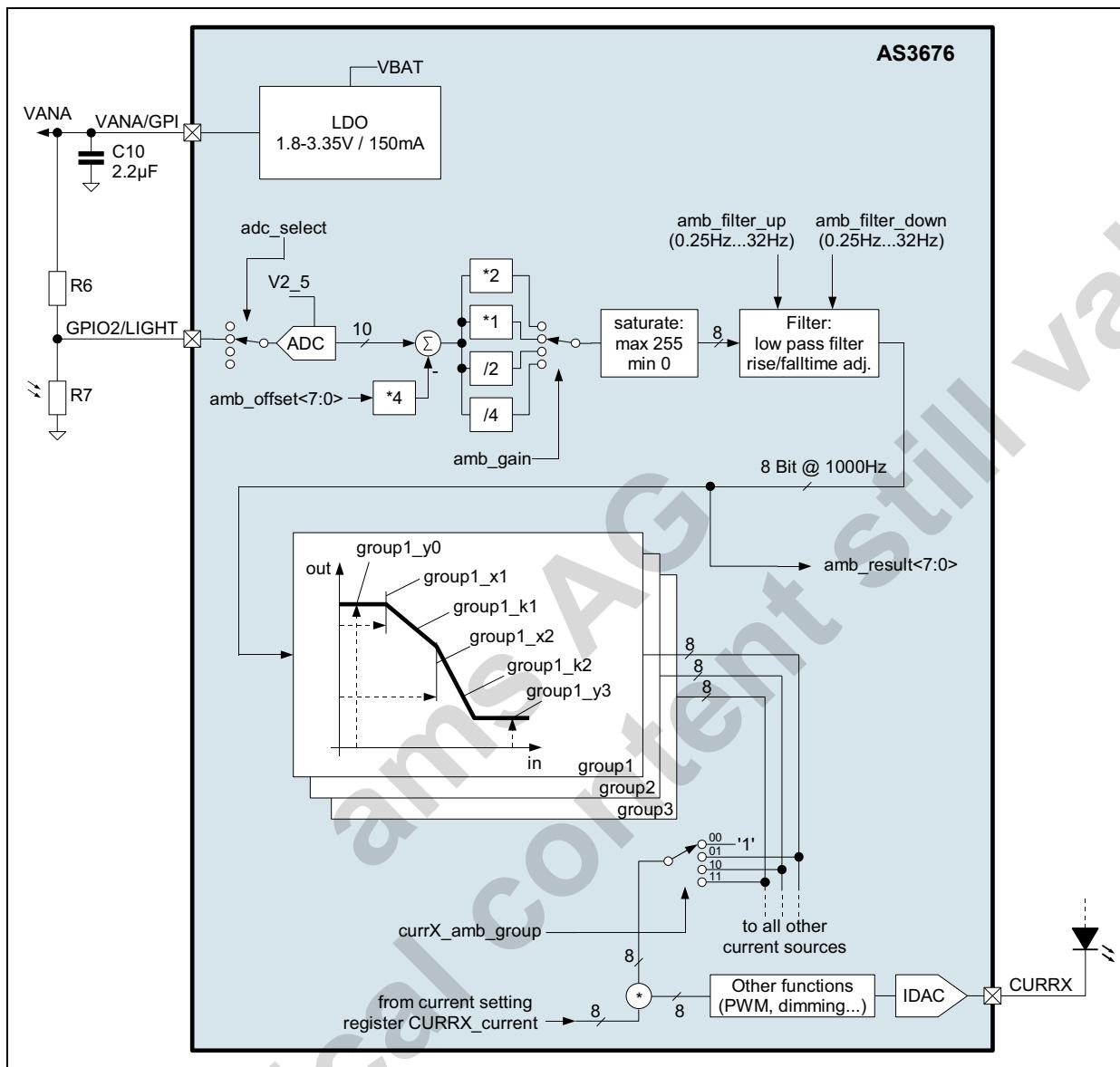
Each current sources has a 2 bit register (`currX_amb_group`) to select None, Group1, Group2 or Group3 of ambient light sensing.

The calculations are done every 1ms resulting in a flicker-free 1000Hz update rate of the current sources.

Note: A current source should not use adder or subtractor current at the same time together with ALS (e.g. `curr1_adder=1` and `curr1_amb_group=01, 10 or 11). For details see austriamicrosystems application note AN3676_ALS_with_adder_current.`

Technical content still valid

Figure 23. Ambient Light Sensor internal circuit



Ambient Light Sensor Registers

Table 65. *ALS control* Register

Addr: 90h		ALS control					
		control ambient light sensing					
Bit	Bit Name	Default	Access	Description			
0	amb_on	0	R/W	Enables the ambient light sensing feature			
				0	ambient light sensor disabled		
				1	ambient light sensor enabled		

Table 71. ALS curr3x group Register

Addr: 96h		ALS curr3x group			
		controls the group mapping for CURR30, CURR31, CURR32 and CURR33			
Bit	Bit Name	Default	Access	Description	
1:0	curr30_amb_group	00	R/W	CURR30 is mapped to ambient light sensor group	
				00	None - no ambient light sensor control
				01	Group 1
				10	Group 2
				11	Group 3
3:2	curr31_amb_group	00	R/W	CURR31 is mapped to ambient light sensor group	
				00	None - no ambient light sensor control
				01	Group 1
				10	Group 2
				11	Group 3
5:4	curr32_amb_group	00	R/W	CURR32 is mapped to ambient light sensor group	
				00	None - no ambient light sensor control
				01	Group 1
				10	Group 2
				11	Group 3
7:6	curr33_amb_group	00	R/W	CURR33 is mapped to ambient light sensor group	
				00	None - no ambient light sensor control
				01	Group 1
				10	Group 2
				11	Group 3

Table 72. ALS curr4x group Register

Addr: 97h		ALS curr4x group			
		controls the group mapping for CURR41, CURR42 and CURR43			
Bit	Bit Name	Default	Access	Description	
1:0	curr41_amb_group	00	R/W	CURR41 is mapped to ambient light sensor group	
				00	None - no ambient light sensor control
				01	Group 1
				10	Group 2
				11	Group 3
3:2	curr42_amb_group	00	R/W	CURR42 is mapped to ambient light sensor group	
				00	None - no ambient light sensor control
				01	Group 1
				10	Group 2
				11	Group 3

Table 87. ALS group 3 X1 Register

Addr: A6h		ALS group 3 X1			
Bit	Bit Name	Default	Access	Description	
7:0	group3_x1	00h	R/W	Group 3 x1 value	

Table 88. ALS group 3 K1 Register

Addr: A7h		ALS group 3 K1			
Bit	Bit Name	Default	Access	Description	
7:0	group3_k1	00h	R/W	Group 3 k1 value - divided by 32 defines first slope	

Table 89. ALS group 3 X2 Register

Addr: A8h		ALS group 3 X2			
Bit	Bit Name	Default	Access	Description	
7:0	group3_x2	00h	R/W	Group 3 x2 value	

Table 90. ALS group 3 K2 Register

Addr: A9h		ALS group 3 K2			
Bit	Bit Name	Default	Access	Description	
7:0	group3_k2	00h	R/W	Group 3 k2 value- value divided by 32 defines second slope	

8.4.9 DLS - Dynamic Luminance Scaling Input

The pin GPIO1/DLS can be used for dynamic backlight scaling input. Dynamic backlight scaling is used to reduce the power of the backlight especially when showing dark picture contents on the display. The control unit to operate DLS is the display processor sending a PWM signal to the AS3676 and in parallel changing the display content to compensate for a reduced brightness backlight.

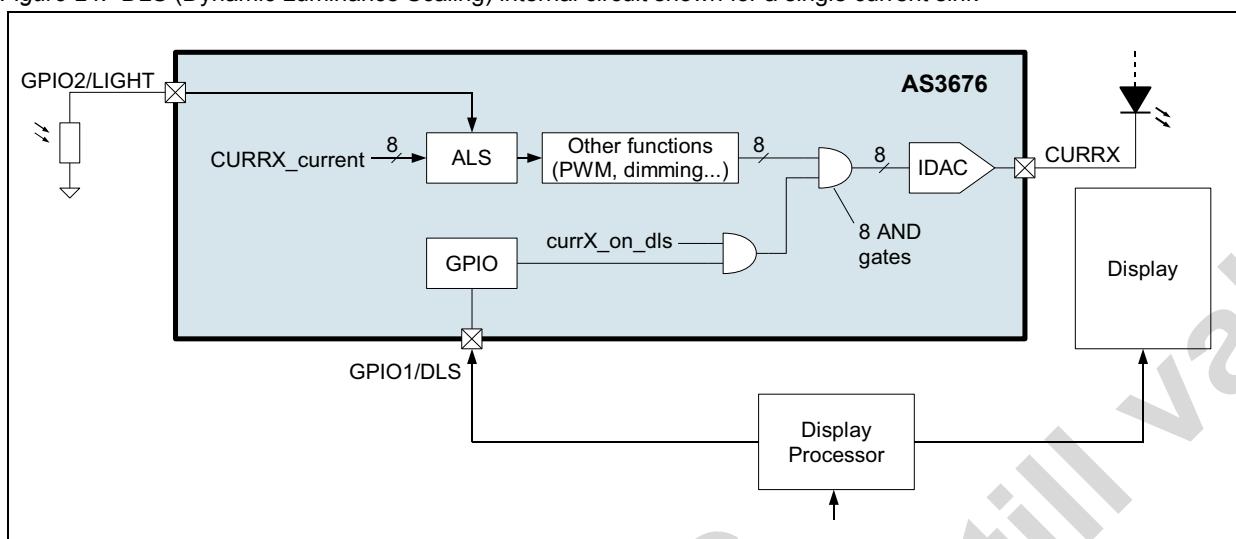
Table 91. DLS Input Parameters

Symbol	Parameter	Condition	Min	Typ	Max	Unit
fDLS	DLS input frequency range	pin GPIO1/DLS if used for DLS (any bit set in register DLS mode control1 or DLS mode control2)	25		1000	kHz

Note: If the input signal from the PWM (e.g. for DLS) is '0', the voltage on the current sink is captured with a S/H. Do not enable DLS operation using the DCDC step up converter if the input GPIO1/DLS is continuously at 0.

When this feature is enabled, the current sink current is disabled when the pin GPIO1/DLS is at 0. If GPIO1/DLS=1, the current sink operates at its programmed current as shown in [Figure 24](#):

Figure 24. DLS (Dynamic Luminance Scaling) internal circuit shown for a single current sink

Table 92. *DLS mode control1* Register

Addr: 56h		DLS mode control1			
		Setup which current sinks are connected to the DLS; if set to '1' the correspond current source (sink) is combined with the DLS input			
Bit	Bit Name	Default	Access	Description	
0	curr30_on_dls	0	R/W	0	CURR30 current sink is not combined with DLS
				1	CURR30 current sink is combined with DLS
1	curr31_on_dls	0	R/W	0	CURR31 current sink is not combined with DLS
				1	CURR31 current sink is combined with DLS
2	curr32_on_dls	0	R/W	0	CURR32 current sink is not combined with DLS
				1	CURR32 current sink is combined with DLS
3	curr33_on_dls	0	R/W	0	CURR33 current sink is not combined with DLS
				1	CURR33 current sink is combined with DLS
4	rgb1_on_dls	0	R/W	0	RGB1 current sink is not combined with DLS
				1	RGB1 current sink is combined with DLS
5	rgb2_on_dls	0	R/W	0	RGB2 current sink is not combined with DLS
				1	RGB2 current sink is combined with DLS
6	rgb3_on_dls	0	R/W	0	RGB3 current sink is not combined with DLS
				1	RGB3 current sink is combined with DLS

Table 93. *DLS mode control2* Register

Addr: 57h		DLS mode control2			
		Setup which current sinks are connected to the DLS; if set to '1' the correspond current source (sink) is combined with the DLS input			
Bit	Bit Name	Default	Access	Description	
0	curr1_on_dls	0	R/W	0	CURR1 current sink is not combined with DLS
				1	CURR1 current sink is combined with DLS

Table 93. *DLS mode control2* Register (Continued)

Addr: 57h		DLS mode control2			
		Setup which current sinks are connected to the DLS; if set to '1' the correspond current source (sink) is combined with the DLS input			
Bit	Bit Name	Default	Access	Description	
1	curr2_on_dls	0	R/W	0	CURR2 current sink is not combined with DLS
				1	CURR2 current sink is combined with DLS
2	curr41_on_dls	0	R/W	0	CURR41 current sink is not combined with DLS
				1	CURR41 current sink is combined with DLS
3	curr42_on_dls	0	R/W	0	CURR42 current sink is not combined with DLS
				1	CURR42 current sink is combined with DLS
4	curr43_on_dls	0	R/W	0	CURR43 current sink is not combined with DLS
				1	CURR43 current sink is combined with DLS
7	curr6_on_dls	0	R/W	0	CURR6 current sink is not combined with DLS
				1	CURR6 current sink is combined with DLS

8.5 General Purpose Input / Output

The GPIOs are a highly-configurable general purpose input/output pins which can be used for the following functionality:

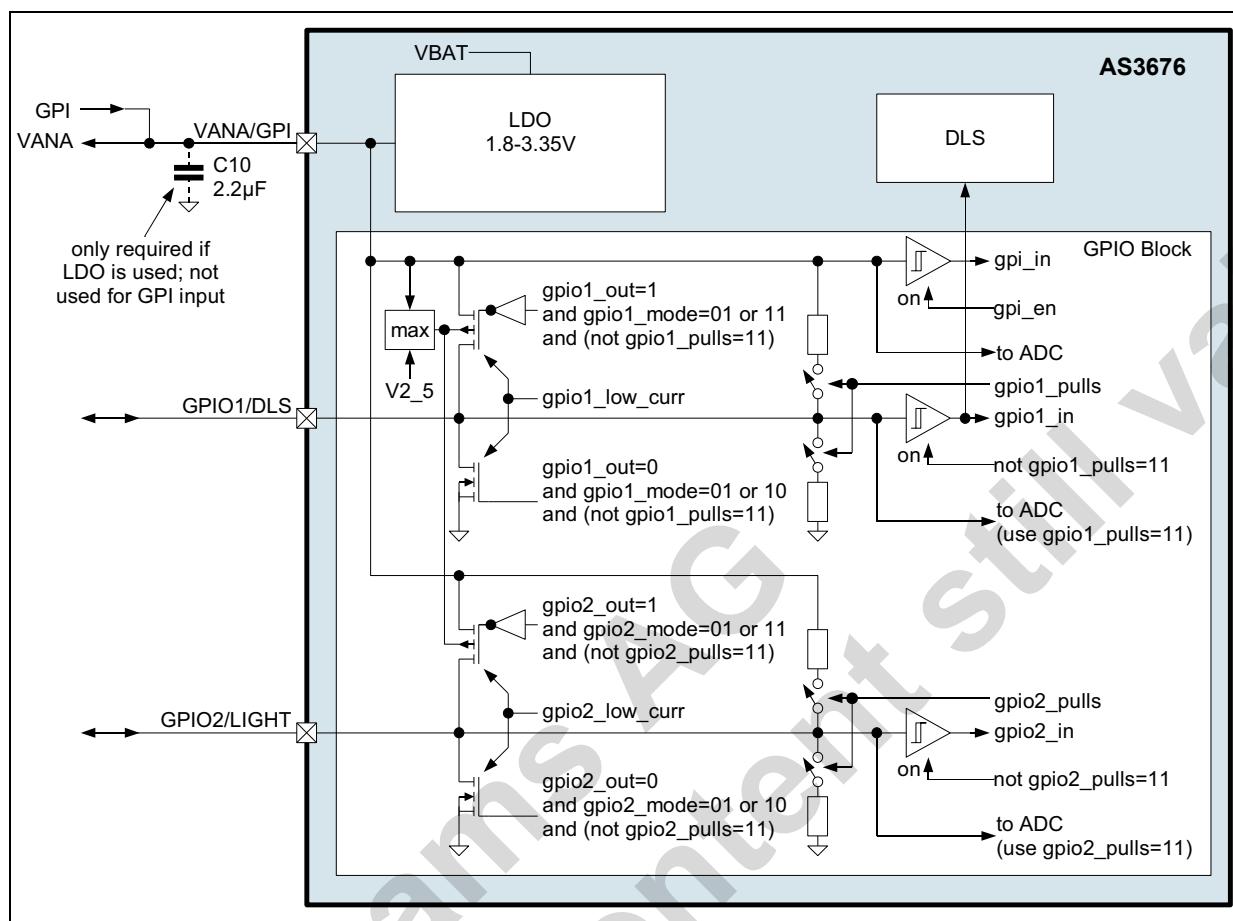
- Digital Schmitt Trigger Input
- Digital Output with 4mA Driving Capability at 2.8V Supply (VANA)
- Tristate Output
- Analog Input to the ADC
- Default Mode for GPIO1/DLS, GPIO2/LIGHT and VANA/GPI is Input (Pull-Down)

Table 94. GPIO Pin Function Summary

GPIO Pin	Configuration	Additional Function
GPIO1/DLS	Digital Input, Totem-Pole Output (Push/Pull), Open Drain (PMOS or NMOS), High-Z, Pull-Down or Pull-Up Resistor	ADC Input, PWM Input, DLS input (see page 55)
GPIO2/LIGHT	Digital Input, Totem-Pole Output (Push/Pull), Open Drain (PMOS or NMOS), High-Z, Pull-Down or Pull-Up Resistor	ADC Input, ALS - light sensor input (see page 47)
VANA/GPI	Digital Input ¹	ADC Input, LDO output

1. As VANA/GPI is used as a power supply for GPIO1/DLS, GPIO2/LIGHT, it is not recommended to use it as a digital input.

Figure 25. GPIOs and VANA/GPI Blockdiagram



8.5.1 Unused GPIO Pins

If the pins GPIO1/DLS or GPIO2/LIGHT are not used, they can be left open (an internal pulldown, which is enabled by default, will pull them to GND).

8.5.2 GPIO Characteristics

Table 95. GPIO DC Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Unit
Rpull	Pull up/Pull down Resistance	enabled by <code>gpio1_pulls</code> and <code>gpio2_pulls</code>	30		75	kΩ
VGPIO	Supply Voltage	=VANA/GPI	1.8		3.35	V
VIH	High Level Input Voltage	pins GPIO1/DLS and GPIO2/LIGHT	55% of VANA/GPI			V
VIL	Low Level Input Voltage				28% of VANA/GPI	V
VHYS	Hysteresis			5% of VANA/GPI		V
ILEAK	Input Leakage Current	To V2_5 or VANA/GPI and VSS	-5		5	µA
VOH	High Level Output Voltage	at Iout	0.8·VANA/GPI			V

Table 95. GPIO DC Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Unit
VOL	Low Level Output Voltage	at Iout			0.2·VANA/GPI	V
IOUT	Driving Capability	VANA/GPI = 2.8V, gpio1_low_curr or gpio2_low_curr= 1	4			mA
		VANA/GPI = 2.8V, gpio1_low_curr or gpio2_low_curr = 0	16			
CLOAD	Capacitive Load				50	pF

8.5.3 GPIO Registers

Table 96. GPIO output 1 Register

Addr: 05h		GPIO output 1				
		This register controls GPIO outputs.				
Bit	Bit Name	Default	Access	Description		
0	gpi_curr1_en	0	R/W	Enables the CURR1 input		
				0	input disabled	
				1	input enabled	
1	gpi_curr2_en	0	R/W	Enables the CURR2 input		
				0	input disabled	
				1	input enabled	
2	gpi_curr6_en	0	R/W	Enables the CURR6 input		
				0	input disabled	
				1	input enabled	
3	gpi_en	0	R/W	Enables the VANA/GPI input		
				0	input disabled	
				1	input enabled	
4	gpi_curr30_en	0	R/W	Enables the CURR30 input		
				0	input disabled	
				1	input enabled	
5	gpi_curr31_en	0	R/W	Enables the CURR31 input		
				0	input disabled	
				1	input enabled	
6	gpi_curr32_en	0	R/W	Enables the CURR32 input		
				0	input disabled	
				1	input enabled	
7	gpi_curr33_en	0	R/W	Enables the CURR33 input		
				0	input disabled	
				1	input enabled	

Table 97. GPIO signal 1 Register

Addr: 06h		GPIO signal 1			
		This register controls GPIO outputs.			
Bit	Bit Name	Default	Access	Description	
0	gpi_curr1_in	N/A	R	Reads a logic signal from pin CURR1; if gpi_curr1_en=1	
1	gpi_curr2_in	N/A	R	Reads a logic signal from pin CURR2; if gpi_curr2_en=1	
2	gpi_curr6_in	N/A	R	Reads a logic signal from pin CURR6; if gpi_curr6_en=1	
3	gpi_in	N/A	R	Reads a logic signal from pin CURR6; if gpi_en=1	
4	gpi_curr30_in	N/A	R	Reads a logic signal from pin CURR30; if gpi_curr30_en=1	
5	gpi_curr31_in	N/A	R	Reads a logic signal from pin CURR31; if gpi_curr31_en=1	
6	gpi_curr32_in	N/A	R	Reads a logic signal from pin CURR32; if gpi_curr32_en=1	
7	gpi_curr33_in	N/A	R	Reads a logic signal from pin CURR33; if gpi_curr33_en=1	

Table 98. GPIO output 2 Register

Addr: 50h		GPIO output 2			
		This register controls GPIO outputs.			
Bit	Bit Name	Default	Access	Description	
0	gpio1_out	0	R/W	Writes a logic signal to pin GPIO1/DLS; this is independent of any other bit setting e.g., gpio1_mode Table 100 .	
1	gpio2_out	0	R/W	Writes a logic signal to pin GPIO1/DLS; this is independent of any other bit setting e.g., gpio2_mode Table 100	
2	gpi_rgb1_en	0	R/W	Enables the RGB1 input	
				0	input disabled
				1	input enabled
3	gpi_rgb2_en	0	R/W	Enables the RGB2 input	
				0	input disabled
				1	input enabled
4	gpi_rgb3_en	0	R/W	Enables the RGB3 input	
				0	input disabled
				1	input enabled
5	gpi_curr41_en	0	R/W	Enables the CURR41 input	
				0	input disabled
				1	input enabled
6	gpi_curr42_en	0	R/W	Enables the CURR42 input	
				0	input disabled
				1	input enabled
7	gpi_curr43_en	0	R/W	Enables the CURR43 input	
				0	input disabled
				1	input enabled

Table 99. *GPIO signal 2 Register*

Addr: 51h		GPIO signal 2			
		This register controls GPIO outputs.			
Bit	Bit Name	Default	Access	Description	
0	gpio1_in	N/A	R	Reads a logic signal from pin GPIO1/DLS; this is independent of any other setting e.g., Table 100 except gpio1_pulls=11	
1	gpio2_in	N/A	R	Reads a logic signal from pin GPIO2/LIGHT; this is independent of any other setting e.g., Table 100 except gpio2_pulls=11	
2	gpi_rgb1_in	N/A	R	Reads a logic signal from pin RGB1; if gpi_rgb1_en=1	
3	gpi_rgb2_in	N/A	R	Reads a logic signal from pin RGB2; if gpi_rgb2_en=1	
4	gpi_rgb3_in	N/A	R	Reads a logic signal from pin RGB3; if gpi_rgb3_en=1	
5	gpi_curr41_in	N/A	R	Reads a logic signal from pin CURR41; if gpi_curr41_en=1	
6	gpi_curr42_in	N/A	R	Reads a logic signal from pin CURR42; if gpi_curr42_en=1	
7	gpi_curr43_in	N/A	R	Reads a logic signal from pin CURR43; if gpi_curr43_en=1	

Table 100. *GPIO control Register*

Addr: 1Eh		GPIO control			
		This register controls GPIO and GPIO1 pin functions.			
Bit	Bit Name	Default	Access	Description	
1:0	gpio1_mode	00	R/W	Defines the direction for pin GPIO1/DLS	
				00	Input only
				01	Output (push and pull)
				10	Output (open drain, only push; only NMOS is active)
				11	Output (open drain, only pull; only PMOS is active)
3:2	gpio1_pulls	01	R/W	Adds the following pullup/pulldown to pin GPIO1/DLS; this is independent of setting of bits gpio1_mode	
				00	None
				01	Pulldown
				10	Pullup
				11	ADC input (gpio1_mode = XX); recommended for analog signals
5:4	gpio2_mode	00	R/W	Defines the direction for pin GPIO2/LIGHT	
				00	Input only
				01	Output (push and pull)
				10	Output (open drain, only push; only NMOS is active)
				11	Output (open drain, only pull; only PMOS is active)

Table 100. *GPIO control* Register (Continued)

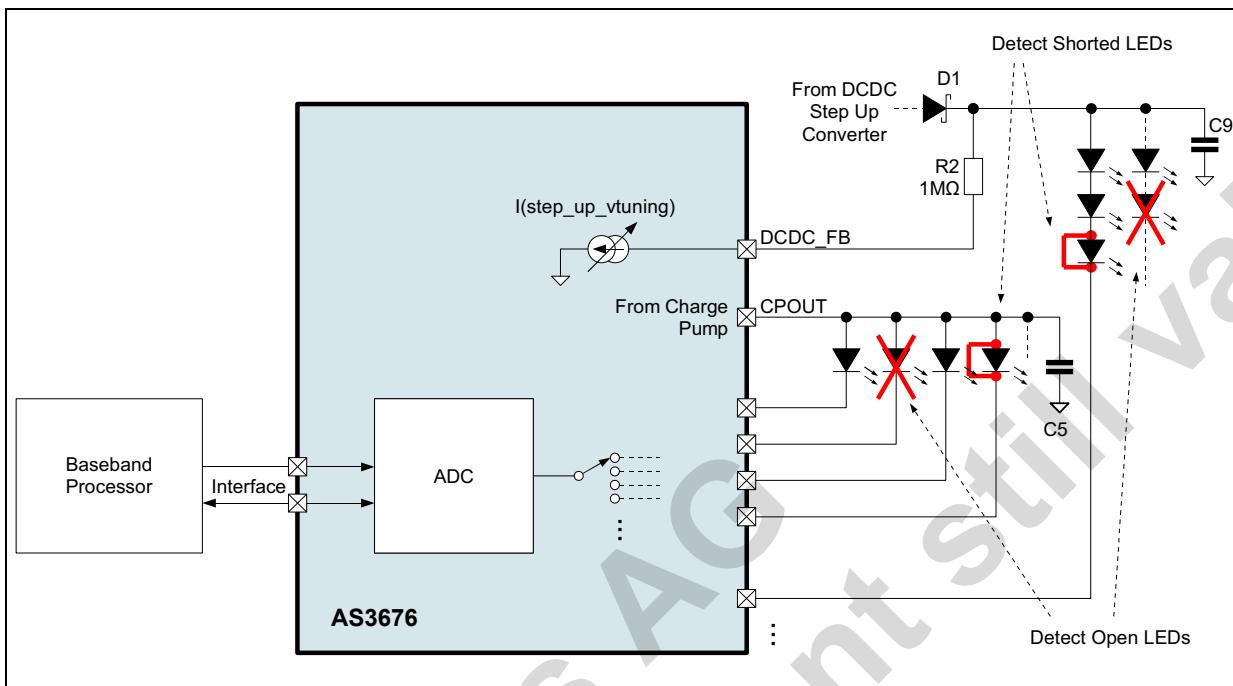
Addr: 1Eh		GPIO control			
		This register controls GPIO and GPIO1 pin functions.			
Bit	Bit Name	Default	Access	Description	
7:6	gpio2_pulls	01	R/W	Adds the following pullup/pulldown to pin GPIO2/LIGHT; this is independent of setting of bits <code>gpio2_mode</code>	
				00	None
				01	Pulldown
				10	Pullup
				11	ADC input (<code>gpio2_mode</code> = XX); recommended for analog signals

Table 101. *GPIO driving cap* Register

Addr: 20h		GPIO driving cap			
		This register enables low current mode for GPIOs.			
Bit	Bit Name	Default	Access	Description	
0	gpio1_low_curr	0	R/W	Defines the driving capability of pin GPIO1/DLS	
				0	Iout
				1	Iout /4
1	gpio2_low_curr	0	R/W	Defines the driving capability of pin GPIO2/LIGHT	
				0	Iout
				1	Iout /4

8.6 LED Test

Figure 26. LED Function Testing



The AS3676 supports the verification of the functionality of all the connected LEDs (open and shorted LEDs can be detected). This feature is especially useful in production test to verify the correct assembly of the LEDs, all its connectors and cables. It can also be used in the field to verify if any of the LEDs is damaged. A damaged LED can then be disabled (to avoid unnecessary currents).

The current sources, charge pump, dc/dc converter and the internal ADC are used to verify the forward voltage of the LEDs. If this forward voltage is within the specified limits of the LEDs, the external circuitry is assumed to operate.

8.6.1 Function Testing for single LEDs connected to the Charge Pump

For any current source connected to the charge pump (CURR30-33) where only one LED is connected between the charge pump and the current sink (see Figure 1) use:

Table 102. Function Testing for LEDs connected to the Charge Pump

Step	Action	Example Code
1	Switch on the charge pump and set it into manual 1:2 mode (to avoid automatic mode switching during measurements)	Reg 23h ≤ 14h (<code>cp_mode = 1:2, manual</code>) Reg 00h ≤ 04h (<code>cp_on = 1</code>)
2	Switch on the current sink for the LED to be tested	e.g. for register CURR31 set to 9mA use Reg 10h ≤ 0Fh (<code>curr3x_other = 9mA</code>) Reg 03h ≤ 0ch (<code>curr31_mode = curr31_other</code>)
3	Measure with the ADC the voltage on CPOUT	Reg 26h ≤ 95h (<code>adc_select=CPOUT,start ADC</code>) Fetch the ADC result from Reg 27h and 28h
4	Measure with the ADC the voltage on the switched on current sink	Reg 26h ≤ 8bh (<code>adc_select=CURR31,start ADC</code>) Fetch the ADC result from Reg 27h and 28h
5	Switch off the current sink for the LED to be tested	Reg 03h ≤ 00h (<code>curr31_mode = off</code>)
6	Compare the difference between the ADC measurements (which is the actual voltage across the tested LED) against the specification limits of the tested LED	Calculation performed in baseband uProcessor

Table 102. Function Testing for LEDs connected to the Charge Pump

Step	Action	Example Code
7	Do the same procedure for the next LED starting from point 2	Jump to 2. If not all the LEDs have been tested
8	Switch off the charge pump set charge pump automatic mode	Reg 00h ≤ 00h (<code>cp_on</code> = 0) Reg 23h ≤ 00h

8.6.2 Function Testing for LEDs connected to the Step Up DCDC Converter

For LEDs connected to the DCDC converter (usually current sinks CURR1,CURR2 and CURR6) use the following procedure:

Table 103. Function Testing for LEDs connected to the DCDC converter

Step	Action	Example Code
1	Switch on the current sink for the LED string to be tested (CURR1,2 or 6)	e.g. Test LEDs on CURR1: Reg 01h ≤ 01h (<code>curr1_mode</code> =on) Reg 09h ≤ 3ch (<code>curr1_current</code> = 9mA)
2	Select the feedback path for the LED string to be tested (e.g. <code>step_up_fb</code> = 01 for LED string on CURR1)	Reg 21h ≤ 02h (<code>step_up_fb</code> =curr1)
3	Set the current for <code>step_up_vtuning</code> exactly above the maximum forward voltage of the tested LED string + 0.6V (for the current sink) + 0.25V; add 6% margin (accuracy of <code>step_up_vtuning</code>); this sets the maximum output voltage limit for the DCDC converter	e.g. 4 LEDs with $U_{fMAX} = 4.1V$ gives $17.25V + 6\% = 18.29V$; if $R2=1M\Omega$ and $R3$ = open, then select <code>step_up_vtuning</code> = 18 (Reg 21h ≤ 92h; results in 19.25V over voltage protection voltage – Table 9 on page 14)
4	Set <code>step_up_prot</code> = 1	Reg 22h ≤ 04h
5	Switch on the DCDC converter	Reg 00h ≤ 08h
6	Wait 80ms (DCDC_FB settling time)	
7	Measure the voltage on DCDC_FB (ADC)	Reg 26h ≤ 96h (<code>adc_select</code> =DCDC_FB, start ADC; Fetch the ADC result from Reg 27h and 28h)
8	If the voltage on DCDC_FB is above 1.0V, the tested LED string is broken – then skip the following steps	(Code >199h)
9	Switch off the over voltage protection (<code>step_up_prot</code> =0)	Reg 22h ≤ 00h
10	Reduce <code>step_up_vtuning</code> step by step until the measured voltage on DCDC_FB (ADC) is above 1.0V. After changing <code>step_up_vtuning</code> always wait 80ms, before AD-conversion	e.g.: Reg 21h ≤ 62h (<code>step_up_vtuning</code> =12): ADC result=1,602V
11	Measure voltage on DCDC_FB	e.g. DCDC_FB=1.602V
12	Switch off the DCDC converter	Reg 00h ≤ 00h
13	The voltage on the LED string can be calculated now as follows (R4 = open): $V_{LEDSTRING} = V(DCDC_FB) + I(\text{step_up_vtuning}) * R2 - 0.5V$ (current sinks feedback voltage: VFB2). $V(DCDC_FB) = \text{ADC Measurement from point 11}$ $I(\text{step_up_vtuning}) = \text{last setting used for point 10}$	e.g.: $V_{LED} = (1.602V + 12V - 0.5V) / 4 = 3.276V$
14	Compare the calculated value against the specification limits of the tested LEDs	

Note: With the above described procedures electrically open and shorted LEDs can be automatically detected

8.7 Analog-to-Digital Converter

The AS3676 has a built-in 10-bit successive approximation analog-to-digital converter (ADC). It is internally supplied by V_{2_5}, which is also the full-scale input range (0V defines the ADC zero-code). For input signals exceeding V_{2_5} (typ. 2.5V) a resistor divider with a gain of 0.4 (Ratioprescaler) is used to scale the input of the ADC converter. Consequently the resolution is:

Table 104. ADC Input Ranges, Compliances and Resolution

Channels (Pins)	Input Range	V _{LSB}	Note
DCDC_FB, GPIO1/DLS, GPIO2/LIGHT, VANA/GPI, audio controlled LED buffer output	0V-2.5V	2.44mV	V _{LSB} =2.5/1024
ADCTEMP_CODE	-30°C to 125°C	1 / ADC _{Tc}	junction temperature
CURR3x, CURR4x, RGBx VBAT, CPOUT	0V-5.5V	6.1mV	V _{LSB} =(2.5/1024)/0.4; internal resistor divider used
CURR1, CURR2, CURR6	0V-1.0V	2.44mV	V _{LSB} =2.5/1024

Table 105. ADC Parameters

Symbol	Parameter	Condition	Min	Typ	Max	Unit
	Resolution		10			Bit
V _{IN}	Input Voltage Range	V _{SUPPLY} = V _{2_5}	V _{SS}		see Table 104	V
DNL	Differential Non-Linearity			± 0.25		LSB
INL	Integral Non-Linearity			± 0.5		LSB
V _{os}	Input Offset Voltage			± 0.25		LSB
R _{in}	Input Impedance		100			MΩ
C _{in}	Input Capacitance				9	pF
V _{SUPPLY} (V _{2_5})	Power Supply Range	± 2%, internally trimmed.		2.5		V
I _{dd}	Power Supply Current	During conversion only.		300		μA
I _{dd}	Power Down Current			100		nA
T _{TOL}	Temperature Sensor Accuracy	@ 25 °C	-10		+10	°C
ADC _{OFFSET}	ADC temperature measurement offset value			375		°C
ADC _{Tc}	Code temperature coefficient	Temperature change per ADC LSB		1.293 9		°C/ Code
RatioPRESCALE_R	Ratio of Prescaler	For all low voltage current sinks, CPOUT and VBAT		0.4		
Transient Parameters (2.5V, 25 °C)						
T _c	Conversion Time	All signals are internally generated and triggered by start_conversion		27		μs
f _c	Clock Frequency			1.0		MHz
t _s	Settling Time of S&H			16		μs

The junction temperature (T_{JUNCTION}) can be calculated with the following formula (ADCTEMP_CODE is the adc conversion result for channel 17h selected by register `adc_select` = 010111b):

$$T_{JUNCTION} [^{\circ}\text{C}] = \text{ADC}_{OFFSET} - \text{ADC}_{Tc} \cdot \text{ADCTEMP_CODE} \quad (\text{EQ } 5)$$

ADC Registers

Table 106. *ADC_MSB result* Register

Addr: 27h		ADC_MSB result			
Together with Register 27h, this register contains the results (MSB) of an ADC cycle.					
Bit	Bit Name	Default	Access	Description	
6:0	D9:D3	N/A	R	ADC results register.	
7	result_not_ready	N/A	R	Indicates end of ADC conversion cycle	
				0	Result is ready
				1	Conversion is running

Table 107. *ADC_LSB result* Register

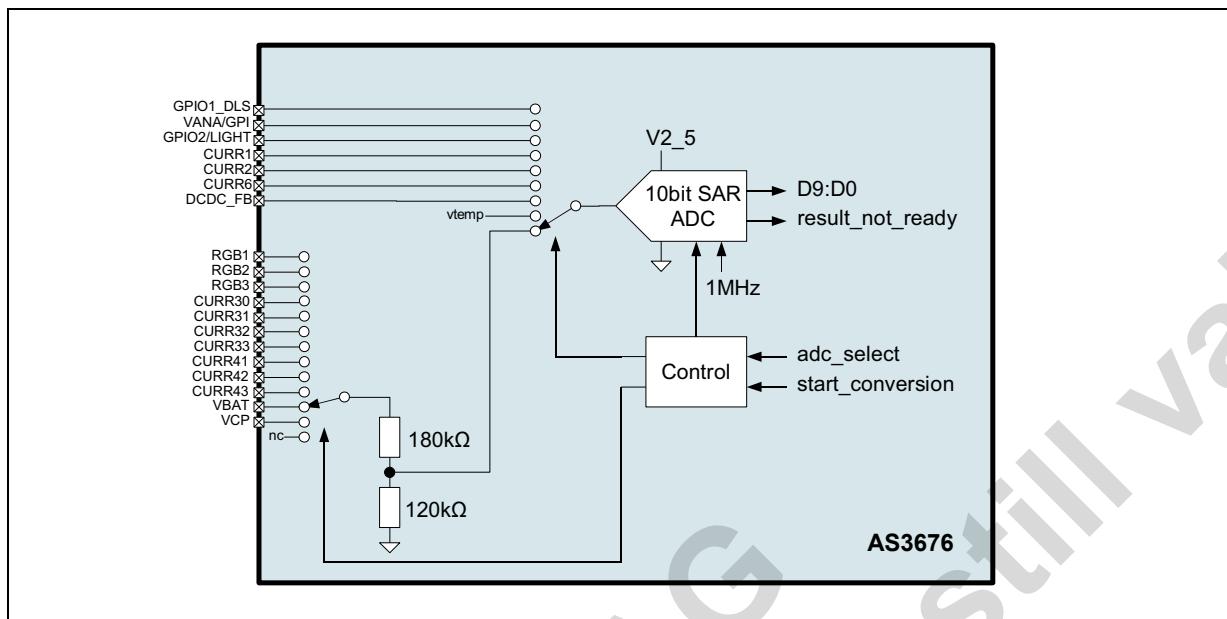
Addr: 28h		ADC_LSB result			
Together with Register 28h, this register contains the results (LSB) of an ADC cycle					
Bit	Bit Name	Default	Access	Description	
2:0	D2:D0	N/A	R	ADC result register	

Table 108. ADC_control Register

Addr: 26h		ADC_control		
		This register input source selection and initialization of ADC		
Bit	Bit Name	Default	Access	Description
5:0	adc_select ¹	03h	R/W	Selects input source as ADC input
				000000 (00h) GPIO2/LIGHT
				000001 (01h) VANA/GPI
				000010 (02h) GPIO1/DLS
				000011 (03h) audio controlled LED buffer output
				000100 (04h) reserved
				000101 (05h) RGB1
				000110 (06h) RGB2
				000111 (07h) RGB3
				001000 (08h) CURR1
				001001 (09h) CURR2
				001010 (0Ah) CURR30
				001011 (0Bh) CURR31
				001100 (0Ch) CURR32
				001101 (0Dh) CURR33
				001110 (0Eh) CURR41
				001111 (0Fh) CURR42
				010000 (10h) CURR43
				010001 (11h) reserved
				010010 (12h) reserved
				010011 (13h) CURR6
				010100 (14h) VBAT
				010101 (15h) CPOUT
				010110 (16h) DCDC_FB
				010111 (17h) ADCTEMP_CODE (junction temperature)
				011xxx, 1xxxxx reserved
6				NA
7	start_conversion	N/A	W	Writing a 1 into this bit starts one ADC conversion cycle.

1. See Table Table 104 for ADC ranges and resolution.

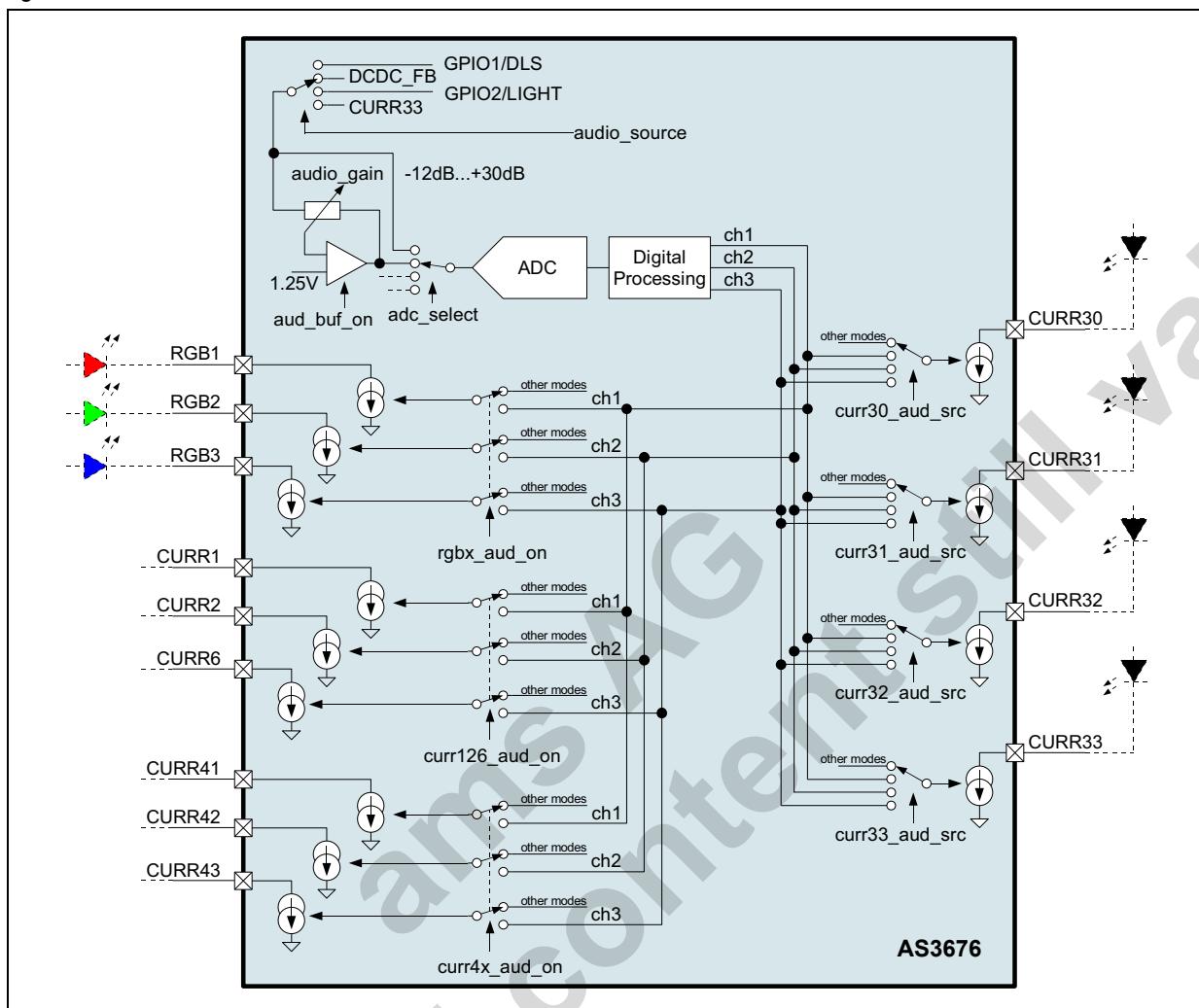
Figure 27. ADC Circuit



8.8 Audio controlled LEDs

Up to four RGB LEDs and/or up to 13 LEDs (number of LEDs is fully configurable) can be controlled by an audio source (connected to the pin CURR33, DCDC_FB or GPIO2/LIGHT). The color of the RGB LED(s) or the brightness of the single color LED(s) is depending on the input amplitude. For the RGB LEDs it starts from black transitions to blue, green, cyan, yellow, red and for high amplitudes white is used (internal lookup table if `audio_color=000b`).

Figure 28. Audio controlled LED internal circuit



The audio controlled LED block is enabled if any of the registers [curr30_aud_src\[1:0\]...curr33_aud_src\[1:0\]](#), [curr126_aud_on](#), [rgbx_aud_on](#) or [curr4x_aud_on](#) not equal zero.

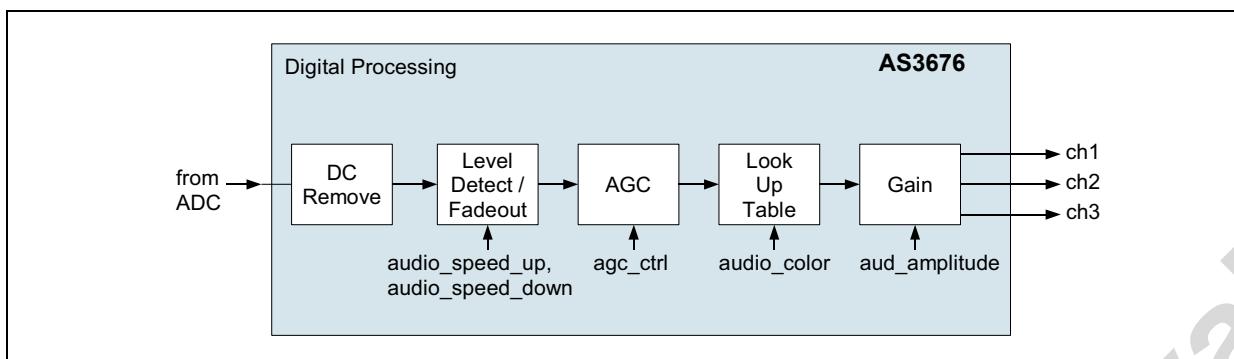
The audio input amplifier (enabled by [aud_buf_on=1](#)) is used to allow the attenuation (or amplification of the input signal) and has the following parameters:

Table 109. Audio input Parameters

Symbol	Parameter	Condition	Min	Typ	Max	Unit
VIN	Input Voltage Range		0		2.5	V
Rin_min	min. Input Impedance	at max. input gain (30dB)		20		kΩ

The signal is converted with the ADC (If the audio controlled LED is active, the internal ADC is continuously running. In this case the ADC cannot be used for any other purpose). The digital processing converts this signal into 3 channels (ch1, ch2, ch3):

Figure 29. Audio controlled LED digital processing internal circuit



These three output channels (ch1, ch2, ch3) can be routed to any of the current sources according to Figure 28.

The input amplitude is mapped into different colors (Look Up Table in Figure 29) for RGB LED(s) or brightness for single color LED(s). The mapping is controlled by the register `audio_color`. If `audio_color` = 000, then the mapping is done as follows:

Very low amplitudes are mapped to black, for higher amplitudes, the color smoothly transitions from blue, green, cyan, yellow, red and eventually to white (for high input amplitudes).

8.8.1 AGC

The AGC is used to ‘compress’ the input signal and to attenuate very low input amplitude signals (this is performed to ensure no light output for low signals especially for noisy input signals).

The AGC monitors the input signal amplitude and filters this amplitude with a filter with a short attack time, but a long decay time (decay time depends on the register `agc_ctrl`). This amplitude measurement (represented by an integer value from 0 to 15) is then used to amplify or attenuate the input signal with one of the following amplification ratios (output to input ratio) – the curve A, B, or C is selected depending on the register `agc_ctrl`:

Figure 30. AGC curve A (x-axis: input amplitude, y-axis: output amplitude; actual value: gain between output to input)

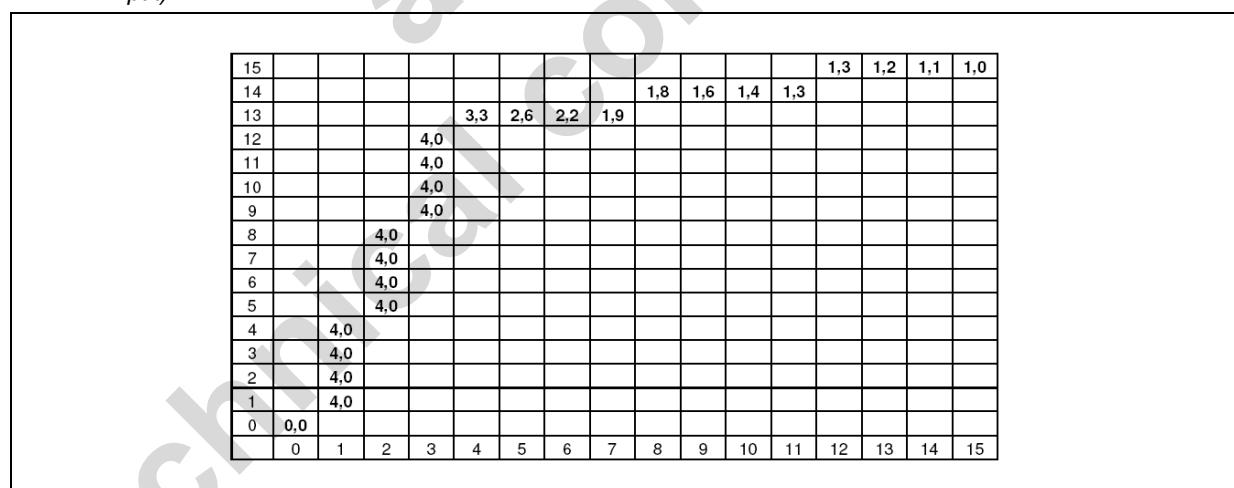


Figure 31. AGC curve B (x-axis: input amplitude, y-axis: output amplitude; actual value: gain between output to input)

15																	1,0	
14																		
13																		
12								2,0	1,7	1,5						1,2	1,1	1,0
11									2,2									
10										2,5								
9										3,0								
8										3,0								
7										3,0								
6										3,0								
5										3,0								
4										3,0								
3										3,0								
2										3,0								
1										3,0								
0	0,0																	
	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15		

Figure 32. AGC curve C (x-axis: input amplitude, y-axis: output amplitude; actual value: gain between output to input)

15																	1,0
14																	
13																	
12																	
11																	
10																	
9																	
8																	
7																	
6																	
5																	
4																	
3																	
2																	
1																	
0	0,0																
	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	

8.8.2 Audio Controlled LED Registers

Table 110. *Audio Control* Register

Addr: 46h		Audio Control				
		Audio Sync Mode control				
Bit	Bit Name	Default	Access	Description		
0	aud_buf_on	0b	R/W	Audio input buffer enable		
				0	off; for audio direct input to ADC use <i>adc_select</i> = 00h (AUDIO_IN)	
				1	on; set <i>adc_select</i> = 03h (buffer output)	
4:2	audio_color	000b	R/W	audio controlled LED color selection (amplitude mode)		
				000	color scheme defined by lookup table	
				001-111	single color scheme (b2=R, b1=G, b0=B)	

Table 110. *Audio Control* Register (Continued)

Addr: 46h		Audio Control			
		Audio Sync Mode control			
Bit	Bit Name	Default	Access	Description	
7:6	audio_speed_down	00b	R/W	Audio controlled LED persistence time (ramping down)	
				00	none
				01	200ms
				10	400ms
				11	800ms

Table 111. *Audio input* Register

Addr: 47h		Audio input			
		Audio Sync input control			
Bit	Bit Name	Default	Access	Description	
2:0	audio_gain	000b	R/W	Audio input buffer gain control	
				000	-12dB
				001	-6dB
				010	0dB
				011	+6dB
				100	+12dB
				101	+18dB
				110	+24dB
				111	+30dB
				Audio input buffer AGC function controls AGC switching threshold	
5:3	agc_ctrl	000b	R/W	000	AGC off
				001	Attenuate low amplitude signals otherwise linear response (to remove e.g. noise)
				010	AGC curve A; slow decay of amplitude detection
				011	AGC curve A; fast decay of amplitude detection
				100	AGC curve B; slow decay of amplitude detection
				101	AGC curve B; fast decay of amplitude detection
				110	AGC curve C; slow decay of amplitude detection
				111	AGC curve C; fast decay of amplitude detection
				Startup Control of audio input buffer (used to charge optional external dc blocking capacitor)	
				0	automatic precharging 300us (if <code>audio_dis_start</code> = 0)
6	audio_man_start ¹	0b	R/W	1	continuously precharging (if <code>aud_buf_on</code> = 1)
				Disable Startup Control of audio input buffer (used to charge optional external dc blocking capacitor)	
				0	precharging enabled
7	audio_dis_start ²	0b	R/W	1	precharging disabled

1. Its safe to keep default value
2. Its safe to keep default value

Table 112. *Audio output Register*

Addr: 48h		Audio output			
		Audio Sync input control			
Bit	Bit Name	Default	Access	Description	
2:0	aud_amplitude	000b	R/W	LED(s) output amplitude control (in percent of selected output current)	
				000	6.25%
				001	12.5%
				010	25%
				011	50%
				100	75%
				101	87.5%
				110	93.75%
				111	100%
4	curr126_aud_on	0b	R/W	Audio controlled LED enable for CURR1, CURR2, CURR6	
				0	off
				1	on, audio controlled LED is enabled
5	rgbx_aud_on	0b	R/W	Audio controlled LED enable for RGB1-RGB3	
				0	off
				1	on, audio controlled LED is enabled
6	curr4x_aud_on	0b	R/W	Audio controlled LED enable for CURR41-CURR43	
				0	off
				1	on, audio controlled LED is enabled

Table 113. *CURR3x audio source Register*

Addr: 53h		CURR3x audio source			
		Controls CURR30,31,32,33 audio outputs and enables audio controlled LED			
Bit	Bit Name	Default	Access	Description	
1:0	curr30_aud_src[1:0]	00b	R/W	Audio controlled LED source for CURR30	
				00	All other modes
				01	ch1 connected to CURR30, audio controlled LED on
				10	ch2 connected to CURR30, audio controlled LED on
				11	ch3 connected to CURR30, audio controlled LED on

Table 113. CURR3x audio source Register (Continued)

Addr: 53h		CURR3x audio source			
		Controls CURR30,31,32,33 audio outputs and enables audio controlled LED			
Bit	Bit Name	Default	Access	Description	
3:2	curr31_aud_src[1:0]	00b	R/W	Audio controlled LED source for CURR31	
				00	All other modes
				01	ch1 connected to CURR31, audio controlled LED on
				10	ch2 connected to CURR31, audio controlled LED on
				11	ch3 connected to CURR32, audio controlled LED on
5:4	curr32_aud_src[1:0]	00b	R/W	Audio controlled LED source for CURR32	
				00	All other modes
				01	ch1 connected to CURR32, audio controlled LED on
				10	ch2 connected to CURR32, audio controlled LED on
				11	ch3 connected to CURR32, audio controlled LED on
7:6	curr33_aud_src[1:0]	00b	R/W	Audio controlled LED source for CURR33	
				00	All other modes
				01	ch1 connected to CURR33, audio controlled LED on
				10	ch2 connected to CURR33, audio controlled LED on
				11	ch3 connected to CURR33, audio controlled LED on

Table 114. Audio Control 2 Register

Addr: 55h		Audio Control 2			
		Audio Mode Control Register			
Bit	Bit Name	Default	Access	Description	
0		NA		not used	
3:1	audio_speed_up	000b	R/W	Audio controlled LED filtering time (ramping up)	
				000	none
				001	50ms
				010	100ms
				011	150ms
				100	200ms
				101	250ms
				110	400ms
				111	800ms

Table 114. *Audio Control 2 Register (Continued)*

Addr: 55h		Audio Control 2				
		Audio Mode Control Register				
Bit	Bit Name	Default	Access	Description		
5:4	audio_source	00b	R/W	Audio Buffer input source		
				00	CURRE33	
				01	DCDC_FB	
				10	GPIO1/DLS	
				11	GPIO2/LIGHT	

8.9 Power-On Reset

The internal reset is controlled by two sources:

- VBAT Supply
- Serial interface state (CLK, DATA)

The internal reset is forced if VBAT is low or if both interface pins (CLK, DATA) are low for more than tPOR_DEB (typ. 100ms)⁶. Then device enters shutdown mode. For details see section [Operating Modes](#) on page 80.

The reset levels control the state of all registers. As long as VBAT and CLK/DATA are below their reset thresholds, the register contents are set to default. Access by serial interface is possible once the reset thresholds are exceeded.

Figure 33. Zero Power Device Wakeup block diagram

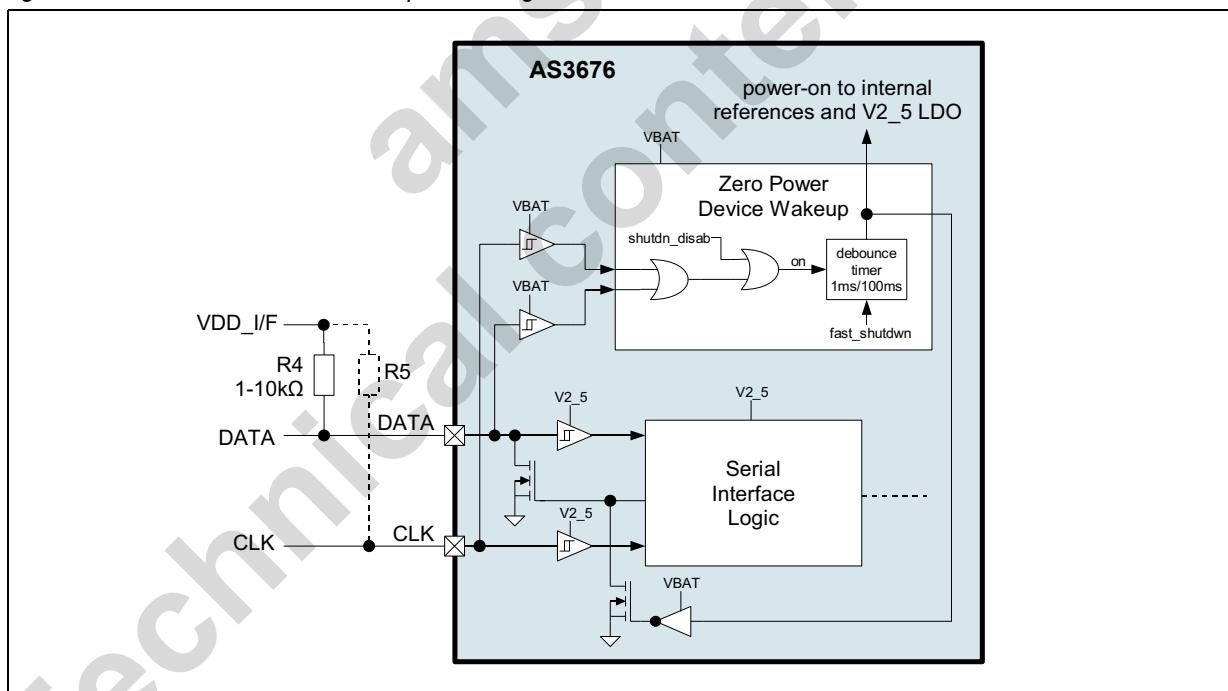


Table 115. Power On Reset Parameters

Symbol	Parameter	Condition	Min	Typ	Max	Unit
VPOR_VBAT	Overall Power-On Reset	Monitor voltage on V2_5; power-on reset for all internal functions.	1.8	2.15	2.4 ¹	V

6. Only if shutdn_enab=1

Table 115. Power On Reset Parameters

Symbol	Parameter	Condition	Min	Typ	Max	Unit
VPOR_PERI	Reset Level for pins CLK, DATA	Monitor voltage on pins CLK, DATA	0.29	1.0	1.38	V
tPOR_DEB	Reset debounce time for pins CLK, DATA		80	100	120	ms
tstart	Interface Startup Time		4	6	8	ms

1. Guaranteed by design - min./max. limits not production tested

8.9.1 Reset control register

Table 116. *Reset and Overtemp control* Register

Addr: 29h		Reset and Overtemp control				
		This register reads and resets the overtemperature flag.				
Bit	Bit Name	Default	Access	Description		
4	shutdown_enab	0	R/W	Enable Shutdown mode and serial interface reset.		
				0	Serial Interface reset disabled. Device does not enter Shutdown mode	
				1	Serial Interface reset enabled, device enters shutdown when SCL and SDA remain low for tPOR_DEB	

8.10 Temperature Supervision

An integrated temperature sensor provides over-temperature protection for the AS3676. This sensor generates a flag if the device temperature reaches the overtemperature threshold of 140°. The threshold has a hysteresis to prevent oscillation effects.

If the device temperature exceeds the T140 threshold all current sources, the charge pump and the dc-dc converter is disabled and the [ov_temp](#) flag is set. After decreasing the temperature by THYST operation is resumed.

The ov_temp flag can only be reset by first writing a 1 and then a 0 to the register bit [rst_ov_temp](#).

Bit [ov_temp_on](#) = 1 activates temperature supervision ([Table 118](#)). It is recommended to leave this bit set (default state).

Table 117. Overtemperature Detection

Symbol	Parameter	Condition			Min	Typ	Max	Unit
T140	ov_temp Rising Threshold					140		°C
THYST	ov_temp Hysteresis					5		°C

Table 118. *Reset and Overtemp control* Register

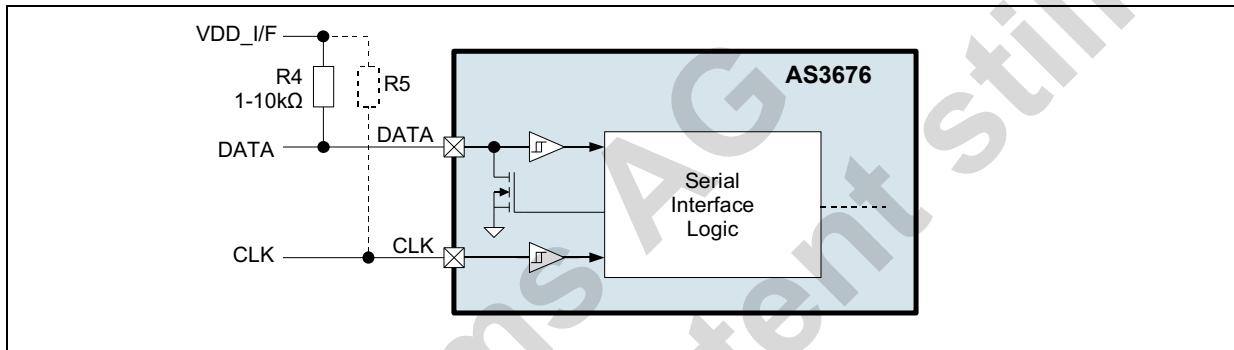
Addr: 29h		Reset and Overtemp control				
		This register reads and resets the overtemperature flag.				
Bit	Bit Name	Default	Access	Description		
0	ov_temp_on	1	W	Activates/deactivates device temperature supervision. Default: Off - all other bits are only valid if this bit is set to 1		
				0	Temperature supervision is disabled. No reset will be generated if the device temperature exceeds 140°C	
				1	Temperature supervision is enabled	

Table 118. *Reset and Overtemp control Register (Continued)*

Addr: 29h		Reset and Overtemp control				
		This register reads and resets the overtemperature flag.				
Bit	Bit Name	Default	Access	Description		
1	ov_temp	N/A	R	1	Indicates that the overtemperature threshold has been reached; this flag is not cleared by an overtemperature reset. It has to be cleared using rst_ov_temp	
2	rst_ov_temp	0	R/W	The ov_temp flag is cleared by first setting this bit to 1, and then setting this bit to 0.		

8.11 Serial Interface

The AS3676 is controlled using serial interface pins CLK and DATA:

Figure 34. *Serial interface block diagram*

The clock line CLK is never held low by the AS3676 (as the AS3676 does not use clock stretching of the bus).

Table 119. *Serial Interface Voltages and Timings*

Symbol	Parameter	Condition	Min	Typ	Max	Unit
V _{IHI} /F	High Level Input Voltage	Pins DATA and CLK	1.38		V _{BAT}	V
V _{ILI} /F	Low Level Input Voltage		0.0		0.52	V
V _{HYSTI} /F	Hysteresis			0.1		V
t _{RISE}	Rise Time		0		1000	ns
t _{FALL}	Fall Time		0		300	ns
t _{CLK_FILTER}	Spike Filter on CLK			100		ns
t _{DATA_FILTER}	Spike Filter on DATA			300		ns

The AS3676 is compatible to the NXP two wire specification http://www.nxp.com/acrobat_download/literature/9398/39340011.pdf, Version 2.1, January 2000 for standard and fast mode (no high speed mode).

8.11.1 Serial Interface Features

- Fast Mode Capability (Maximum Clock Frequency is 400 kHz)
- 7-bit Addressing Mode
- Write Formats
 - Single-Byte Write
 - Page-Write

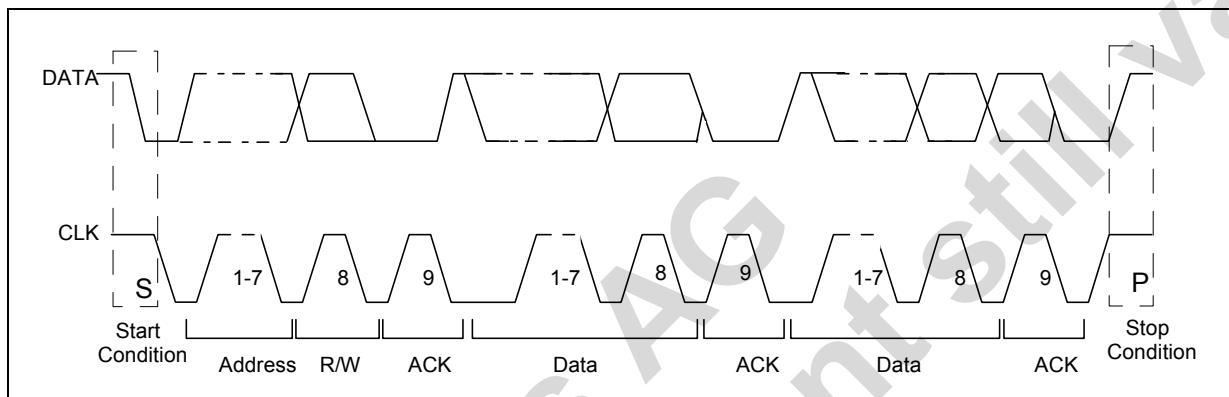
- Read Formats
 - Current-Address Read
 - Random-Read
 - Sequential-Read
- DATA Input Delay and CLK spike filtering by integrated RC components

8.11.2 Device Address Selection

The serial interface address of the AS3676 has the following address:

- 80h – Write Commands
- 81h – Read Commands

Figure 35. Complete Serial Data Transfer



Serial Data Transfer Formats

Definitions used in the serial data transfer format diagrams are listed in the following table:

Table 120. Serial Data Transfer Byte Definitions

Symbol	Definition	R/W (AS3676 Slave)	Note
S	Start Condition after Stop	R	1 bit
Sr	Repeated Start	R	1 bit
DW	Device Address for Write	R	10000000b (80h).
DR	Device Address for Read	R	10000001b (81h)
WA	Word Address	R	8 bits
A	Acknowledge	W	1 bit
N	Not Acknowledge	R	1 bit
reg_data	Register Data/Write	R	8 bits
data (n)	Register Data/read	R	1 bit
P	Stop Condition	R	8 bits
WA++	Increment Word Address Internally	R	During Acknowledge

Figure 36. Serial Interface Byte Write

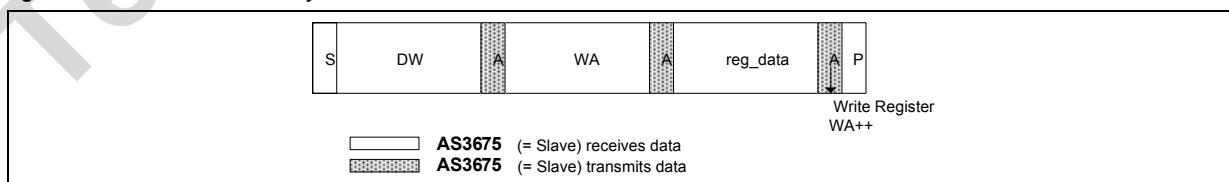
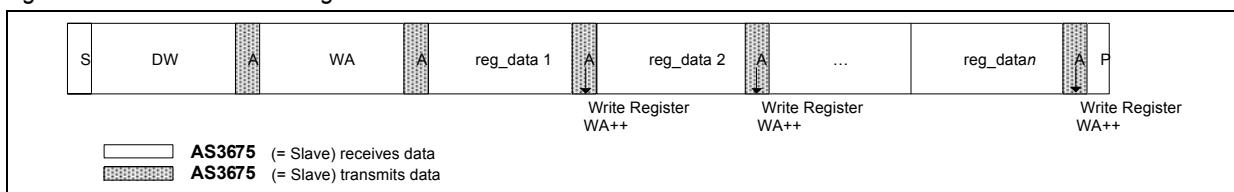


Figure 37. Serial Interface Page Write



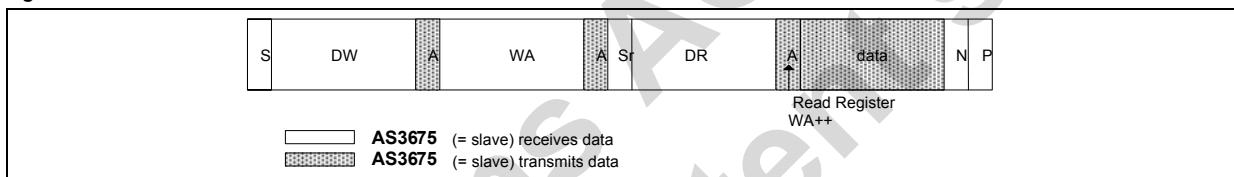
Byte Write and Page Write formats are used to write data to the slave.

The transmission begins with the START condition, which is generated by the master when the bus is in IDLE state (the bus is free). The device-write address is followed by the word address. After the word address any number of data bytes can be sent to the slave. The word address is incremented internally, in order to write subsequent data bytes on subsequent address locations.

For reading data from the slave device, the master has to change the transfer direction. This can be done either with a repeated START condition followed by the device-read address, or simply with a new transmission START followed by the device-read address, when the bus is in IDLE state. The device-read address is always followed by the 1st register byte transmitted from the slave. In Read Mode any number of subsequent register bytes can be read from the slave. The word address is incremented internally.

The following diagrams show the serial read formats supported by the AS3676.

Figure 38. Serial Interface Random Read

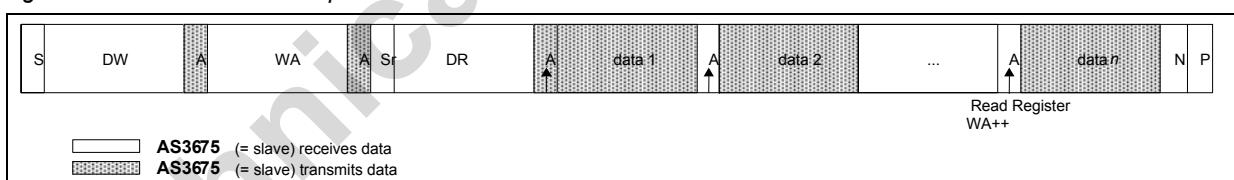


Random Read and Sequential Read are combined formats. The repeated START condition is used to change the direction after the data transfer from the master.

The word address transfer is initiated with a START condition issued by the master while the bus is idle. The START condition is followed by the device-write address and the word address.

In order to change the data direction a repeated START condition is issued on the 1st CLKpulse after the ACKNOWLEDGE bit of the word address transfer. After the reception of the device-read address, the slave becomes the transmitter. In this state the slave transmits register data located by the previous received word address vector. The master responds to the data byte with a NOT ACKNOWLEDGE, and issues a STOP condition on the bus.

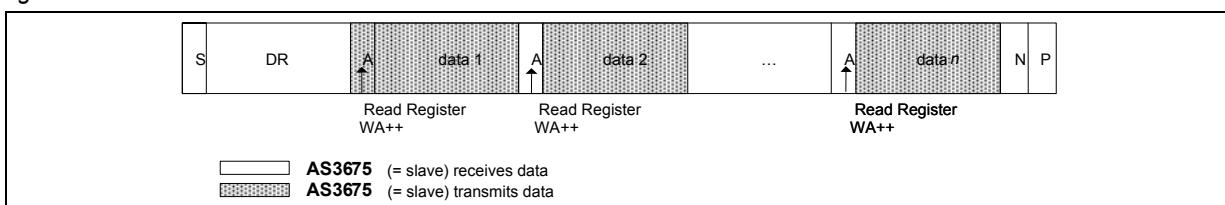
Figure 39. Serial Interface Sequential Read



Sequential Read is the extended form of Random Read, as multiple register-data bytes are subsequently transferred.

In contrast to the Random Read, in a sequential read the transferred register-data bytes are responded by an acknowledge from the master. The number of data bytes transferred in one sequence is unlimited (consider the behavior of the word-address counter). To terminate the transmission the master has to send a NOT ACKNOWLEDGE following the last data byte and subsequently generate the STOP condition.

Figure 40. Serial Interface Current Address Read



To keep the access time as small as possible, this format allows a read access without the word address transfer in advance to the data transfer. The bus is idle and the master issues a START condition followed by the Device-Read address.

Analogous to Random Read, a single byte transfer is terminated with a NOT ACKNOWLEDGE after the 1st register byte. Analogous to Sequential Read an unlimited number of data bytes can be transferred, where the data bytes must be responded to with an ACKNOWLEDGE from the master.

For termination of the transmission the master sends a NOT ACKNOWLEDGE following the last data byte and a subsequent STOP condition.

8.12 Operating Modes

If the voltages on CLK and DATA is less than VPOR_PERI for $> t_{POR_DEB}$ (see Table 115 on page 75), the AS3676 is in shutdown mode and its current consumption is minimized ($I_{BAT} = I_{SHUTDOWN}$) and all internal registers are reset to their default values.

If the voltage at CLK or DATA rises above VPOR_PERI, the AS3676 serial interface is enabled and the AS3676 and the standby mode is selected. The AS3676 is switched automatically from standby mode ($I_{BAT} = I_{STANBY}$) into normal mode ($I_{BAT} = I_{ACTIVE}$) and back, if one of the following blocks are activated:

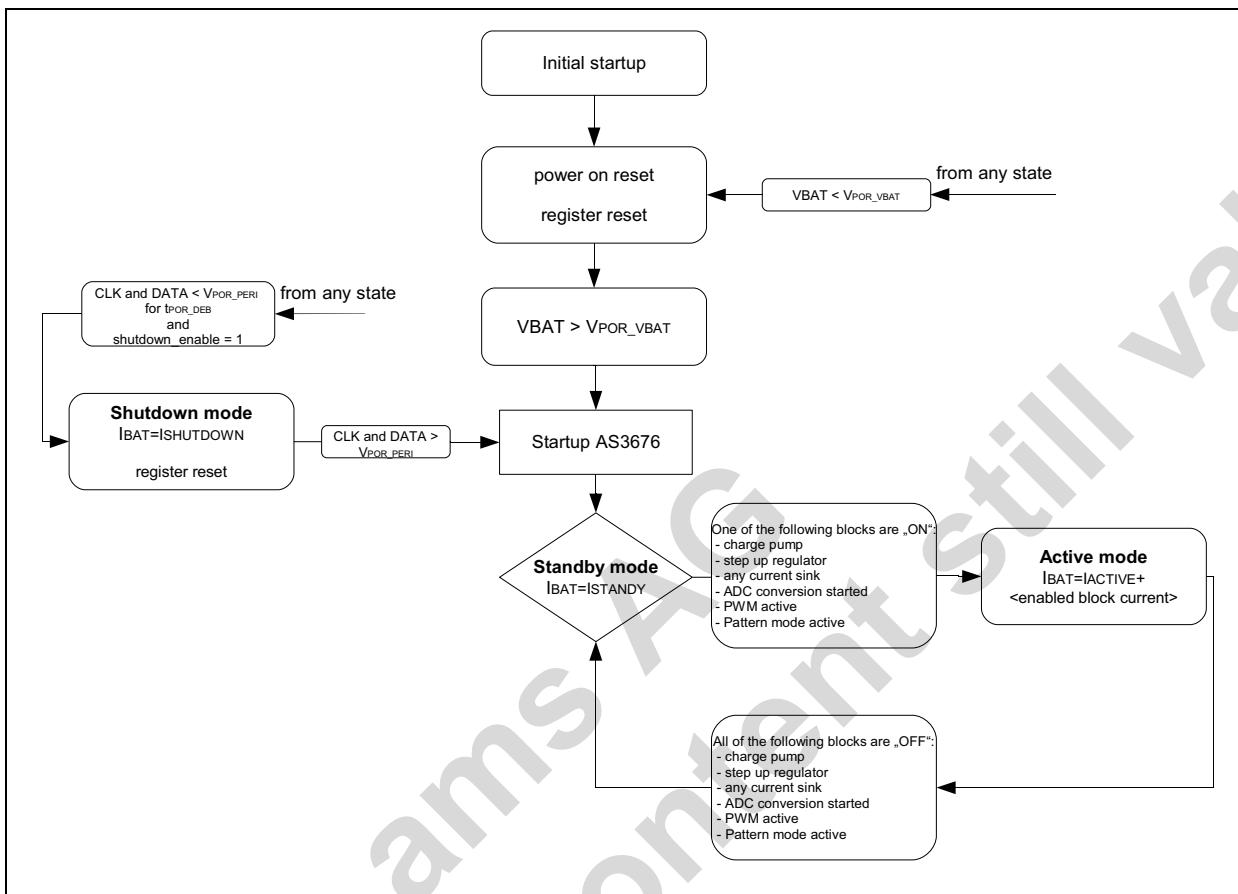
- Charge pump
- Step up regulator
- Any current sink
- ADC conversion started
- PWM active
- Pattern mode active.

If any of these blocks are already switched on the internal oscillator is running and a write instruction to the registers is directly evaluated within 1 internal CLK cycle (typ. 1 μ s)

If all these blocks are disabled, a write instruction to enable these blocks is delayed by 64 CLK cycles (oscillator will startup, within max 200 μ s).

The mode switching is shown in [Figure 41](#):

Figure 41. Startup and Operating Mode Selection



9 Register Map

Table 121. Registermap

Register Definition	Addr	Default	Content												
			b7	b6	b5	b4	b3	b2	b1	b0					
Reg. control	00h	00					step_up_on	cp_on		ldo_on					
curr12 control	01h	00h				curr2_mode			curr1_mode						
curr rgb control	02h	00h	curr6_mode		rgb3_mode		rgb2_mode		rgb1_mode						
curr3 control1	03h	00h	curr33_mode		curr32_mode		curr31_mode		curr30_mode						
curr4 control	04h	00h				curr43_mode	curr42_mode		curr41_mode						
GPIO output 1	05h	00h	gpi_curr_33_en	gpi_curr_32_en	gpi_curr_31_en	gpi_curr_30_en	gpi_en	gpi_curr_6_en	gpi_curr_2_en	gpi_curr_1_en					
GPIO signal 1	06h	00h	gpi_curr_33_in	gpi_curr_32_in	gpi_curr_31_in	gpi_curr_30_in	gpi_in	gpi_curr_6_in	gpi_curr_2_in	gpi_curr_1_in					
LDO Voltage	07h	00h				ldo_voltage									
Curr1 current	09h	00h	curr1_current												
Curr2 current	0Ah	00h	curr2_current												
Rgb1 current	0Bh	00h	rgb1_current												
Rgb2 current	0Ch	00h	rgb2_current												
Rgb3 current	0Dh	00h	rgb3_current												
Curr3x strobe	0Eh	00h				curr3x_strobe									
Curr3x preview	0Fh	00h				curr3x_preview									
Curr3x other	10h	00h				curr3x_other									
Curr3 strobe control	11h	00h	strobe_timing				strobe_mode		strobe_ctrl						
Curr3 control2	12h	00h	strobe_p_in		curr3x_strobe_hi	gh			preview_ctrl	preview_off_after_strobe					
Curr41 current	13h	00h	curr41_current												
Curr42 current	14h	00h	curr42_current												
Curr43 current	15h	00h	curr43_current												
Pwm control	16h	00h			pwm_dim_speed			pwm_dim_mode							
pwm code	17h	00h	pwm_code												
Pattern control	18h	00h	curr33_pattern	curr32_pattern	curr31_pattern	curr30_pattern	softdim_pattern	pattern_delay		pattern_color					
Pattern data0	19h	00h	pattern_data[7:0]												
Pattern data1	1Ah	00h	pattern_data[15:8]												
Pattern data2	1Bh	00h	pattern_data[23:16]												
Pattern data3	1Ch	00h	pattern_data[31:24]												
GPIO control	1Eh	44h	gpio2_pulls		gpio2_mode		gpio1_pulls		gpio1_mode						
GPIO driving cap	20h	00h							gpio2_low_curr	gpio1_low_curr					

Table 121. Registermap

Register Definition	Addr	Default	Content											
			b7	b6	b5	b4	b3	b2	b1	b0				
DCDC control1	21h	00h	step_up_vtuning				step_up_fb		step_up_frequ					
DCDC control2	22h	04h	step_up_fb_auto				step_up_lowcur	step_up_prot	skip_fast	step_up_res				
CP control	23h	00h		cp_auto_on	cp_start_debounce	cp_mode_switching	cp_mode		cp_clk					
CP mode Switch1	24h	00h		rgb3_on_cp	rgb2_on_cp	rgb1_on_cp	curr33_on_cp	curr32_on_cp	curr31_on_cp	curr30_on_cp				
CP mode Switch2	25h	00h	curr6_on_cp			curr43_on_cp	curr42_on_cp	curr41_on_cp	curr2_on_cp	curr1_on_cp				
ADC_control	26h	03h	start_convnersion		adc_select									
ADC_MSB result	27h	NA	result_not_ready	D9:D3										
ADC_LSB result	28h	NA						D2:D0						
Reset and Overtemp control	29h	01h			shutdwn_enab		rst_ov_temp	ov_temp	ov_temp_on					
Curr low voltage status1	2Ah	NA	curr6_low_v	rgb3_low_v	rgb2_low_v	rgb1_low_v	curr33_low_v	curr32_low_v	curr31_low_v	curr30_low_v				
Curr low voltage status2	2Bh	NA				curr43_low_v	curr42_low_v	curr41_low_v	curr2_low_v	curr1_low_v				
gpio current	2Ch	00h		pattern_slow		pattern_delay2								
curr6 current	2Fh	00h	curr6_current											
Adder Current 1	30h	00h	adder_current1 (can be enabled for CURR30, CURR1, RGB1, CURR41)											
Adder Current 2	31h	00h	adder_current2 (can be enabled for CURR31, CURR2, RGB2, CURR42)											
Adder Current 3	32h	00h	adder_current3 (can be enabled for CURR32, CURR6, RGB3, CURR43)											
Adder Enable 1	33h	00h		curr43_adder	curr42_adder	curr41_adder	rgb3_adder	rgb2_adder	rgb1_adder					
Adder Enable 2	34h	00h	curr33_adder	curr32_adder	curr31_adder	curr30_adder	curr6_adder	curr2_adder	curr1_adder					
Subtract Enable	35h	00h					sub_en4	sub_en3	sub_en2	sub_en1				
ASIC ID1	3Eh	AEh	1	0	1	0	1	1	1	0				
ASIC ID2	3Fh	5Xh	0	1	0	1	revision							
Curr30 current	40h	00h	curr30_current											
Curr31 current	41h	00h	curr31_current											
Curr32 current	42h	00h	curr32_current											
Curr33 current	43h	00h	curr33_current											
Audio Control	46h	00h	audio_speed_down			audio_color				aud_buf_on				
Audio input	47h	00h	audio_di_s_start	audio_m_an_start	agc_ctrl			audio_gain						

Table 121. Registermap

Register Definition	Addr	Default	Content																
			b7	b6	b5	b4	b3	b2	b1	b0									
Audio output	48h	00h		curr4x_aud_on	rgbx_aud_on	curr126_aud_on		aud_amplitude											
GPIO output 2	50h	00h	gpi_curr43_en	gpi_curr42_en	gpi_curr41_en	gpi_rgb3_en	gpi_rgb2_en	gpi_rgb1_en	gpio2_o ut	gpio1_o ut									
GPIO signal 2	51h	00h	gpi_curr43_in	gpi_curr42_in	gpi_curr41_in	gpi_rgb3_in	gpi_rgb2_in	gpi_rgb1_in	gpio2_in	gpio1_in									
Adder Current 4	52h	00h	adder_current4 (can be enabled for CURR33)																
CURR3x audio source	53h	00h	curr33_aud_src[1:0]	curr32_aud_src[1:0]	curr31_aud_src[1:0]	curr30_aud_src[1:0]													
Pattern End	54h	00h								pattern_end									
Audio Control 2	55h	00h		audio_source		audio_speed_up													
DLS mode control1	56h	00h		rgb3_on_dls	rgb2_on_dls	rgb1_on_dls	curr33_on_dls	curr32_on_dls	curr31_on_dls	curr30_on_dls									
DLS mode control2	57h	00h	curr6_on_dls			curr43_on_dls	curr42_on_dls	curr41_on_dls	curr2_on_dls	curr1_on_dls									
ALS control	90h	00h					amb_gain		amb_on										
ALS filter	91h	00h	amb_filter_down				amb_filter_up												
ALS offset	92h	00h	amb_offset																
ALS result	93h	00h	amb_result																
ALS curr12 group	94h	00h					curr2_amb_group		curr1_amb_group										
ALS rgb group	95h	00h	curr6_amb_group	rgb3_amb_group		rgb2_amb_group		rgb1_amb_group											
ALS curr3x group	96h	00h	curr33_amb_group	curr32_amb_group		curr31_amb_group		curr30_amb_group											
ALS curr4x group	97h	00h		curr43_amb_group		curr42_amb_group		curr41_amb_group											
ALS group 1 Y0	98h	00h	group1_y0																
ALS group 1 Y3	99h	00h	group1_y3																
ALS group 1 X1	9Ah	00h	group1_x1																
ALS group 1 K1	9Bh	00h	group1_k1																
ALS group 1 X2	9Ch	00h	group1_x2																
ALS group 1 K2	9Dh	00h	group1_k2																
ALS group 2 Y0	9Eh	00h	group2_y0																
ALS group 2 Y3	9Fh	00h	group2_y3																
ALS group 2 X1	A0h	00h	group2_x1																
ALS group 2 K1	A1h	00h	group2_k1																
ALS group 2 X2	A2h	00h	group2_x2																
ALS group 2 K2	A3h	00h	group2_k2																
ALS group 3 Y0	A4h	00h	group3_y0																
ALS group 3 Y3	A5h	00h	group3_y3																
ALS group 3 X1	A6h	00h	group3_x1																

Table 121. Registermap

Register Definition	Addr	Default	Content							
			b7	b6	b5	b4	b3	b2	b1	b0
ALS group 3 K1	A7h	00h								group3_k1
ALS group 3 X2	A8h	00h								group3_x2
ALS group 3 K2	A9h	00h								group3_k2

Note: If writing to register, write 0 to unused bits

Write to read only bits will be ignored

yellow color = read only

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10 External Components

Table 122. External Components List

Part Number	Min	Value Typ	Max	tol. (min.)	Rating (max)	Notes	Package (min.) ¹
C1		1µF		±20%	6.3V	Ceramic, X5R (V2_5 output) (e.g. Taiyo Yuden JMK105BJ105KV-F)	0402
C2		1µF		±20%	6.3V	Ceramic, X5R (VBAT) (e.g. Taiyo Yuden JMK105BJ105KV-F)	0402
C3		1µF		±20%	6.3V	Ceramic, X5R (Charge Pump) (e.g. Taiyo Yuden JMK105BJ105KV-F)	0402
C4		1µF		±20%	6.3V	Ceramic, X5R (Charge Pump) (e.g. Taiyo Yuden JMK105BJ105KV-F)	0402
C5		2.2µF		±20%	6.3V	Ceramic, X5R (Charge Pump Output) (e.g. Taiyo Yuden JMK107BJ225MA-T)	0403
C6		1µF		±20%	6.3V	Ceramic, X5R (Step Up DCDC input) (e.g. Taiyo Yuden JMK105BJ105KV-F)	0402
C7 - optional	only required in voltage feedback mode of DCDC	1.5nF		±20%	25V	Ceramic, X5R (Step Up DCDC Feedback, 150pF for over voltage protection)	0402
C8 - optional		15nF		±20%	6.3V	Ceramic, X5R (Step Up DCDC Feedback, 1.5nF for over voltage protection)	0402
C9		10µF		±20%	25V	Ceramic, X5R, X7R (Step Up DCDC output) e.g. Murata GRM188R61E106MA73 ²	0603
		4.7µF				e.g. Taiyo Yuden TMK316BJ475KD	3.2x1.6x 0.85mm
C10		2.2µF		±20%	6.3V	Ceramic, X5R (VANA/GPI output) (e.g. Taiyo Yuden JMK107BJ225MA-T)	0402
R1		100mΩ		±5%		Shunt Resistor	0603
R2		1MΩ		±1%		Step Up DC/DC Converter Voltage Feedback	0201
R3		100kΩ		±1%		Step Up DC/DC Converter Voltage Feedback - not required for over voltage protection	0201
R4, R5		1-10kΩ		±1%		DATA, CLK Pullup resistor – usually already inside master	0201
L1		10µH (15µH ³)		±20%		Recommended Type: Coilcraft LPS3010-123 ⁴ or Murata LQH3NPN100NJ0, LQH3NPN150NJ0 or Panasonic ELLSFG100MA or TDK VLF3012Aor Taiyo Yuden NRH3012T100MN	3x3x1mm

Table 122. External Components List

Part Number	Min	Value Typ	Max	tol. (min.)	Rating (max)	Notes	Package (min.) ¹
Q1 (+ D1)	Fairchild FDFMA3N109					Integrated NMOS and Schottky diode	MicroFET 2x2mm
	Fairchild FDFME3N311ZT or OnSemi NTLUF4189NZ					Integrated NMOS and Schottky diode; recommended for configurations up to 12 LEDs	MicroFET 1.6x1.6mm
D2:D20	LED					As required by application	

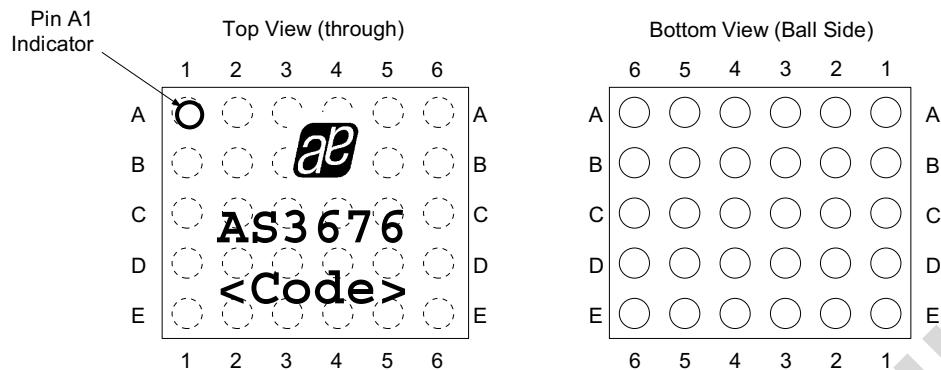
1. in 1/100 inch (unless otherwise specified)
2. Specified >1µF at 20V; use up to 20V DCDC output voltage
3. Results in improved efficiency compared to 10µH
4. For highest efficiency

As system efficiency is depending on LED configurations, external components and operating conditions, see austriamicrosystems application note 'Mobile Backlight Selection Guide 1vx.pdf' for optimizing efficiency and components size (coil L1, transistor Q1, diode D1).

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11 Package Drawings and Markings

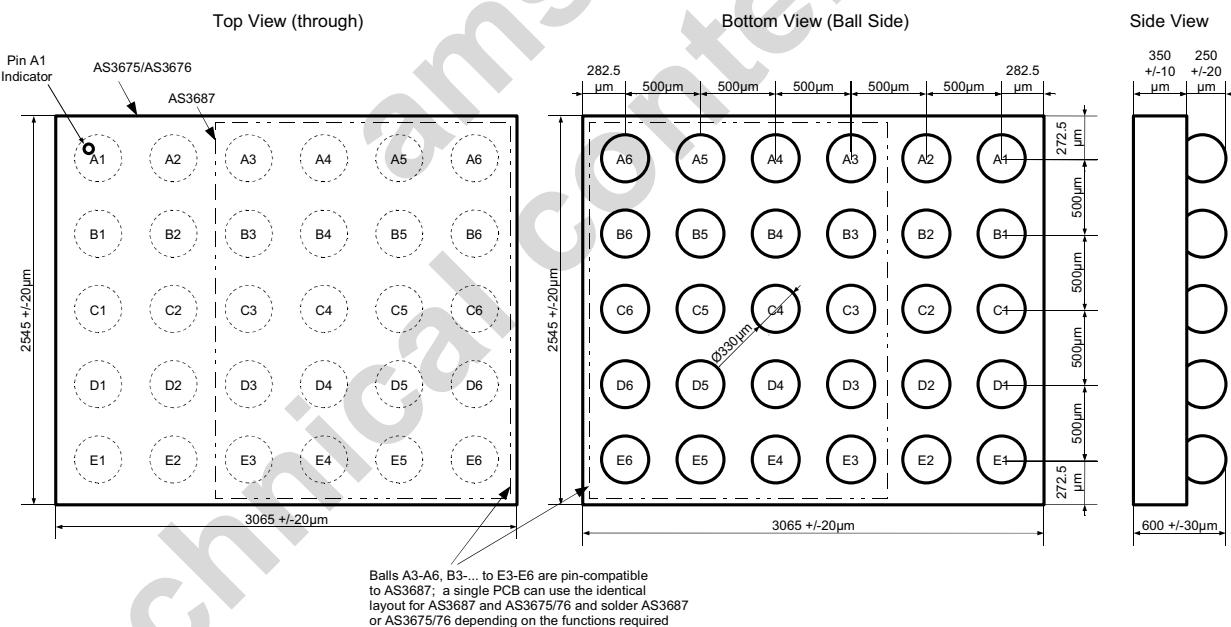
Figure 42. WL-CSP30 3x2.5mm 6x5 Balls Package Drawing



Note:

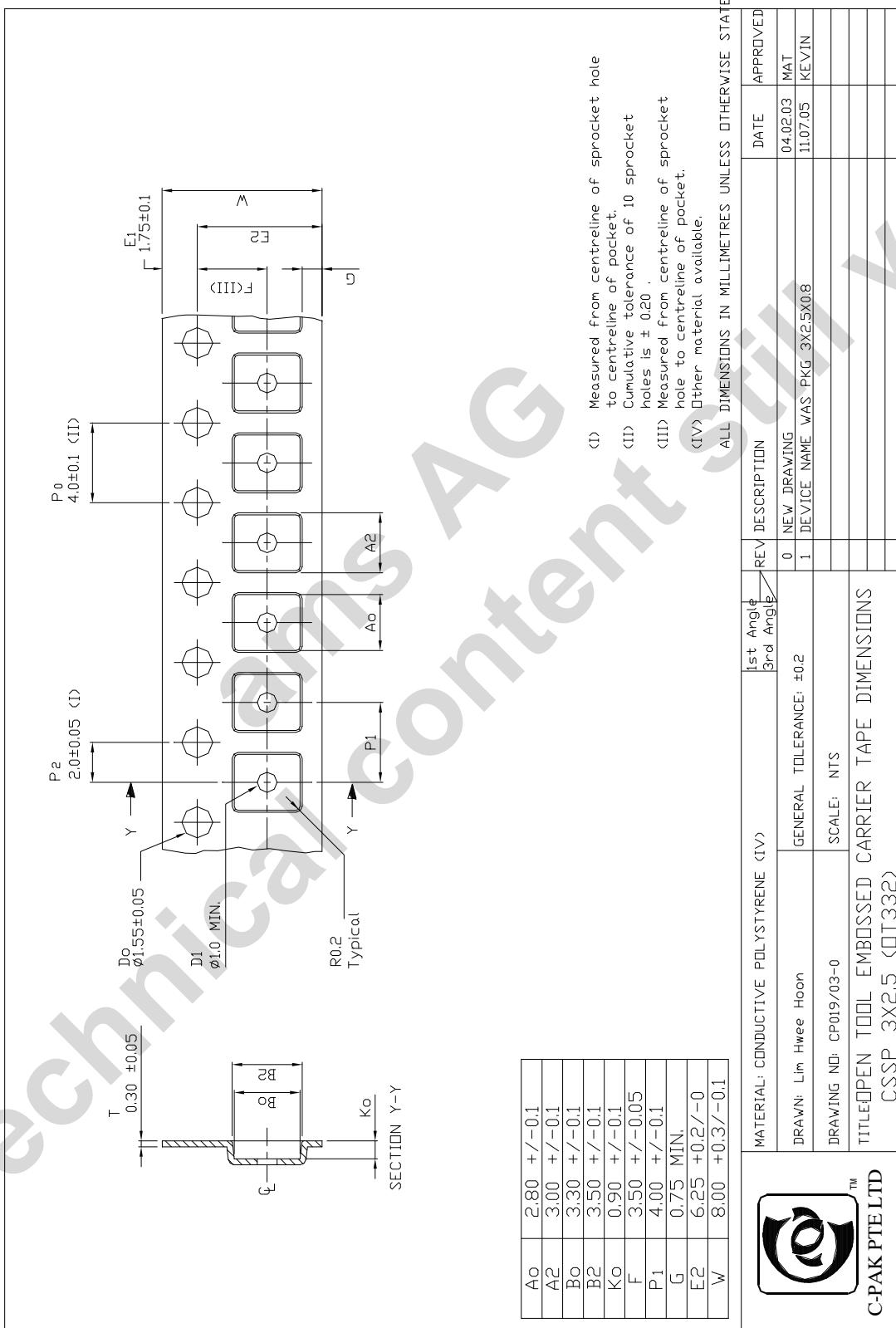
- Line 1: austriamicrosystems logo
- Line 2: AS3676
- Line 3: <Code>
Encode datecode 4 characters

Figure 43. WL-CSP30 3x2.5mm 6x5 Balls Detail Dimensions



11.1 Tape & Reel Information

Figure 44. Tape & Reel Dimensions



12 Ordering Information

The devices are available as the standard products shown in [Table 123](#).

Table 123. Ordering Information

Model	Description	Delivery Form	Package
AS3676-ZWLT	AS3676 Wafer Level Chip Scale Package, size 3x2.5mm, 6x5 balls, 0.5mm pitch, Pb-Free	Tape & Reel	30pin WL-CSP (3x2.5mm) RoHS compliant / Pb-Free

Note: All products are RoHS compliant and austriamicrosystems green.

Buy our products or get free samples online at ICdirect: <http://www.austriamicrosystems.com/ICdirect>

Technical support is found at <http://www.austriamicrosystems.com/Technical-Support>

For further information and requests, please contact us <mailto:sales@austriamicrosystems.com>
or find your local distributor at <http://www.austriamicrosystems.com/distributor>

Note: AS3676-ZWLT

AS3676-

Z Temperature Range: -30°C - 85°C

WL Package: Wafer Level Chip Scale Package (WL-CSP) 3x2.5mm

T Delivery Form: Tape & Reel

Technical content still valid

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