

**Data Sheet**
**HMC1040CHIPS**
**FEATURES**

**Low noise figure: 2 dB typical**

**High gain: 25.0 dB typical**

**P1dB output power: 13.5 dBm, 24 GHz to 40 GHz**

**High output IP3: 25.5 dBm typical**

**Die size: 1.309 mm × 1.48 × 0.102 mm**

**APPLICATIONS**

**Software defined radios**

**Electronic warfare**

**Radar applications**

**Satellite communication**

**Electronic warfare**

**Instrumentation**

**Telecommunications**

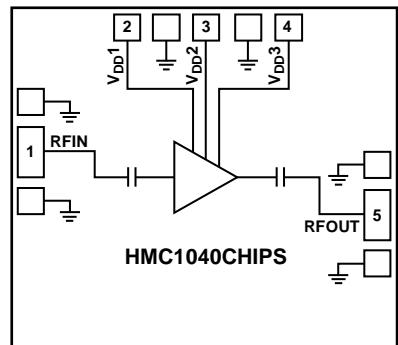
**FUNCTIONAL BLOCK DIAGRAM**


Figure 1.

**GENERAL DESCRIPTION**

The HMC1040CHIPS is a gallium arsenide (GaAs), pseudomorphic high electron mobility transistor (pHEMT), monolithic microwave integrated circuit (MMIC), low noise wideband amplifier that operates from 20 GHz to 44 GHz. The HMC1040CHIPS is self biased and provides a typical gain of 25.0 dB, a 2 dB typical noise figure, and a typical output third-order intercept (IP3) of 25.5 dBm typical, requiring only 65 mA from a 2.5 V supply voltage. The typical saturated

output power ( $P_{SAT}$ ) of 15.5 dBm enables the low noise amplifier (LNA) to function as a local oscillator (LO) driver for many of Analog Devices, Inc., balanced, in phase quadrature (I/Q) or image rejection mixers.

The HMC1040CHIPS also feature inputs and outputs that are internally matched to  $50\ \Omega$ , making it ideal for surface-mounted technology (SMT)-based, high capacity microwave radio applications.

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## REVISION HISTORY

4/2018—Revision 0: Initial Version

## SPECIFICATIONS

### 20 GHz TO 24 GHz FREQUENCY RANGE

$T_A = 25^\circ\text{C}$ , supply voltage ( $V_{DD} = 2.5 \text{ V}$ ), and supply current ( $I_{DQ} = 65 \text{ mA}$ , unless otherwise noted.

Table 1.

| Parameter                         | Symbol           | Min | Typ   | Max | Unit                 |
|-----------------------------------|------------------|-----|-------|-----|----------------------|
| FREQUENCY RANGE                   |                  | 20  |       | 24  | GHz                  |
| GAIN                              |                  |     | 24.5  |     | dB                   |
| Gain Variation Over Temperature   |                  |     | 0.018 |     | dB/ $^\circ\text{C}$ |
| NOISE FIGURE                      | NF               |     | 4     |     | dB                   |
| RETURN LOSS                       |                  |     |       |     |                      |
| Input                             |                  |     | 18    |     | dB                   |
| Output                            |                  |     | 18    |     | dB                   |
| OUTPUT                            |                  |     |       |     |                      |
| Output Power for 1 dB Compression | P <sub>1dB</sub> |     | 12.5  |     | dBm                  |
| Saturated Output Power            | P <sub>SAT</sub> |     | 13.5  |     | dBm                  |
| Output Third-Order Intercept      | IP3              |     | 21    |     | dBm                  |
| SUPPLY                            |                  |     |       |     |                      |
| Current                           | I <sub>DQ</sub>  |     | 65    |     | mA                   |
| Voltage                           | V <sub>DD</sub>  | 2   | 2.5   | 3.5 | V                    |

### 24 GHz TO 32 GHz FREQUENCY RANGE

$T_A = 25^\circ\text{C}$ ,  $V_{DD} = 2.5 \text{ V}$ , and  $I_{DQ} = 65 \text{ mA}$ , unless otherwise noted.

Table 2.

| Parameter                         | Symbol           | Min | Typ   | Max | Unit                 |
|-----------------------------------|------------------|-----|-------|-----|----------------------|
| FREQUENCY RANGE                   |                  | 24  |       | 32  | GHz                  |
| GAIN                              |                  | 23  | 25.0  |     | dB                   |
| Gain Variation Over Temperature   |                  |     | 0.021 |     | dB/ $^\circ\text{C}$ |
| NOISE FIGURE                      | NF               |     | 2.5   | 2.7 | dB                   |
| RETURN LOSS                       |                  |     |       |     |                      |
| Input                             |                  |     | 13    |     | dB                   |
| Output                            |                  |     | 13    |     | dB                   |
| OUTPUT                            |                  |     |       |     |                      |
| Output Power for 1 dB Compression | P <sub>1dB</sub> |     | 13.5  |     | dBm                  |
| Saturated Output Power            | P <sub>SAT</sub> |     | 14.5  |     | dBm                  |
| Output Third-Order Intercept      | IP3              |     | 22.5  |     | dBm                  |
| SUPPLY                            |                  |     |       |     |                      |
| Current                           | I <sub>DQ</sub>  |     | 65    |     | mA                   |
| Voltage                           | V <sub>DD</sub>  | 2   | 2.5   | 3.5 | V                    |

**32 GHz TO 40 GHz FREQUENCY RANGE**

$T_A = 25^\circ\text{C}$ ,  $V_{DD} = 2.5 \text{ V}$ , and  $I_{DQ} = 65 \text{ mA}$ , unless otherwise noted.

**Table 3.**

| Parameter                               | Symbol           | Min | Typ         | Max | Unit                       |
|---|------------------|-----|-------------|-----|----------------------------|
| FREQUENCY RANGE                         |                  | 32  |             | 40  | GHz                        |
| GAIN<br>Gain Variation Over Temperature |                  | 21  | 23<br>0.021 |     | dB<br>dB/ $^\circ\text{C}$ |
| NOISE FIGURE                            | NF               | 2   |             | 2.7 | dB                         |
| RETURN LOSS                             |                  |     |             |     |                            |
| Input                                   |                  |     | 11          |     | dB                         |
| Output                                  |                  |     | 13          |     | dB                         |
| OUTPUT                                  |                  |     |             |     |                            |
| Output Power for 1 dB Compression       | P <sub>1dB</sub> |     | 13.5        |     | dBm                        |
| Saturated Output Power                  | P <sub>SAT</sub> |     | 15.5        |     | dBm                        |
| Output Third-Order Intercept            | IP <sub>3</sub>  |     | 24.5        |     | dBm                        |
| SUPPLY                                  |                  |     |             |     |                            |
| Current                                 | I <sub>DQ</sub>  |     | 65          |     | mA                         |
| Voltage                                 | V <sub>DD</sub>  | 2   | 2.5         | 3.5 | V                          |

**40 GHz TO 44 GHz FREQUENCY RANGE**

$T_A = 25^\circ\text{C}$ ,  $V_{DD} = 2.5 \text{ V}$ , and  $I_{DQ} = 65 \text{ mA}$ , unless otherwise noted.

**Table 4.**

| Parameter                               | Symbol           | Min | Typ         | Max | Unit                       |
|---|------------------|-----|-------------|-----|----------------------------|
| FREQUENCY RANGE                         |                  | 40  |             | 44  | GHz                        |
| GAIN<br>Gain Variation Over Temperature |                  | 19  | 21<br>0.023 |     | dB<br>dB/ $^\circ\text{C}$ |
| NOISE FIGURE                            | NF               | 2.5 |             | 3.2 | dB                         |
| RETURN LOSS                             |                  |     |             |     |                            |
| Input                                   |                  |     | 6           |     | dB                         |
| Output                                  |                  |     | 13          |     | dB                         |
| OUTPUT                                  |                  |     |             |     |                            |
| Output Power for 1 dB Compression       | P <sub>1dB</sub> |     | 14          |     | dBm                        |
| Saturated Output Power                  | P <sub>SAT</sub> |     | 16          |     | dBm                        |
| Output Third-Order Intercept            | IP <sub>3</sub>  |     | 25.5        |     | dBm                        |
| SUPPLY                                  |                  |     |             |     |                            |
| Current                                 | I <sub>DQ</sub>  |     | 65          |     | mA                         |
| Voltage                                 | V <sub>DD</sub>  | 2   | 2.5         | 3.5 | V                          |

## ABSOLUTE MAXIMUM RATINGS

Table 5.

| Parameter  | Rating  |
|--|---|
| Drain Bias Voltage ( $V_{DD}$ )  | 4 V dc  |
| Radio Frequency (RF) Input Power (RFIN)  | 5 dBm   |
| Continuous Power Dissipation ( $P_{DISS}$ ),<br>$T = 85^{\circ}\text{C}$ (Derate 5.46 mW/ $^{\circ}\text{C}$ Above 85 $^{\circ}\text{C}$ ) | 0.49 W  |
| Channel Temperature  | 175 $^{\circ}\text{C}$                            |
| Storage Temperature Range  | -65 $^{\circ}\text{C}$ to +150 $^{\circ}\text{C}$ |
| Operating Temperature Range  | -55 $^{\circ}\text{C}$ to +85 $^{\circ}\text{C}$  |
| Electrostatic Discharge (ESD) Sensitivity<br>Human Body Model (HBM)  | Class 0 passed,<br>100 V                          |

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

## THERMAL RESISTANCE

Thermal performance is directly linked to system design and operating environment.

Careful attention to PCB thermal design is required.

$\theta_{JC}$  is the channel to case thermal resistance, channel to bottom of die.

Table 6. Thermal Resistance

| Package Type | $\theta_{JC}$ | Unit                 |
|--------------|---------------|----------------------|
| C-5-6        | 183           | $^{\circ}\text{C/W}$ |

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.**  
Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

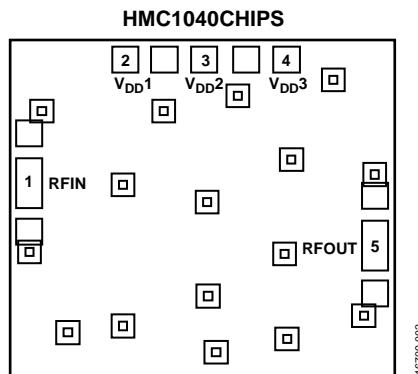


Figure 2. Pad Configuration

Table 7. Pad Function Descriptions

| Pad No.    | Mnemonic   | Description   |
|------------|--|---|
| 1          | RFIN   | Radio Frequency Input. This pad ac couples the RF signal, has a 5 kΩ resistor connected to GND, and is matched to 50 Ω. See Figure 3 for the interface schematic. |
| 2, 3, 4    | V <sub>DD1</sub> , V <sub>DD2</sub> , V <sub>DD3</sub> | Power Supply Voltages for the Amplifier. Connect a dc bias to provide drain current (I <sub>DD</sub> ). See Figure 4 for the interface schematic.                 |
| 5          | RFOUT  | RF Output. This pad ac couples the RF signal, has a 5 kΩ resistor connected to GND, and is matched to 50 Ω. See Figure 5 for the interface schematic.             |
| Die Bottom | GND  | Ground. Die bottom must be connected to RF/dc ground. See Figure 6 for the interface schematic.   |

## INTERFACE SCHEMATICS

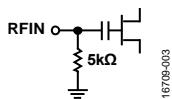


Figure 3. RFIN Interface Schematic

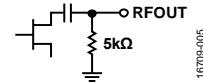


Figure 5. RFOUT Interface Schematic

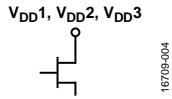


Figure 4. V<sub>DD1</sub>, V<sub>DD2</sub>, V<sub>DD3</sub>, Interface Schematic



Figure 6. GND Interface Schematic

## TYPICAL PERFORMANCE CHARACTERISTICS

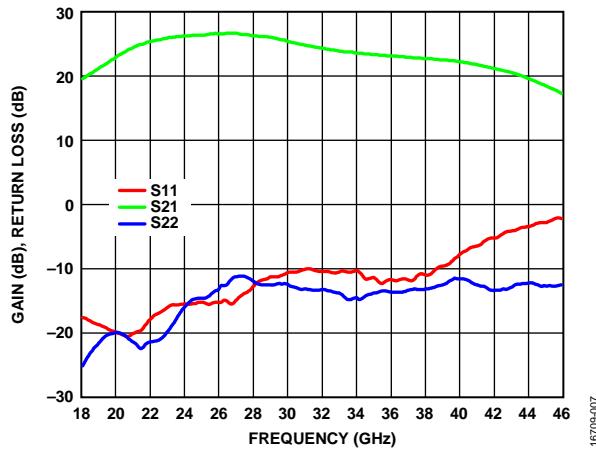


Figure 7. Gain and Return Loss vs. Frequency

16709-007

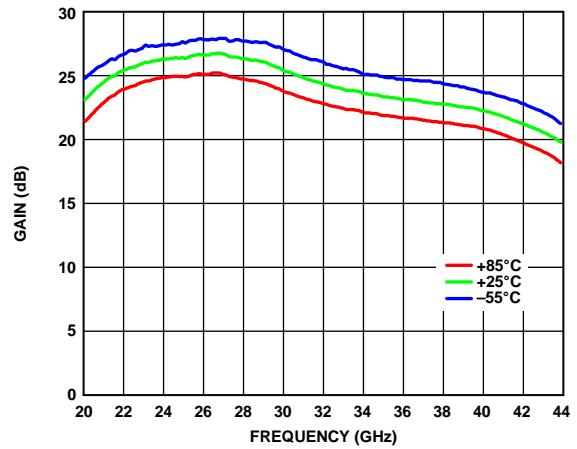


Figure 10. Gain vs. Frequency for Various Temperatures

16709-010

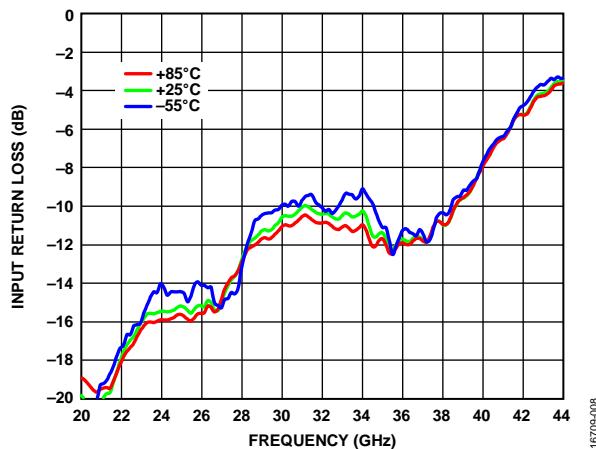


Figure 8. Input Return Loss vs. Frequency for Various Temperatures

16709-008

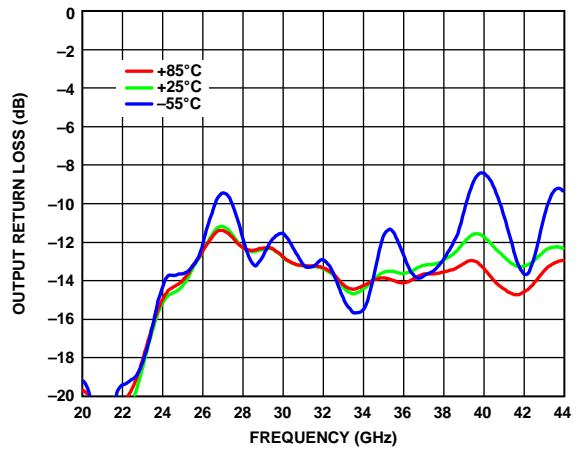


Figure 11. Output Return Loss vs. Frequency for Various Temperatures

16709-011

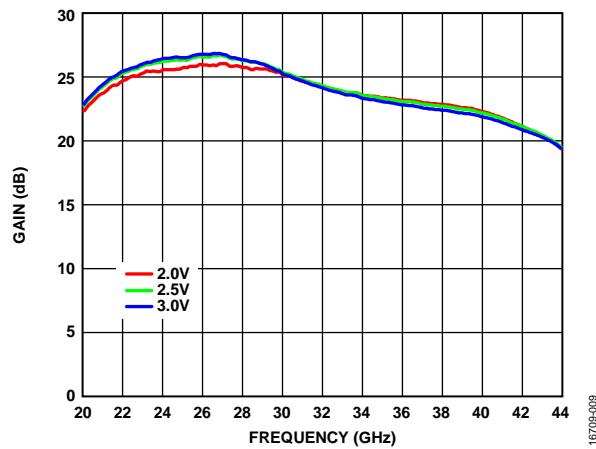


Figure 9. Gain vs. Frequency for Various Supply Voltages

16709-008

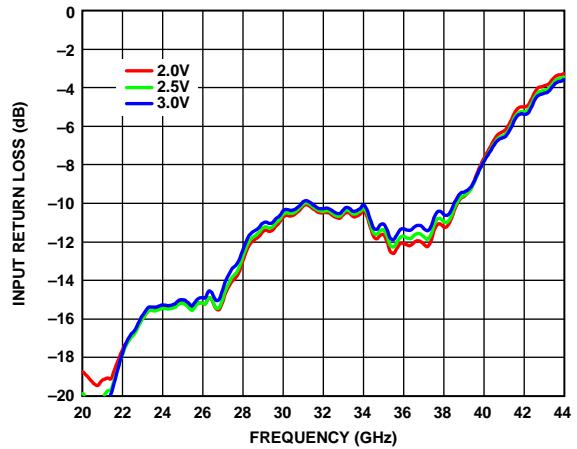
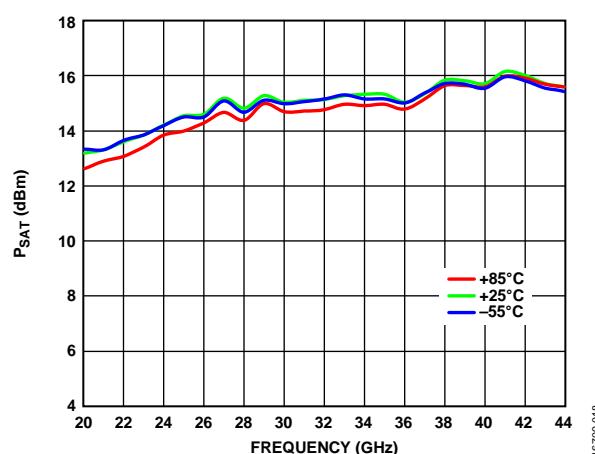
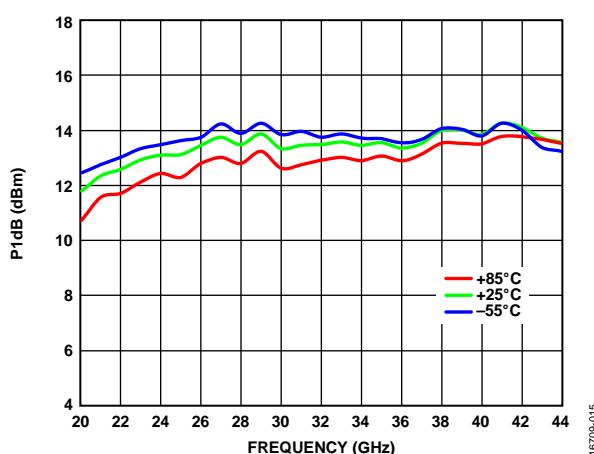
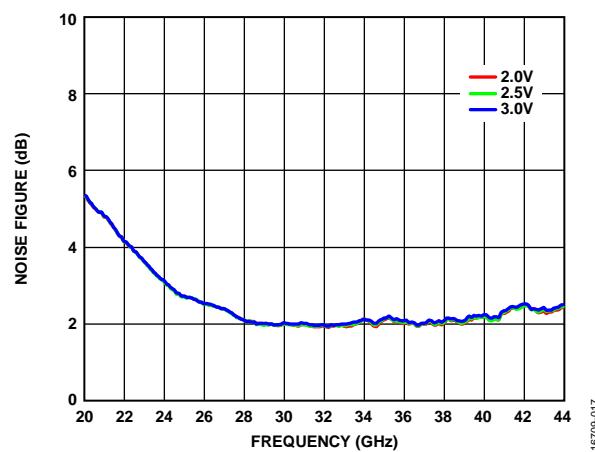
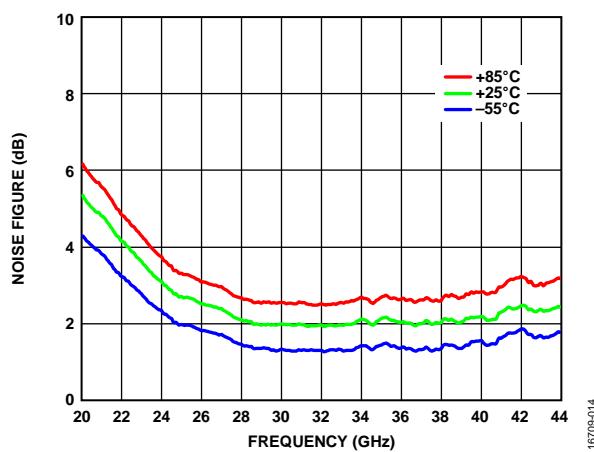
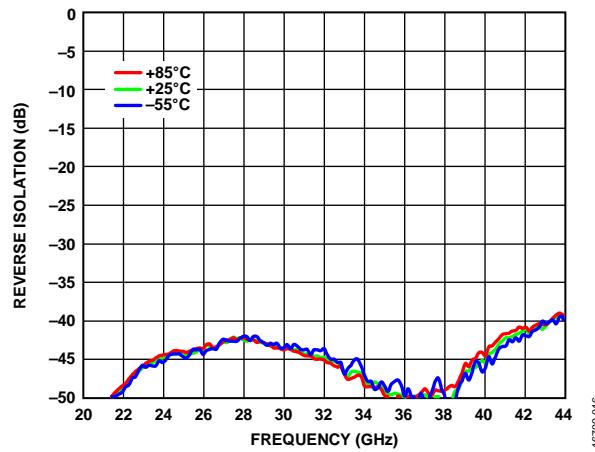
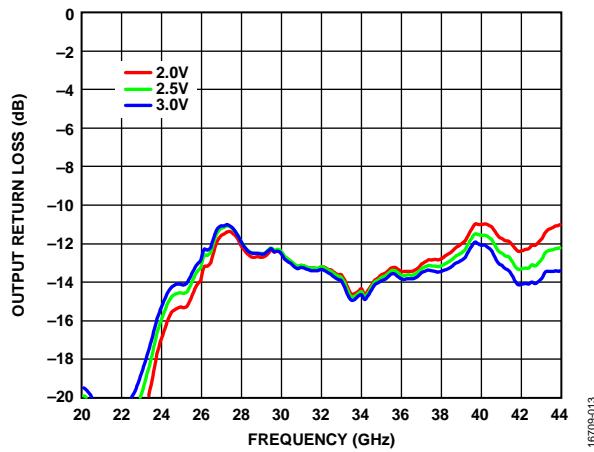
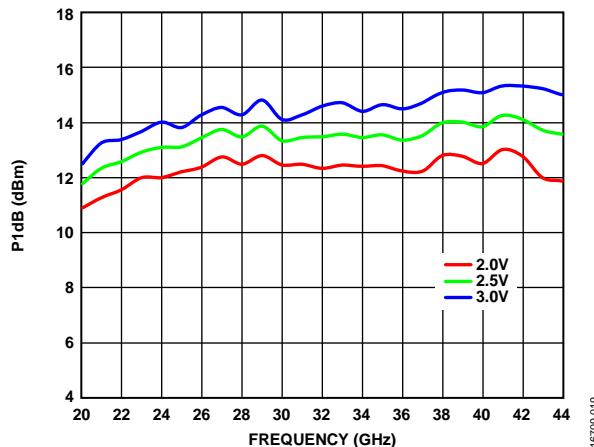


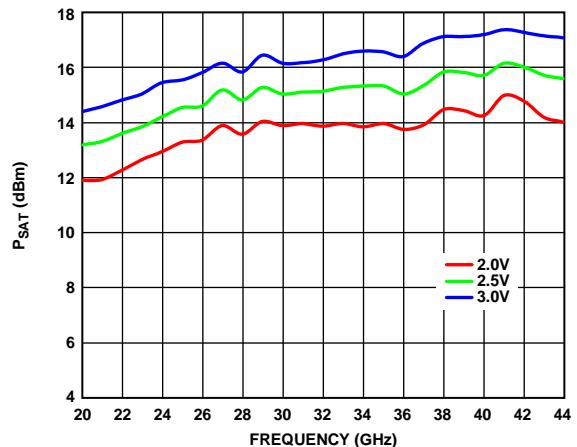
Figure 12. Input Return Loss vs. Frequency for Various Supply Voltages

16709-012

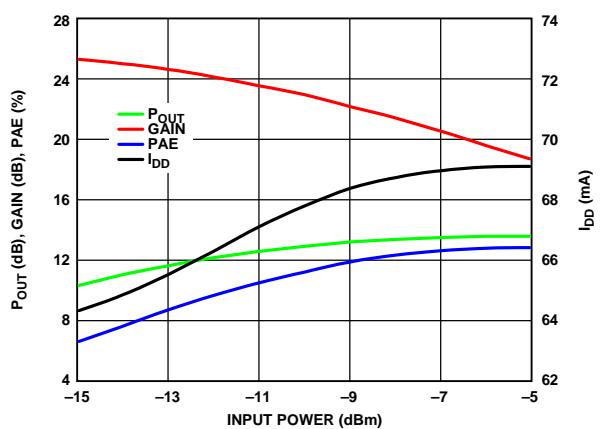




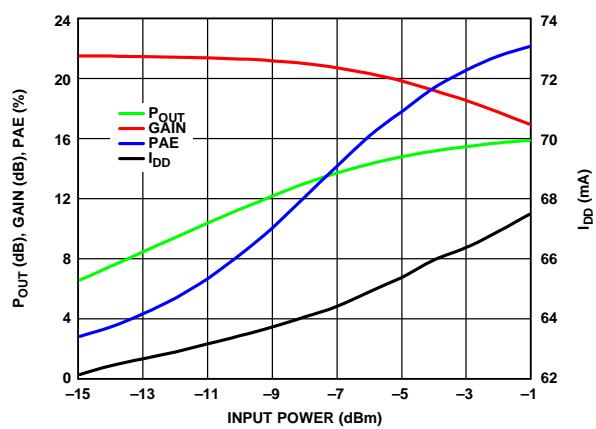
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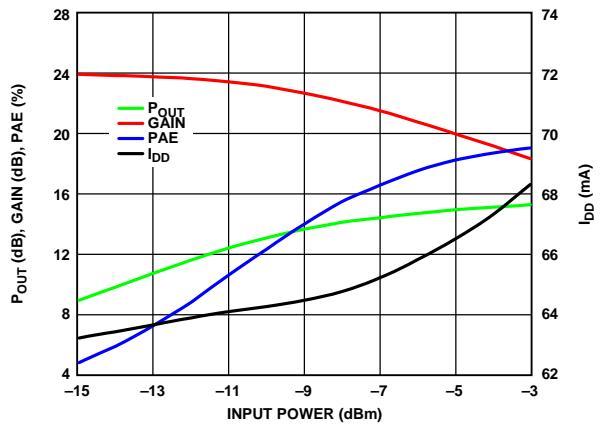
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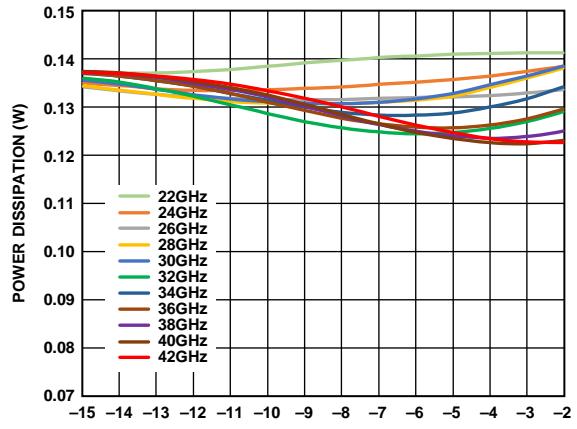
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16708-021



16708-024

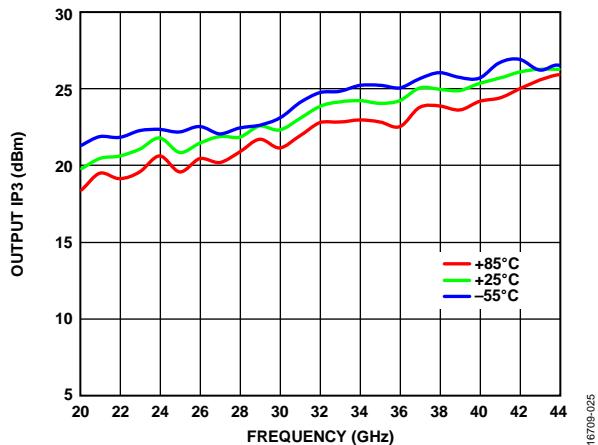


Figure 25. Output IP3 vs. Frequency for Various Temperatures,  
 $P_{\text{OUT/Tone}} = 4 \text{ dBm}$

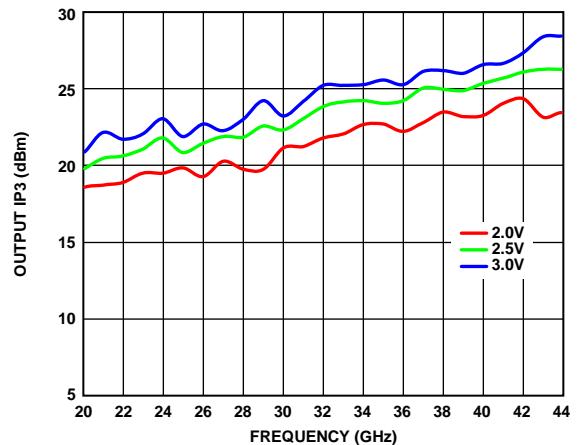


Figure 28. Output IP3 vs. Frequency for Various Supply Voltages,  
 $P_{\text{OUT/Tone}} = 4 \text{ dBm}$

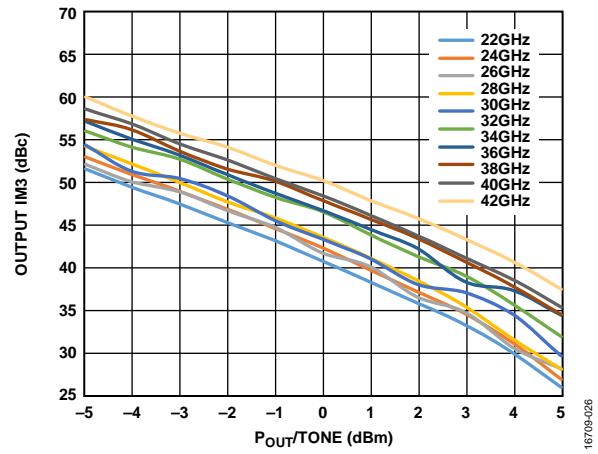


Figure 26. Output Third-Order Intermodulation (IM3) vs.  $P_{\text{OUT/Tone}}$  for Various Frequencies at  $V_{\text{DD}} = 2.0 \text{ V}$

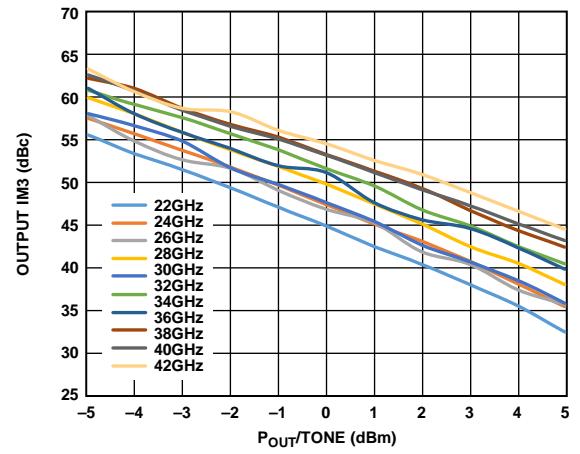


Figure 29. Output IM3 vs.  $P_{\text{OUT/Tone}}$  for Various Frequencies at  $V_{\text{DD}} = 3.0 \text{ V}$

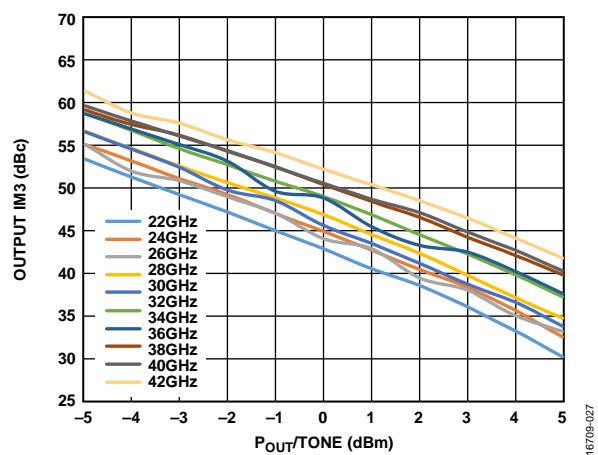
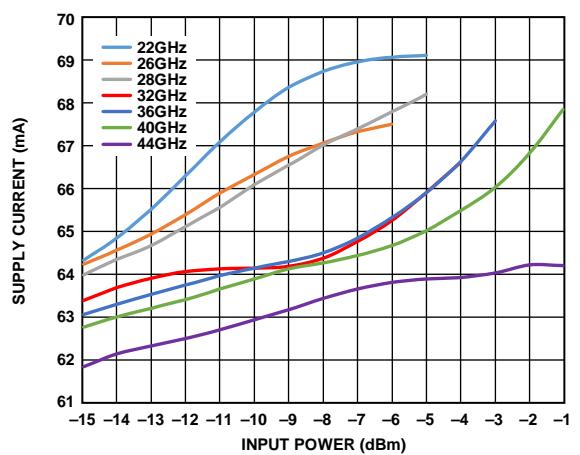


Figure 27. Output IM3 vs.  $P_{\text{OUT/Tone}}$  for Various Frequencies at  $V_{\text{DD}} = 2.5 \text{ V}$



## THEORY OF OPERATION

The HMC1040CHIPS is a GaAs, pHEMT, MMIC, low noise, wideband amplifier. The basic architecture consists of three amplifier stages, optimized for low noise figure and high gain. Self bias removes the need for a negative bias voltage supply for the gate at each stage. A negative voltage is generated across the gate to the source when a typical voltage of 2.5 V is applied on  $V_{DDx}$  at each stage.

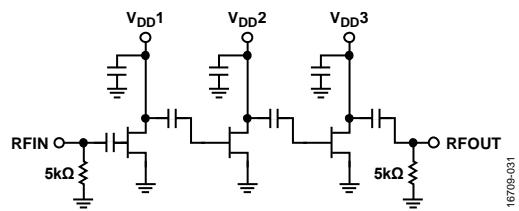


Figure 31. Architecture and Simplified Schematic

16709-031

## APPLICATIONS INFORMATION

### RECOMMENDED BIAS SEQUENCING

Capacitive bypassing is required for  $V_{DDX}$ , as shown in the typical application circuit in Figure 34.

The recommended bias sequence during power-up is as follows:

1. Set  $V_{DDX}$  to 2.5 V (this results in an  $I_{DQ}$  near its specified typical value).
2. Apply the RF input signal.

The recommended bias sequence during power-down is as follows:

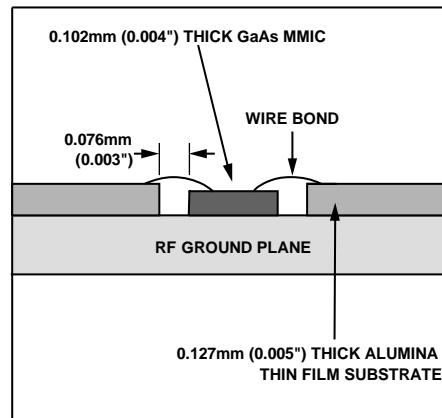
1. Turn off the RF input signal.
2. Set  $V_{DDX}$  to 0 V.

Unless otherwise noted, all measurements and data shown were taken using the typical application circuit (see Figure 34), configured as shown in the assembly diagram (see Figure 35) and biased per the conditions in the Specifications section. The bias conditions shown in the Specifications section are the operating points recommended to optimize the overall performance. Operation using other bias conditions may provide performance that differs from what is shown in this data sheet. To obtain the best performance while not damaging the device, follow the recommended biasing sequence outlined in this section.

## MOUNTING AND BONDING TECHNIQUES FOR MILLIMETERWAVE GaAs MMICS

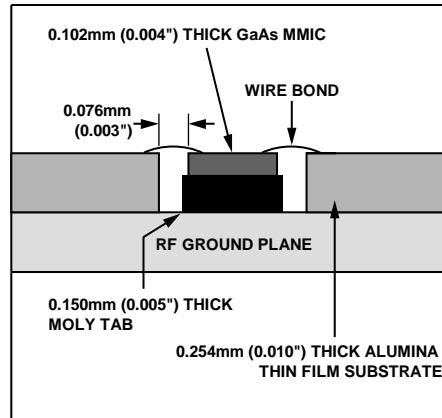
Attach the die directly to the ground plane eutectically or with conductive epoxy (see the Handling Precautions section).

To bring the radio frequency to and from the chip, implementing 50  $\Omega$  transmission lines using a microstrip or coplanar waveguide on 0.127 mm (0.005") thick alumina, thin film substrates is recommended (see Figure 32). When using 0.254 mm (0.010") thick alumina, it is recommended that the die be raised to ensure that the die and substrate surfaces are coplanar. Raise the die 0.150 mm (0.005") to ensure that the surface of the die is coplanar with the surface of the substrate. To accomplish this, attach the 0.102 mm (0.004") thick die to a 0.150 mm (0.005") thick, molybdenum (Mo) heat spreader (moly tab), which can then be attached to the ground plane (see Figure 32 and Figure 33).



16709-032

Figure 32. Die Without the Moly Tab



16709-033

Figure 33. Die With the Moly Tab

Place microstrip substrates as close to the die as possible to minimize bond wire length. Typical die to substrate spacing is 0.076 mm to 0.152 mm (0.003" to 0.006").

### Handling Precautions

To avoid permanent damage, follow these storage, cleanliness, static sensitivity, transient, and general handling precautions:

- Place all bare die in either waffle or gel-based ESD protective containers and then seal the die in an ESD protective bag for shipment. After the sealed ESD protective bag is opened, store all die in a dry nitrogen environment.
- Handle the chips in a clean environment. Do not attempt to clean the chip using liquid cleaning systems.
- Follow ESD precautions to protect against ESD strikes.
- While bias is applied, suppress instrument and bias supply transients. Use shielded signal and bias cables to minimize inductive pickup.
- Handle the chip along the edges with a vacuum collet or with a sharp pair of bent tweezers. The surface of the chip may have fragile air bridges and must not be touched.

## TYPICAL APPLICATION CIRCUIT

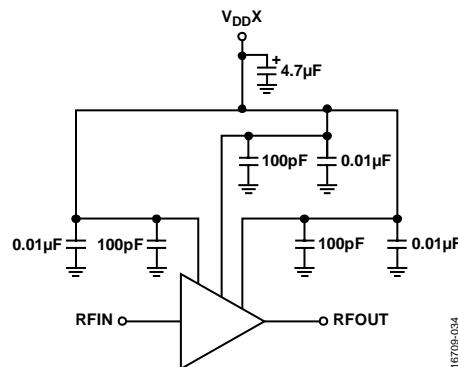


Figure 34. Typical Application Circuit

## ASSEMBLY DIAGRAM

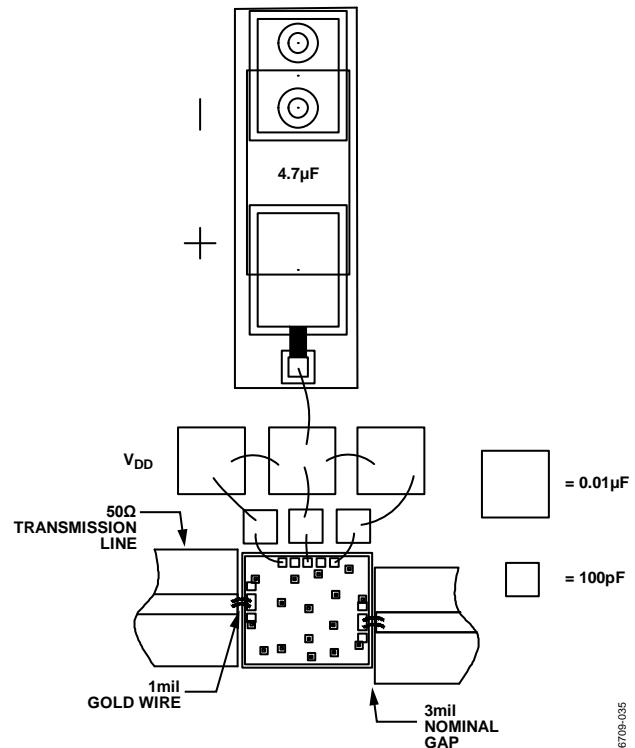
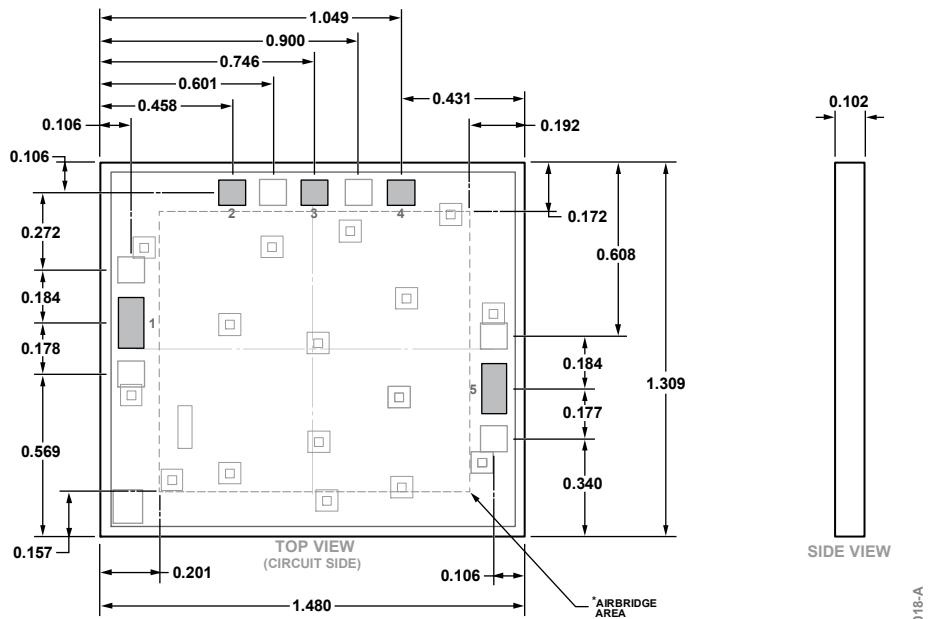


Figure 35. Assembly Diagram

## OUTLINE DIMENSIONS



\*This die utilizes fragile air bridges. Any pickup tools used must not contact this area.

04-26-2018-A

Figure 36. 5-Pad Bare Die [CHIP]

(C-5-6)

Dimensions shown in millimeters

## ORDERING GUIDE

| Model <sup>1</sup> | Temperature Range | Package Description   | Package Option |
|--------------------|-------------------|-----------------------|----------------|
| HMC1040CHIPS       | -55°C to +85°C    | 5-Pad Bare Die [CHIP] | C-5-6          |
| HMC1040CHIPS-SX    | -55°C to +85°C    | 5-Pad Bare Die [CHIP] | C-5-6          |

<sup>1</sup> The HMC1040CHIPS and HMC1040CHIPS-SX are RoHS Compliant Parts.